

White LED Step-Up Converter In Tiny SOT-23 Package

DESCRIPTION

The EUP2584 is a constant current step-up converter specifically designed to drive white LEDs. The Step-up converter topology allows series connection of the white LEDs so the LED currents are identical for uniform brightness. The EUP2584 switches at 1.1MHz, allowing the use of tiny external components. The input and output capacitor can be as small as $1\mu F$, saving space and cost versus alternative solutions. A low 0.25V feedback voltage minimizes power loss in the current setting resistor for better efficiency. EUP2584 is enhanced with Soft-Start function and that could significantly reduce noise induced by capacitor.

The EUP2584 is available in low profile SOT23-6 package.

FEATURES

- 2.6V to 5.5V Input Range
- 27V Output with Over Voltage Protection
- High Efficiency :85 % Typical
- Internal Soft-Start
- PWM Dimming Control
- Internal High Power 30V MOSFET Switch
- Fast 1.1MHz Switching Frequency
- Small, Low-Profile Inductors and Capacitors
- SOT23-6 Package
- RoHS Compliant and 100% Lead (Pb)-Free

APPLICATIONS

- Mobile Phone
- Digital Still Camera
- PDAs, Handheld Computers
- MP3 Players
- GPS Receivers

Typical Application Circuit

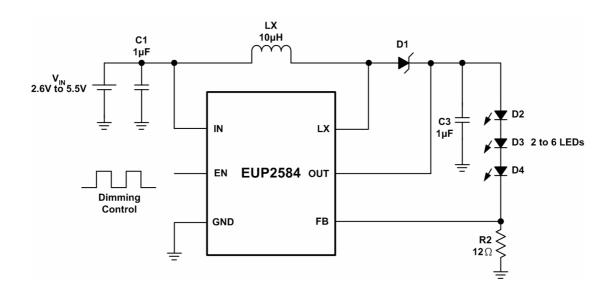


Figure 1. White LED Application



Pin Configurations

Package Type	Pin Configurations			
		IN	OUT	EN
	_			
		6	5	4
SOT23-6				
		• 1	2	3
		LX	GND	FB

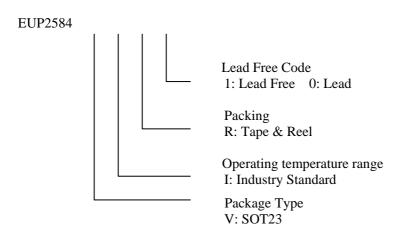
Pin Description

PIN	SOT23-6	DESCRIPTION		
LX	1	Switch Pin. Connect inductor/diode here. Minimize trace area at this pin to reduce EMI.		
GND	2	Common Ground		
FB	3	Feedback Pin. Reference voltage is 0.25V. Connect cathode of lowest LED and resistor here. Calculate resistor value according to the formula: R_{FB} =0.25/ I_{LED}		
EN	4	Chip Enable Pin. Connect to 1.4V or higher to enable device, 0.3V or less to disable device.		
IN	6	Input Supply Voltage		
		Overvoltage Sense. When V_{OUT} is greater than 27V, the internal N-channel MOSFET turns off until V_{OUT} drops below 25V, then the IC reenters start. Connect a 1uF capacitor from OUT to GND.		



Ordering Information

Order Number	Package Type	Marking	Operating Temperature range
EUP2584VIR1	SOT23-6	q0	-40 °C to 85°C



Block Diagram

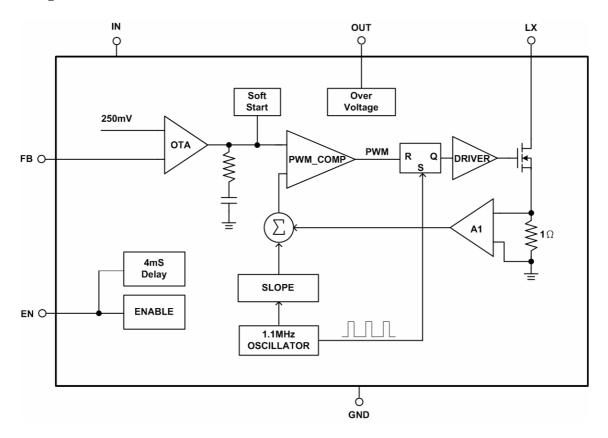


Figure 2.



Absolute Maximum Ratings

•	Supply Voltage ,V _{IN}	-0.3V to 6V
	LX,OUT	-0.3V to 30V
•	The Other Pins	-0.3V to 6V
	Power dissipation, P _D @ T _A =25°C	
	SOT23-6	0.4W
	Package Thermal Resistance	
	SOT23-6, _{JA}	250°C/W
	Maximum Junction Temperature	125°C
	Lead Temperature (Soldering, 10sec.)	260°C
•	Storage Temperature Range	-65°C to 150°C

Operating Conditions

■ Supply Voltage , V_{IN}------- 2.6V to 5.5V

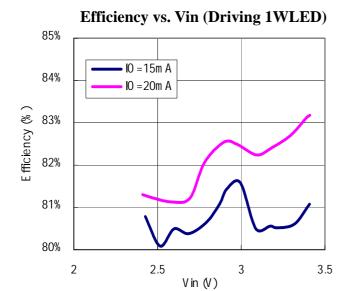
Electrical Characteristics

 $(V_{IN}=\!3.6V,V_{OUT}\!=\!18V,C_{OUT}\!=\!1\mu\text{F},C_{IN}\!=\!1\mu\text{F},RSENSE}\!=\!12 \quad,T_{A}\!=\!-40^{\circ}\text{C to }85^{\circ}\text{C}. \ Unless otherwise noted.}$ Typical values are at $T_{A}\!=\!25^{\circ}\text{C})$

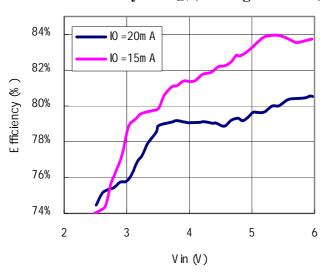
	Donomoton	C 1'4'		EUP2584				
Symbol	Parameter	Conditions	Min	Тур	Max.	Unit		
	Supply Voltage				6	V		
UVLO	Under Voltage Leek Out	V _{IN} Rising	2.2	2.4	2.6	V		
UVLO	Under Voltage Lock Out	V _{IN} Falling	1.95	2.15	2.35	V		
	Maximum Output Voltage	No Switching			30	V		
I_{CC1}	Supply Current	V _{CC} =6V, Continuous Switching		0.8	1.3	mA		
I_{CC2}	Quiescent Current	V _{CC} =6V, FB=1.3V, No Switching		115	150	μΑ		
I_{CC3}	Shutdown current	$V_{CC}=6V, V_{EN}<0.3V$		0.1	1	μΑ		
Oscillator								
Fosc	Operation Frequency		0.8	1.1	1.3	MHz		
Dmax	Maximum Duty Cycle		89	92	96	%		
Reference	Reference Voltage							
V_{FB}	Feedback Voltage		230	250	270	mV		
MOSFET				•		•		
Rds (on)	On resistance of MOSFET			1	1.5			
ILX	Current Limit		0.4	0.8	1.2	A		
Control ar	d Protection		•	•	•	•		
V_{EN1}	Shutdown Voltage		0.3	0.7		V		
V _{EN2}	Enable Voltage			0.7	1.4	V		
I _{EN}	EN Pin Pull Low Current				0.1	μA		
	OVP Threshold	Falling	24.5	25.7	26.5	1		
OVP		Rising	26.1	27.3	28.1	V		



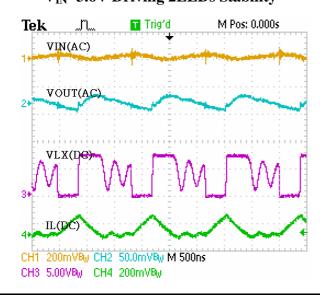
Typical Operating Characteristics



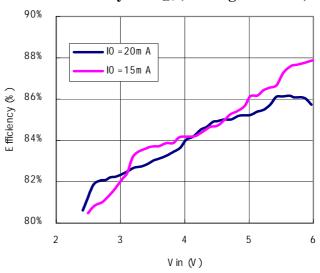
Efficiency vs. V_{IN} (Driving 6WLEDs)



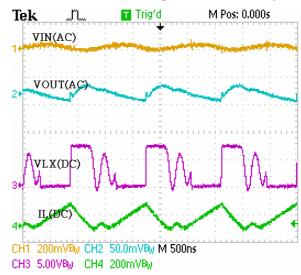
V_{IN}=3.6V Driving 2LEDs Stability



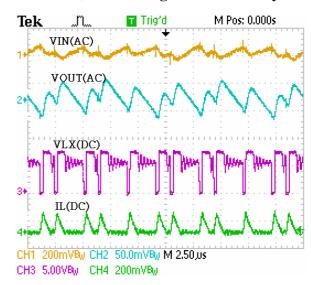
Efficiency vs. V_{IN} (Driving 3WLEDs)



V_{IN}=2.6V Driving 2LEDs Stability

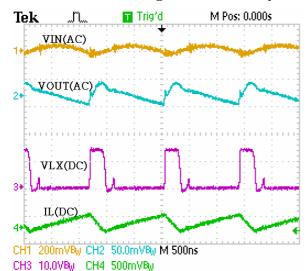


V_{IN}=5V Driving 2LEDs Stability

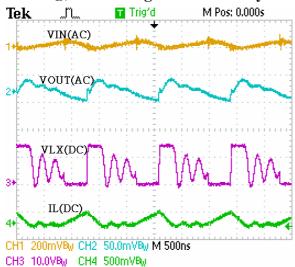




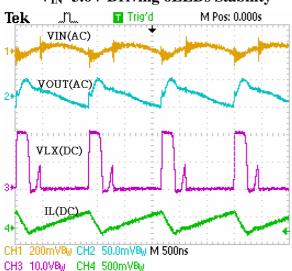
V_{IN}=2.6V Driving 4LEDs Stability



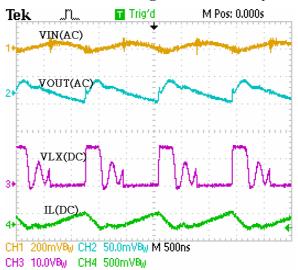
V_{IN}=5V Driving 4LEDs Stability



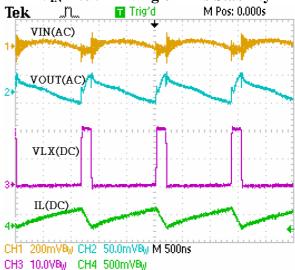
V_{IN}=3.6V Driving 6LEDs Stability



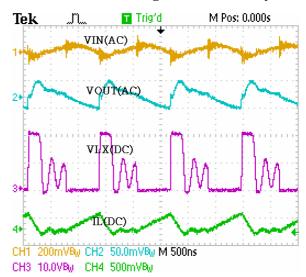
V_{IN}=3.6V Driving 4LEDs Stability



V_{IN}=2.6V Driving 6LEDs Stability

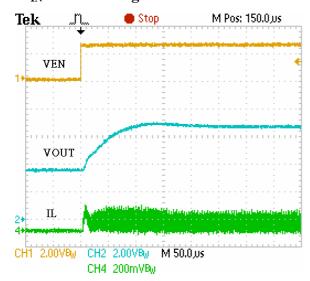


V_{IN}=5V Driving 6LEDs Stability

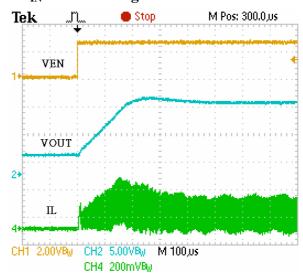




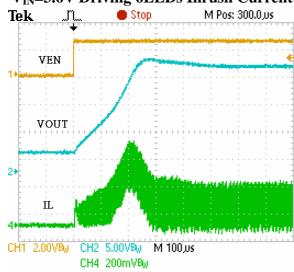
V_{IN}=3.6V Driving 2LEDs Inrush Current



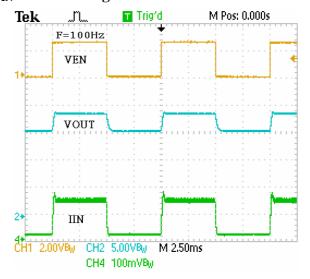
V_{IN}=3.6V Driving 4LEDs Inrush Current



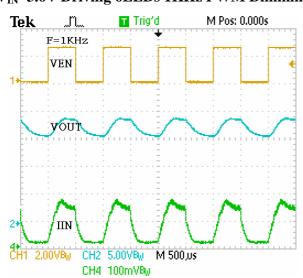
V_{IN}=3.6V Driving 6LEDs Inrush Current



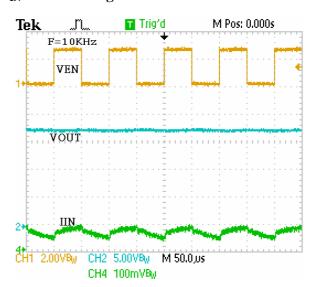
V_{IN}=3.6V Driving 6LEDs 100Hz PWM Dimming



V_{IN}=3.6V Driving 6LEDs 1KHz PWM Dimming

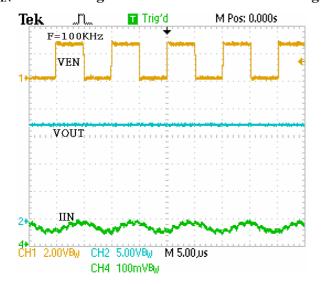


V_{IN}=3.6V Driving 6LEDs 10KHz PWM Dimming





V_{IN} =3.6V Driving 6LEDs 100KHz PWM Dimming





Application Information

LED Current Control

The EUP2584 regulates the LED current by setting the current sense resistor (R2) connecting to feedback and ground. The internal feedback reference voltage is 0.25V. The LED current can be set from following equation easily.

$$R2 = \frac{0.25 \text{V}}{I_{LED}}$$
 -----(1)

In order to have an accurate LED current, precision resistors are preferred (1% is recommended). The table for R2 selection is shown below.

R2 Resistor Value selection

I _{LED} (mA)	R2()			
5	50			
10	25			
12	21			
15	16.7			
20	12.5			

Inductor Selection

The recommended value of inductor for 2 to 6 WLEDs applications are 4.7 to $22\mu H$. Small size and better efficiency are the major concerns for portable device, such as EUP2584 used for mobile phone. The inductor should have low core loss at 1.1MHz and low DCR for better efficiency. To avoid inductor saturation current rating should be considered.

Dimming Control

a. Using a PWM Signal to EN Pin

For controlling the LED brightness, the EUP2584 can perform the dimming control by applying a PWM signal to EN pin. The internal soft-start and wide range dimming frequency from 100Hz to 100KHz can insignificantly reduce audio noise when dimming. The average LED current is proportional to the PWM signal duty cycle. The magnitude of the PWM signal should be higher than the maximum enable voltage of EN pin, in order to let the dimming control perform correctly.

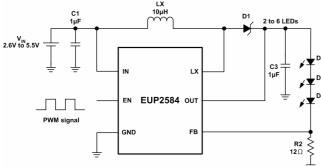


Figure 3. PWM Dimming Control Using the EN Pin

b. Using a DC Voltage

Using a variable DC voltage to adjust the brightness is a popular method in some applications. The dimming control using a DC voltage circuit is shown in Figure 4. According to the Superposition Theorem, as the DC voltage increases, the voltage contributed to V_{FB} increases and the voltage drop on R2 decreases, i.e. the LED current decreases. For example, if the V_{DC} range is from 0V to 2.8V, the selection of resistors in Figure 4 sets dimming control of LED current from 20mA to 0mA.

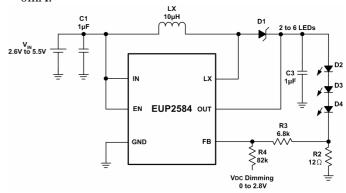


Figure 4. Dimming Control Using a DC Voltage

c. Using a Filtered PWM signal

Another common application is using a filtered PWM signal as an adjustable DC voltage for LED dimming control. A filtered PWM signal acts as the DC voltage to regulate the output current. The recommended application circuit is shown in the Figure 6. In this circuit, the output ripple depends on the frequency of PWM signal. For smaller output voltage ripple (<100mV), the recommended frequency of 2.8V PWM signal should be above 2kHz. To fix the frequency of PWM signal and change the duty cycle of PWM signal can get different output current. According to the application circuit of Figure 5, output current is from 20.5mA to 5.5mA by adjusting the PWM duty cycle from 10% to 90%.

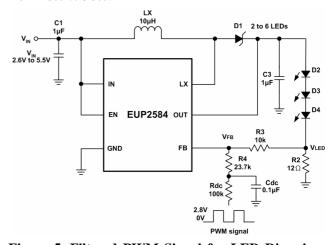


Figure 5. Filtered PWM Signal for LED Dimming Control



Constant Output Voltage for Backlight of Main Panel and Flashlight

Figure 6 is an application of EUP2584 for backlight of main panel and flashlight. Setting the divider-resistors (R1 & R2) is to get a constant output voltage that depends on the forward voltage and the numbers of series-LEDs. There are three kinds of mode controlled by the switches - backlight mode /flashlight mode /backlight + flashlight mode. It can turn on backlight or flashlight at one time or both at the same time. Applying different duty cycle of PWM signal above 22kHz to backlight's switch can also control the brightness. The following formula (2)(3) can determine R3 and R4.

R3 =
$$\frac{V_{OUT} - 3VFb - V_{DS}}{Ib}$$
 -----(2)

$$R4 = \frac{V_{OUT} - 3VFf - V_{DS}}{If} - \dots (3)$$

$$V_{DS} = Ib \times R_{DS(ON)}$$
 -----(4)

Constant Output Voltage for Backlight of Main Panel and Keypad

Figure 7 is another application of EUP2584 for backlight and keypad. Setting the divider-resistors (R1 & R2) is to get a constant output voltage that depends on the forward voltage and the numbers of series-LEDs. It can turn on backlight of main panel and keypad at the same time. Applying different duty cycle of PWM signal above 22kHz to the backlight's switch can also control the brightness of main panel's backlight. The keypad's backlight will keep the same brightness during the dimming control of main panel. Otherwise the brightness of keypad's s backlight can also change during the dimming control of main panel by using the application circuit as figure 4. The following formula (5)(6) can determine the resistors of Figure 8.

$$R3 = \frac{V_{OUT} - 3VFb - V_{DS}}{Ib} - \dots (5)$$

$$R4 = R5 = R6 = \frac{V_{OUT} - 3VFk}{Ik}$$
 ----(6)

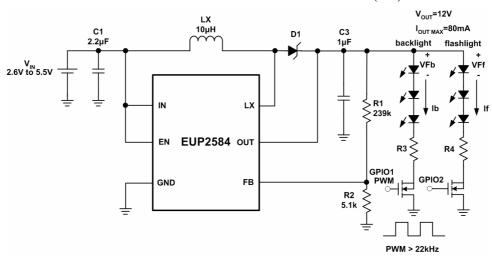


Figure 6. Constant Output Voltage for Backlight and Flashlight

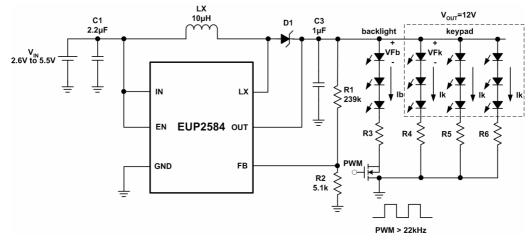
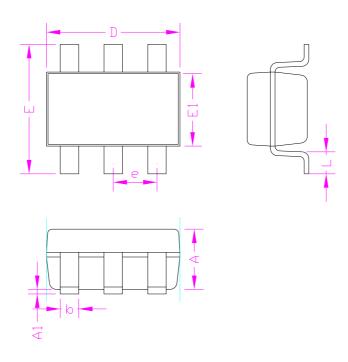


Figure 7. Constant Output Voltage for Backlight and Keypad



Packaging Information

SOT23-6



SYMBOLS	MILLIM	IETERS	INCHES		
STIVIBOLS	MIN.	MAX.	MIN.	MAX.	
Α	-	1.45	-	0.057	
A1	0.00	0.15	0.000	0.006	
b	0.30	0.50	0.012	0.020	
D	2.90		0.114		
E1	1.	1.60		0.063	
е	0.	95	0.037		
Е	2.60	3.00	0.102	0.118	
L	0.3	0.60	0.012	0.024	

