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# Single/Dual/Quad High-Speed, Fast-Settling, High Output Current Operational Amplifier 


#### Abstract

General Description The MAX408/428/448 are high speed general purpose monolithic operational amplifiers in a single, dual or quad package, that are useful for signal frequencies extending into the video range. These Op Amps function in gain configurations greater-than or equal-to 3. High output current allows large capacitive loads to be driven at high speeds.

Open-loop voltage gain of 10k V/V and high slew rate of $90 \mathrm{~V} / \mu \mathrm{s}$ make the MAX408/428/448 ideal for analog amplification and high speed signal processing. 100 MHz gain bandwidth and a $\pm 0.1 \%$ settling time of I50ns make each amplifier ideal for fast data conversion systems. The amplifiers are capable of driving back terminated transmission lines of $75 \Omega$ with amplitudes of 5 V peak-to-peak. Along with the high speed and output drive capability, a 35nA offset current and trimmable offset voltage make the MAX408/428/448 optimal for signal conditioning applications where accuracy must be maintained.


## Applications

Video Amplifiers
Test Equipment
Waveform Generators
Video Distribution
Pulse Amplifiers

- Fast Settling Time: $\pm 0.1 \%$ In 150 ns
- High Slew Rate: 90V/us
- Large Gain Bandwidth: 100MHz
- Full Power Bandwidth: 4.8 MHz at 6 V p-p
- Ease of Use: Internally Compensated for

AcL $\geq 3$ with $50^{\circ}-60^{\circ}$ Phase Margin

- Low Supply Voltage Operation: $\pm 4 \mathrm{~V}$
- Wide Input Voltage Range: Within 1.5V of V+ and 0.5 V of V -
- Minimal Crosstalk: >90dB Separation (MAX428/448)
- Short Circuit Protection

| PART | TEMP. RANGE | PIN-PACKAGE |
| :---: | :---: | :---: |
| MAX408ACPA | $0^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}$ | 8 Lead Plastic DIP |
| MAX408ACSA | $0^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}$ | 8 Lead Small Outline |
| MAX408CPA | $0^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}$ | 8 Lead Plastic DIP |
| MAX408CSA | $0^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}$ | 8 Lead Small Outline |
| MAX408C/D | $0^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}$ | Dice |

Pin Configurations


For free samples \& the latest literature: http://www.maxim-ic.com, or phone 1-800-998-8800. For small orders, phone 1-800-835-8769.

## Single/Dual/Quad High-Speed, Fast-Settling, High Output Current Operational Amplifier

## ABSOLUTE MAXIMUM RATINGS

Supply Voltages $\qquad$ $+6 \mathrm{~V}$
Differential Input Voltage $+9 \mathrm{~V}$
Common Mode Input Voltage ......................................|Vs| -0.5 V
Output Short Circuit Current Duration $\qquad$ ndefinite Continuous Power Dissipation ( $\mathrm{T}_{\mathrm{A}}=+70^{\circ} \mathrm{C}$ )
8-Pin Plastic DIP (derate $9.09 \mathrm{~mW} /{ }^{\circ} \mathrm{C}$ above $+70^{\circ} \mathrm{C}$ ) $\ldots . .727 \mathrm{~mW}$
8 -Pin SO (derate $5.88 \mathrm{~mW} /{ }^{\circ} \mathrm{C}$ above $+70^{\circ} \mathrm{C}$ )................. 471 mW

14-Pin Plastic DIP (derate $10.00 \mathrm{~mW} /{ }^{\circ} \mathrm{C}$ above $+70^{\circ} \mathrm{C}$ ) $\qquad$ .800 mW 14-Pin SO (derate $8.33 \mathrm{~mW} /{ }^{\circ} \mathrm{C}$ above $+70^{\circ} \mathrm{C}$ )............... 667 mW Operating Temperature Range
Commercial (MAX4_8AC/C) . $0^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}$
Storage Temperature Range $-65^{\circ} \mathrm{C}$ to $+150^{\circ} \mathrm{C}$
Lead Temperature (Soldering, 60 seconds)................... $+300^{\circ} \mathrm{C}$

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## ELECTRICAL CHARACTERISTICS-MAX408

( $\mathrm{V}_{\mathrm{S}}= \pm 5 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$, unless otherwise noted.)

| PARAMETER | SYMBOL | CONDITIONS | MAX408C |  |  | MAX408AC |  |  | UNITS |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | MIN | TYP | MAX | MIN | TYP | MAX |  |
| Input Offset Voltage | Vos | $\begin{aligned} & \mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C} \\ & 0^{\circ} \mathrm{C} \leq \mathrm{T}_{\mathrm{A}} \leq 70^{\circ} \mathrm{C} \end{aligned}$ |  | $\begin{aligned} & 5 \\ & 8 \end{aligned}$ | $\begin{aligned} & \hline 12 \\ & 16 \end{aligned}$ |  | 3 | $\begin{gathered} 6 \\ 10 \end{gathered}$ | mV |
| Average Offset Voltage Drift | $\Delta \mathrm{VOS} / \Delta \mathrm{T}$ | $0^{\circ} \mathrm{C} \leq \mathrm{T}_{\mathrm{A}} \leq 70^{\circ} \mathrm{C}$ |  | 20 |  |  | 20 |  | $\mu \mathrm{V} /{ }^{\circ} \mathrm{C}$ |
| Input Bias Current | IB |  |  | 650 | 1100 |  | 650 | 1100 | nA |
| Input Offset Current | los | $\begin{aligned} & \mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C} \\ & 0^{\circ} \mathrm{C} \leq \mathrm{T}_{\mathrm{A}} \leq 70^{\circ} \mathrm{C} \end{aligned}$ |  | $\begin{aligned} & 35 \\ & 70 \end{aligned}$ | $\begin{aligned} & 120 \\ & 200 \end{aligned}$ |  | $\begin{aligned} & 35 \\ & 70 \end{aligned}$ | $\begin{aligned} & 120 \\ & 200 \end{aligned}$ | nA |
| Input Common Mode Range | Vсм |  | $\begin{gathered} +3 \\ -4 \end{gathered}$ | $\begin{gathered} +3.5 \\ -4.5 \end{gathered}$ |  | $\begin{aligned} & +3 \\ & -4 \end{aligned}$ | $\begin{gathered} +3.5 \\ -4.5 \end{gathered}$ |  | V |
| Differential Input Resistance | Rind | (Note 1) | 3 | 10 |  | 3 | 10 |  | $\mathrm{M} \Omega$ |
| Common Mode Input Resistance | Rinc | ( Note 1) | 4 | 8 |  | 4 | 8 |  | $\mathrm{M} \Omega$ |
| Differential Input Capacitance | CIND |  |  | 2 |  |  | 2 |  | pF |
| Common Mode Input Capacitance | Cinc |  |  | 3 |  |  | 3 |  | pF |
| Input Voltage Noise | eN | BW $=10 \mathrm{~Hz}$ to 100 kHz |  | 12 |  |  | 12 |  | $\mu \mathrm{V}$ RMS |
| Open Loop Voltage Gain | Av | VOUT $= \pm 3 \mathrm{~V}, \mathrm{R}_{\mathrm{L}}=2 \mathrm{k} \Omega$ | 2 | 5 |  | 5 | 10 |  | $\mathrm{V} / \mathrm{mV}$ |
| Output Voltage Swing | Vout | $\begin{aligned} & \mathrm{R} \mathrm{~L}=2 \mathrm{k} \Omega \\ & \mathrm{RL}=150 \Omega \end{aligned}$ | $\begin{aligned} & \pm 3.5 \\ & \pm 2.0 \end{aligned}$ | $\pm 2.4$ |  | $\begin{aligned} & \pm 3.5 \\ & \pm 2.5 \end{aligned}$ | $\pm 2.7$ |  | V |
| Power Supply Current | Is |  |  | 7 | 10 |  | 7 | 10 | mA |
| Common Mode Rejection Ratio | CMRR | $\mathrm{V}_{\text {CM }}= \pm 2 \mathrm{~V}$ | 60 | 70 |  | 60 | 70 |  | dB |
| Power Supply Rejection Ratio | PSRR | $\Delta \mathrm{VPS}= \pm 0.5 \mathrm{~V}$ | 60 | 66 |  | 60 | 66 |  | dB |
| Slew Rate (Note 1) | SR | 10-90\% of Leading Edge (Figure 1) | 60 | 90 |  | 60 | 90 |  | V/ $/ \mathrm{S}$ |
| Settling Time | ts | To $\pm 0.1 \%( \pm 4 \mathrm{mV})$ of Final Value (Figure 1) (Note 1) |  | 150 | 200 |  | 150 | 200 | ns |
| Gain Bandwidth Product | GBW |  |  | 100 |  |  | 100 |  | MHz |

Note 1: Not tested, guaranteed by design.

## Single/Dual/Quad High-Speed, Fast-Settling, High Output Current Operational Amplifier

## ELECTRICAL CHARACTERISTICS—MAX428

( $\mathrm{V}_{\mathrm{S}}= \pm 5 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$, unless otherwise noted.)

| PARAMETER | SYMBOL | CONDITIONS | MAX428C |  |  | MAX428AC |  |  | UNITS |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | MIN | TYP | MAX | MIN | TYP | MAX |  |
| Input Offset Voltage | Vos | $\begin{aligned} & \mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C} \\ & 0^{\circ} \mathrm{C} \leq \mathrm{T}_{\mathrm{A}} \leq 70^{\circ} \mathrm{C} \end{aligned}$ |  | $\begin{aligned} & 5 \\ & 8 \end{aligned}$ | $\begin{aligned} & 12 \\ & 16 \end{aligned}$ |  | $\begin{aligned} & 3 \\ & 5 \end{aligned}$ | $\begin{gathered} 6 \\ 10 \end{gathered}$ | mV |
| Average Offset Voltage Drift | $\Delta \mathrm{Vos} / \Delta \mathrm{T}$ | $0^{\circ} \mathrm{C} \leq \mathrm{T}_{\mathrm{A}} \leq 70^{\circ} \mathrm{C}$ |  | 20 |  |  | 20 |  | $\mu \mathrm{V} /{ }^{\circ} \mathrm{C}$ |
| Input Bias Current | IB | $\begin{aligned} & \mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C} \\ & 0^{\circ} \mathrm{C} \leq \mathrm{T}_{\mathrm{A}} \leq 70^{\circ} \mathrm{C} \end{aligned}$ |  | 650 | $\begin{aligned} & 1100 \\ & 1700 \end{aligned}$ |  | 650 | $\begin{aligned} & 1100 \\ & 1700 \end{aligned}$ | mA |
| Input Offset Current | los |  |  | 35 | 120 |  | 35 | 120 | nA |
| Input Common Mode Range | VCM |  | $\begin{aligned} & +3 \\ & -4 \end{aligned}$ | $\begin{aligned} & +3.5 \\ & -4.5 \end{aligned}$ |  | $\begin{aligned} & +3 \\ & -4 \end{aligned}$ | $\begin{gathered} +3.5 \\ -4.5 \end{gathered}$ |  | V |
| Differential Input Resistance | Rind | (Note 1) | 3 | 10 |  | 3 | 10 |  | $\mathrm{M} \Omega$ |
| Common Mode Input Resistance | Rinc | (Note 1) | 4 | 8 |  | 4 | 8 |  | $\mathrm{M} \Omega$ |
| Differential Input Capacitance | CIND |  |  | 2 |  |  | 2 |  | pF |
| Common Mode Input Capacitance | Cinc |  |  | 3 |  |  | 3 |  | pF |
| Input Voltage Noise | eN | BW $=10 \mathrm{~Hz}$ to 100 kHz |  | 12 |  |  | 12 |  | $\mu \mathrm{V}$ RMS |
| Open Loop Voltage Gain | Av | VOUT $= \pm 3 \mathrm{~V}, \mathrm{RL}=2 \mathrm{k} \Omega$ | 2 | 5 |  | 5 | 10 |  | $\mathrm{V} / \mathrm{mV}$ |
| Output Voltage Swing | Vout | $\begin{aligned} & \mathrm{RL}=2 \mathrm{k} \Omega \\ & \mathrm{RL}=150 \Omega \end{aligned}$ | $\begin{aligned} & \pm 3.5 \\ & \pm 2.0 \end{aligned}$ | $\pm 2.4$ |  | $\begin{aligned} & \pm 3.5 \\ & \pm 2.5 \end{aligned}$ | $\pm 2.7$ |  | V |
| Power Supply Current (Both Amplifiers) | Is |  |  | 15 | 20 |  | 15 | 20 | mA |
| Common Mode Rejection Ratio | CMRR | $\mathrm{V}_{\text {CM }}= \pm 2 \mathrm{~V}$ | 60 | 70 |  | 60 | 70 |  | dB |
| Power Supply Rejection Ratio | PSRR | $\Delta \mathrm{V}$ PS $= \pm 0.5 \mathrm{~V}$ | 60 | 66 |  | 60 | 66 |  | dB |
| Slew Rate (Note 1) | SR | 10-90\% of Leading Edge (Figure 1) | 60 | 90 |  | 60 | 90 |  | V/ $\mu \mathrm{S}$ |
| Settling Time | ts | To $\pm 0.1 \%( \pm 4 \mathrm{mV})$ of Final Value (Figure 1) (Note 1) |  | 150 | 200 |  | 150 | 200 | ns |
| Gain Bandwidth Product | GBW |  |  | 100 |  |  | 100 |  | MHz |

Note 1: Not tested, guaranteed by design.

## Single/Dual/Quad High-Speed, Fast-Settling, High Output Current Operational Amplifier

ELECTRICAL CHARACTERISTICS—MAX448
( $\mathrm{V}_{\mathrm{S}}= \pm 5 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$, unless otherwise noted.)

| PARAMETER | SYMBOL | CONDITIONS | MAX408C |  |  | MAX408AC |  |  | UNITS |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | MIN | TYP | MAX | MIN | TYP | MAX |  |
| Input Offset Voltage | Vos | $\begin{aligned} & \mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C} \\ & 0^{\circ} \mathrm{C} \leq \mathrm{T}_{\mathrm{A}} \leq 70^{\circ} \mathrm{C} \end{aligned}$ |  | $\begin{aligned} & 5 \\ & 8 \end{aligned}$ | $\begin{aligned} & 12 \\ & 16 \end{aligned}$ |  | $\begin{aligned} & 3 \\ & 5 \end{aligned}$ | $\begin{gathered} 6 \\ 10 \end{gathered}$ | mV |
| Average Offset Voltage Drift | $\Delta \mathrm{V}$ os/ $/ \mathrm{T}$ | $0^{\circ} \mathrm{C}<\mathrm{T}_{\mathrm{A}} \leq 70^{\circ} \mathrm{C}$ |  | 20 |  |  | 20 |  | $\mu \mathrm{V} /{ }^{\circ} \mathrm{C}$ |
| Input Bias Current | IB | $\begin{aligned} & \mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C} \\ & 0^{\circ} \mathrm{C} \leq \mathrm{T}_{\mathrm{A}} \leq 70^{\circ} \mathrm{C} \end{aligned}$ |  | 650 | $\begin{aligned} & 1100 \\ & 1700 \end{aligned}$ |  | 650 | $\begin{aligned} & 1100 \\ & 1700 \end{aligned}$ | nA |
| Input Offset Current | los |  |  | 35 | 120 |  | 35 | 120 | nA |
| Input Common Mode Range | VCM |  | $\begin{gathered} +3 \\ -4 \end{gathered}$ | $\begin{gathered} +3.5 \\ -4.5 \end{gathered}$ |  | $\begin{gathered} +3 \\ -4 \end{gathered}$ | $\begin{gathered} +3.5 \\ -4.5 \end{gathered}$ |  | V |
| Differential Input Resistance | Rind | (Note 1) | 3 | 10 |  | 3 | 10 |  | $\mathrm{M} \Omega$ |
| Common Mode Input Resistance | Rinc | (Note 1) | 4 | 8 |  | 4 | 8 |  | $\mathrm{M} \Omega$ |
| Differential Input Capacitance | CIND |  |  | 2 |  |  |  |  | pF |
| Common Mode Input Capacitance | Cinc |  |  | 3 |  |  | 3 |  | pF |
| Input Voltage Noise | eN | $B W=10 \mathrm{~Hz}$ to 100 kHz |  | 12 |  |  | 12 |  | $\mu \mathrm{V}_{\text {RMS }}$ |
| Open Loop Voltage Gain | Av | VOUT $= \pm 3 \mathrm{~V}, \mathrm{R}_{\mathrm{L}}=2 \mathrm{k} \Omega$ | 2 | 5 |  | 4 | 10 |  | $\mathrm{V} / \mathrm{mV}$ |
| Output Voltage Swing | Vout | $\begin{aligned} & \mathrm{RL}=2 \mathrm{k} \Omega \\ & \mathrm{RL}=150 \Omega \end{aligned}$ | $\begin{aligned} & \pm 3.5 \\ & \pm 2.0 \end{aligned}$ | $\pm 2.4$ |  | $\begin{aligned} & \pm 3.5 \\ & \pm 2.5 \end{aligned}$ | $\pm 2.7$ |  | V |
| Power Supply Current (All Four Amplifiers) | Is |  |  | 30 | 40 |  | 30 | 40 | mA |
| Power Supply Rejection Ratio | PSRR | $\Delta \mathrm{VPS}= \pm 0.5 \mathrm{~V}$ | 60 | 66 |  | 60 | 66 |  | dB |
| Common Mode Rejection Ratio | CMRR | $\mathrm{V}_{\mathrm{CM}}= \pm 2 \mathrm{~V}$ | 60 | 70 |  | 60 | 70 |  | dB |
| Slew Rate (Note 1) | SR | 10-90\% of Leading Edge (Figure 1) | 60 | 90 |  | 60 | 90 |  | $\mathrm{V} / \mu \mathrm{S}$ |
| Settling Time | ts | To $\pm 0.1 \%( \pm 4 \mathrm{mV})$ of Final Value (Figure 1) (Note 1) |  | 150 | 200 |  | 150 | 200 | ns |
| Gain Bandwidth Product | GBW |  |  | 100 |  |  | 100 |  | MHz |

Note 1: Not tested, guaranteed by design.

## AC CHARACTERISTICS—MAX408/428/448

( $\mathrm{V}_{\mathrm{S}}= \pm 5 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$, unless otherwise specified.)

| PARAMETER | SYMBOL | CONDITIONS | MAX4XXC |  |  | MAX4XXC |  |  | UNITS |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | MIN | TYP | MAX | MIN | TYP | MAX |  |
| Small Signal Rise/Fall Time | tr/tf | $\begin{aligned} & \text { eo }= \pm 100 \mathrm{mV} \\ & 10-90 \% \text { (Figure 1) } \end{aligned}$ |  | 7 |  |  | 7 |  | ns |
| Full Power Bandwidth | BWFP | $\begin{aligned} & \mathrm{RL}=2 \mathrm{k} \Omega, \mathrm{CL}_{\mathrm{L}}=50 \mathrm{pF} \\ & \text { VOUT }=6 \mathrm{Vp}-\mathrm{p} \end{aligned}$ |  | 4.8 |  |  | 4.8 |  | MHz |
| Amp-Amp Crosstalk (MAX428/448) |  | Input Referenced $f=10 \mathrm{kHz}$ |  | -96 |  |  | -96 |  | dB |

$\qquad$

## Single/Dual/Quad High-Speed, Fast-Settling, High Output Current Operational Amplifier

Typical Operating Characteristics
( $\mathrm{V}_{\mathrm{S}}= \pm 5, \mathrm{~T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$, unless otherwise stated and apply for each individual op amp where applicable.)


## Single/Dual/Quad High-Speed, Fast-Settling, High Output Current Operational Amplifier

Typical Operating Characteristics
( $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$, unless otherwise noted.)







## Single/Dual/Quad High-Speed, Fast-Settling, High Output Current Operational Amplifier



Figure 1A. Settling Time and Slew Rate Test Circuit


Figure 1B. Large Signal Response


Figure 1C. Small Signal Response

## Application Information

## AC Characteristics

The 35 MHz 10 dB crossover point of the MAX408/ 428/448 is achieved without feed forward compensation, a technique which can produce long tails in the recovery characteristic. The single pole rolloff follows the classic $20 \mathrm{~dB} / \mathrm{dec} a d e$ slope to frequencies approaching 50 MHz . The $10 \mathrm{~dB}(3.2 \mathrm{~V} / \mathrm{V})$ phase margin of $50^{\circ}$, even with a capacitive load of 50 pF , gives stable and predictable performance down to non-inverting gain configurations of approximately 3V/V (inverting gains of $-2 \mathrm{~V} / \mathrm{V}$ ). At frequencies beyond 50 MHz , the $20 \mathrm{~dB} / \mathrm{decade}$ slope is disturbed by an output stage zero, the damping factor of which is dependent upon the RL, CL load combination. This results in loss of gain

# Single/Dual/Quad High-Speed, Fast-Settling, High Output Current Operational Amplifier 

margin (gain at loop phase $=360^{\circ}$ ) at frequencies of 70 to 100 MHz which at a gain margin of $5 \mathrm{~dB}\left(\mathrm{R}_{\mathrm{L}}=2 \mathrm{k}, \mathrm{C}_{\mathrm{L}}\right.$ $=5 p F$ ) results in a peak in the gain of 3 amplifier configurations as shown in Figures 3 and 4.
Figure 3 shows a blow up of the open loop characteristics in the 10 MHz to 200 MHz frequency range, as well as the corresponding closed loop characteristics for a gain of three non-inverting amplifier at similar load conditions. It should be noted that the open loop characteristic does not show the additional phase shift covered by the input capacitance pole. This is why the closed loop peaking at 30 to 40 MHz is greater than what would be expected from the 50 to 60 degrees of phase margin indicated by the open loop characteristics. Corresponding small signal step response characteristics show well-behaved pulse waveforms with 16-33\% overshoot.
The input capacitive pole can be neutralized by adding a feedback capacitor to $\mathrm{R}_{2}$. The value of capacitance is selected according to $\mathrm{R}_{1} \mathrm{CIN}=\mathrm{R}_{2} \mathrm{CFB}_{\mathrm{FB}}$, where $\mathrm{CIN}_{\text {IN }}$ is the sum of the common mode and differential input capacitance $\approx 5 p$ F. For $R_{2}=2 R_{1}, C_{F B}=C_{I N} / 2 \approx 2.5 p F$.
Figure 4 shows the results of this feedback capacitor addition. Neutralizing the input capacitance demonstrates the peaking that can result from the loss of gain margin at 70 to 100 MHz . As the load time constant
( $\mathrm{R}_{\mathrm{L} C L}$ ) increases the peaking gets progressively worse $\approx 6 \mathrm{~dB}$ at $R_{L}=2 K, C_{L}=50 \mathrm{pF}$. The step response waveforms are as expected with a very strong 88 MHz ring being exhibited at $R_{L}=2 k, C_{L}=50 p F$ and no overshoot at $R L=50 \Omega, C L=5 p F$.

## Layout Considerations

As with any high-speed wideband amplifier, certain layout considerations are necessary to ensure stable operation. All connections to the amplifier should remain as short as possible, and the power supplies bypassed with $0.1 \mu \mathrm{~F}$ capacitors to signal ground. It is suggested that a ground plane be considered as the best method for ensuring stability because it minimizes stray inductance and unwanted coupling in the ground signal paths.
To minimize capacitive effects, resistor values should be kept as small as possible, consistent with the application.

## MAX408 Offset Voltage Nulling

The configuration of Figure 2 will give a typical Vos nulling range of $\pm 15 \mathrm{mV}$. If a smaller adjustment range is desired, resistor values R1 and R2 can be increased accordingly. For example, at $\mathrm{R} 1=3.6 \mathrm{k} \Omega$, the adjustment range is $\pm 5 \mathrm{mV}$. Since pins 1 and 5 are not part of the signal path, AC characteristics are left undisturbed.

Simplified Schematic. For MAX428/448 omit balance pins.


Figure 2. VOS Nulling Method for MAX408


## Single/Dual/Quad High-Speed, Fast-Settling, High Output Current Operational Amplifier



Figure 3. Frequency and Time Domain Response Characteristics, $A v=3$

Single/Dual/Quad High-Speed, Fast-Settling, High Output Current Operational Amplifier



[^0]
## Single/Dual/Quad High-Speed, Fast-Settling, High Output Current Operational Amplifier

## __Ordering Information (continued)

| PART | TEMP. RANGE | PIN-PACKAGE |
| :--- | :--- | :--- |
| MAX428A_CPA | $0^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}$ | 8 Lead Plastic DIP |
| MAX428ACSA | $0^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}$ | 8 Lead Small Outline |
| MAX428CPA | $0^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}$ | 8 Lead Plastic DIP |
| MAX428CSA | $0^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}$ | 8 Lead Small Outline |
| MAX428C/D | $0^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}$ | Dice |
| MAX448ACPD | $0^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}$ | 14 Lead Plastic DIP |
| MAX448ACSD | $0^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}$ | 14 Lead Small <br> Outline |
| MAX448CPD | $0^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}$ | 14 Lead Plastic DIP |
| MAX448CSD | $0^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}$ | 14 Lead Small <br> Outline |
| MAX448C/D | $0^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}$ | Dice |

## Single/Dual/Quad High-Speed, Fast-Settling, High Output Current Operational Amplifier

## NOTES

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[^0]:    Figure 4. Response Characteristics with Input Pole Cancellation, $A v=3$

