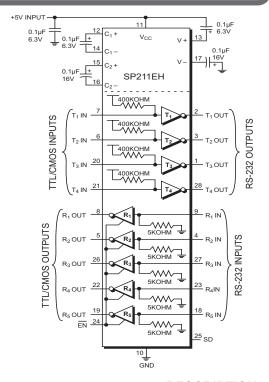


## SP211EH/213EH

# High Speed +5V High Performance RS-232 Transceivers

- Single +5V Supply Operation
- 0.1µF External Charge Pump Capacitors
- 500kbps Data Rate Under Load
- Standard SOIC and SSOP Footprints
- Lower Supply Current Than Competition (typical 3mA)
- 1µA Shutdown Mode
- Wake-Up Feature in Shutdown Mode
- Tri-State Receiver Outputs
- Ideal for High Speed RS-232 Applications
- Improved ESD Specifications:
   ±15kV Human Body Model
   ±15kV IEC 61000-4-2 Air Discharge
   ±8kV IEC 61000-4-2 Contact Discharge

Now Available in Lead Free Packaging



#### DESCRIPTION

The **SP211EH/213E**H devices are high speed enhanced multi-channel RS-232 line transceivers with improved electrical performance. The **SP211EH/213E**H series is a superior drop-in replacement to our previous versions as well as popular industry standards. All devices feature very low power CMOS construction and the Exar-patented (5,306,954) on-board charge pump circuitry that generates the +/-10V RS-232 voltage levels using 0.1µF charge pump capacitors. The **SP211EH** and **SP213EH** devices feature a low-power shutdown mode, which reduces power supply drain to 1µA. Enhancements to this series include a higher transmission rate of 500kbps, a lower power supply current at 3mA typical (no load), and superior ESD performance. The ESD tolerance has been improved for this series to over +/-15kV for both Human Body Model and IEC61000-4-2 Air Discharge test methods.

		f RS-232	No. of Receivers	No. of External			TTL
Model	Drivers	Receivers	Active in Shutdown	0.1µF Capacitors	Shutdown	Wake-Up	Tri-State
SP211EH	4	5	0	4	Yes	No	Yes
SP213EH	4	5	2	4	Yes	Yes	Yes

Table 1. Model Selection Table

## **ABSOLUTE MAXIMUM RATINGS**

These are stress ratings only and functional operation of the device at these or any other above those indicated in the operation sections of the specifications below is not implied. Exposure to absolute maximum rating conditions for extended periods of time may affect reliability.

V <sub>CC</sub>	+6V (V <sub>cc</sub> – 0.3V) to +13.2V
V	13.2V
Input Voltages	
T <sub>IN</sub>	0.3V to (V <sub>CC</sub> +0.3V) ±20V
R <sub>IN</sub>	±20V
Output Voltages	
Т <sub>оит</sub>	$(V^+, +0.3V)$ to $(V^-, -0.3V)$
R <sub>OUT</sub>	–0.3V to (V <sub>cc</sub> +0.3V)
Short Circuit Duration or	. (V $^+$ , +0.3V) to (V $^-$ , -0.3V)0.3V to (V $_{\rm CC}$ +0.3V) n T $_{\rm OUT}$ Continuous

### **Power Dissipation Per Package**

28-pin SSOP (derate 11.2mW/°C above +70°C)....900mW 28-pin SOIC (derate 12.7mW/°C above +70°C)...1000mW

## **SPECIFICATIONS**

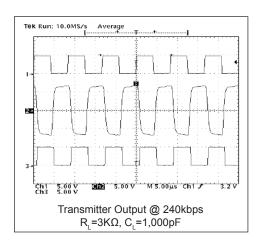
 $V_{_{CC}}$  at nominal ratings; 0.1µF charge pump capacitors;  $T_{_{MIN}}$  to  $T_{_{MAX}}$ , unless otherwise noted.

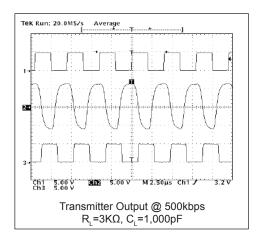
PARAMETER	MIN.	TYP.	MAX.	UNIT	CONDITIONS
TTL INPUTS	•	•			T <sub>IN</sub> , EN, SD
Logic Threshold V <sub>IL</sub>			0.8	Volts	
Logic Threshold V <sub>IH</sub>	2.0			Volts	
Logic Pull-Up Current		15	200	μA	T <sub>IN</sub> = 0V
Maximum Transmission Rate	480	ĺ		kbps	$C_{L} = 1000 pF, R_{L} = 3k\Omega$
TTL OUTPUTS	•	-			
Compatibility		TTL/CMO	s		
V <sub>OL</sub>			0.4	Volts	I <sub>OUT</sub> = 3.2mA: Vcc = +5V
V <sub>OH</sub>	3.5	Ì		Volts	I <sub>OUT</sub> = -1.0mA
Leakage Current		0.05	+/-10	μΑ	$0V \le V_{OUT} \le Vcc$ ; SP211EH E = 0V; SP213EH EN = Vcc, $T_A = +25$ °C
RS-232 OUTPUT					
Output Voltage Swing	+/-5	+/-7		Volts	All transmitter outputs loaded with 3kΩ to ground
Output Resistance	300			Ω	Vcc = 0V; V <sub>OUT</sub> = +/-2V
Output Short Circuit Current		+/-25		mA	Infinite Duration, V <sub>OUT</sub> = 0V
RS-232 INPUT	•				
Voltage Range	-15		+15	Volts	
Voltage Threshold Low	0.8	1.2		Volts	Vcc = 5V, T <sub>A</sub> = +25°C
Voltage Threshold High		1.7	2.8	Volts	Vcc = 5V, T <sub>A</sub> = +25°C
Hysteresis	0.2	0.5	1.0	Volts	Vcc = 5V
Resistance	3	5	7	kΩ	V <sub>IN</sub> = +/-15V, T <sub>A</sub> = +25°C
DYNAMIC CHARACTERISTICS	i .				
Driver Propagation Delay		250		ns	TTL to RS-232
Receiver Propagation Delay		200	500	ns	RS-232 to TTL
Instantaneous Slew Rate			TBD	V/µs	$C_L = 2500 \text{pF}, R_L = 3 \text{k}\Omega;$ $T_A = +25^{\circ}\text{C}; \text{ from +/-3V}$
TATTATTAT				nn	SP2 E 13 H /01 060

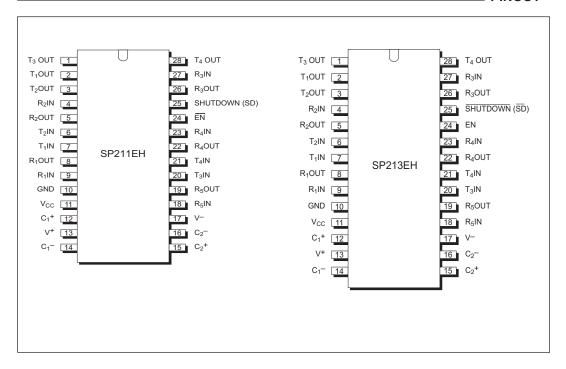
## **SPECIFICATIONS**

 $V_{_{CC}} \ at \ nominal \ ratings; \ 0.1 \mu F \ charge \ pump \ capacitors; \ T_{_{MIN}} \ to \ T_{_{MAX'}} \ unless \ otherwise \ noted.$ 

PARAMETER	MIN.	TYP.	MAX.	UNIT	CONDITIONS	
DYNAMIC CHARACTERISTICS continued						
Transition Time			TBD	μs	$C_L$ = 2500pF, $R_L$ = 3k $\Omega$ , Measured from -3V to +3V or +3V to -3V	
Output Enable Time		400		ns		
Output Disable Time		250		ns		
POWER REQUIREMENTS						
Vcc	4.50	5.00	5.50	Volts		
Icc		3	6	mA	No Load: Vcc = +/-10%	
Icc		15		mA	All Transmitters $R_L = 3k\Omega$	
Shutdown Current		1	10	μΑ	T <sub>A</sub> = +25°C	
ENVIRONMENTAL AND MECHA	NICAL					
Operating Temperature						
Commercial, _C	0		+70	°C		
Extended, _E	-40		+85	°C		
Storage Temperature	-65		+125	°C		
Package _A _T	Shrink (SSOP) small outline Wide (SOIC) small outline					







#### **FEATURES**

As in the original RS-232 multi-channel products, the **SP211EH** and **SP213EH** high speed multi-channel RS-232 line transceivers provide a variety of configurations to fit most designs, especially high speed applications where +/-12V is not available. The **SP211EH/213EH** is a superior high speed drop-in replacement to our previous versions as well as popular industry standards.

All devices in this series feature low–power CMOS construction and **Exar's**–proprietary on-board charge pump circuitry to generate the RS-232 signal voltages. The ability to use 0.1µF charge pump capacitors saves board space and reduces production costs. The devices in this series provide different driver/receiver combinations to match any application requirement.

The **SP211EH** and **SP213EH** models feature a low–power shutdown mode, which reduces power supply drain to 1µA. The **SP213EH** includes a Wake-Up function which keeps two receivers active in the shutdown mode, unless disabled by the EN pin.

The family is available in 28 SOIC (wide) and SSOP (shrink) small outline packages. Devices can be specified for commercial (0°C to +70°C) and industrial/extended (-40°C to +85°C) operating temperatures.

## THEORY OF OPERATION

The **SP211EH** and **SP213EH** devices are made up of three basic circuit blocks — 1) transmitter/driver, 2) receiver and 3) the **EXAR**–proprietary charge pump. Each model within the series incorporates variations of these circuits to achieve the desired configuration and performance.

## Transmitter/Drivers

The drivers of this series can maintain a typical data rate of 500kbps.

This superior RS-232 data transmission rate makes the **SP211EH/213EH** series an ideal match for many designs in personal computer peripherals and LAN applications that demand high speed performance.

The drivers are inverting transmitters, which accept either TTL or CMOS inputs and output the RS-232 signals with an inverted sense relative to the input logic levels. Typically, the RS-232 output voltage swing is +/-9V with no load, and +/-5V minimum with full load. The transmitter outputs are protected against infinite short—circuits to ground without degradation in reliability. The drivers of the **SP211EH**, and **SP213EH** can be tri—stated by using the SHUTDOWN function.

In the "power-off" state, the output impedance will remain greater than 300 ohms, again satisfying the RS-232 specifications. Should the input of the driver be left open, an internal 400k ohm pull-up resistor to Vcc forces the input high, thus committing the output to a low state. The slew rate of the transmitter output is internally limited to a maximum of 30V/µs in order to meet the EIA/RS-232 and ITU V.28 standards. The transition of the output from high to low also meets the monotonicity requirements of the standard.

#### Receivers

The high performance receivers of the SP211EH/213EH devices can accept input signals at a typical data rate or 500kbps. The receivers convert RS-232 level input signals to inverted TTL level signals. Since the input is usually from a transmission line where long cable lengths and system interference can degrade the signal, the inputs have a typical hysteresis margin of 500mV. This ensures that the receiver is virtually immune to noisy transmission lines. Should an input be left unconnected, a  $5k\Omega$  pull-down resistor to ground will commit the output of the receiver to a high state.

#### SHUTDOWN MODE

The **SP211EH** and **SP213EH** both feature a control input which will disable the device and reduce power supply to less than 10µA, making the parts ideal for battery-powered systems. In the "shutdown" mode the receivers and transmitters will both be tristated. The V+ output of the charge pump will discharge to Vcc and the V- output will discharge to ground.

For complete shutdown to occur and the 10µA power drain to be realized, the following conditions must be met:

#### **SP211EH:**

- +5V must be applied to the SD pin.
- ENABLE must be either 0V, +5V or not connected
- the transmitter inputs must be either
  +5.0V or not connected
- Vcc must be +5V
- Receiver inputs must be >0V and <+5V</li>

#### SP213EH:

- 0V must be applied top the SD pin
- ENABLE must be either 0V, +5.0V or not connected
- the transmitter inputs must be either +5.0V or not connected
- Vcc must be +5V
- Receiver inputs must be >0V and <+5V</li>

#### **ENABLE**

SP211EH and SP213EH feature an enable input, which allows the receiver outputs to be either tri–stated or enabled. This can be especially useful when the receiver is tied directly to a shared microprocessor data bus. For the SP211EH, enable is active low; that is, 0V applied to the ENABLE pin will enable the receiver outputs. For the SP213EH, enable is active high; that is, +5V applied to the ENABLE pin will enable the receiver outputs.

### **WAKE-UP FUNCTION**

The **SP213EH** has a wake-up feature that keeps two receivers (R4 and R5) in an enabled state when the device is in shutdown mode. With only the receivers active during shutdown, the devices draw  $5-10\mu\text{A}$  of supply current.

Atypical application of this function would be where a modem is interfaced to a computer in a power-down mode. The ring indicator signal from the modem could be passed through an active receiver in the SP213EH that is itself in the shutdown mode. The ring indicator signal would propagate through the SP213EH to the power management circuitry of the computer to power up the microprocessor and the SP213EH drivers. After the supply voltage to the SP213EH reaches +5.0V, the SHUTDOWN pin can be disabled, taking the SP213EH out of the shutdown mode.

All receivers that are active during shutdown maintain 500mV (typical) of hysteresis.

SD	EN	Power Up/Down	Receiver Outputs
1	1	Up	Enable
1	0	Up	Tri-state
0	1	Down	Enable
0	0	Down	Tri-State

Table 2. SP213EH Wake-Up Truth Table

SD	EN	Power Up/Down	Receiver Outputs
0	0	Up	Enable
0	1	Up	Tri-state
1	0	Down	Enable
1	1	Down	Tri-State

Table 3. SP211EH Receiver Truth Table

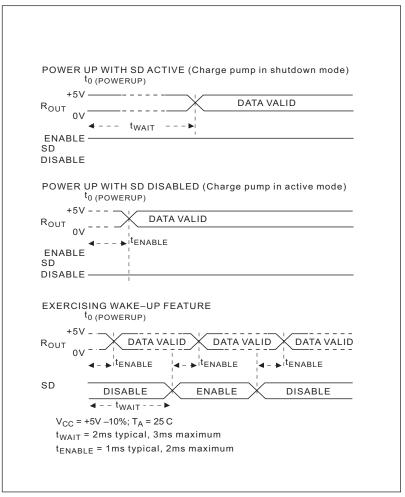


Figure 6. Wake-Up Timing

## Charge-Pump

The charge pump is a **Exar**-patented design (5,306,954) and uses a unique approach compared to older less-efficient designs. The charge pumps still requires four external capacitors, but uses a four-phase voltage shifting technique to attain symmetrical 10V power supplies. Figure 3a shows the waveform found on the positive side of capacitor C2, and Figure 3b shows the negative side of capacitor C2. There is a free-running oscillator that controls the four phases of the voltage shifting. A description of each phase is as follows:

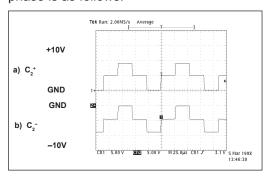


Figure 5. Typical waveforms seen on capacitor C2 when all drivers are at maximum load.

#### Phase 1

 $-V_{ss}$  charge storage —During this phase of the clock cycle, the positive side of capacitors  $C_1$  and  $C_2$  are initially charged to +5V.  $C_1^+$  is then switched to ground and charge on  $C_1^-$  is transferred to  $C_2^-$ . Since  $C_2^+$  is connected to +5V, the voltage potential across capacitor  $C_2^-$  is now 10V.

#### Phase 2

 $-\rm V_{SS}$  transfer — Phase two of the clock connects the negative terminal of  $\rm C_2$  to the V  $_{SS}$  storage capacitor and the positive terminal of C  $_2$  to ground, and transfers the generated –I0V to C  $_3$ . Simultaneously, the positive side of capacitor C  $_1$  is switched to +5V and the negative side is connected to ground.

## Phase 3

V<sub>DD</sub> charge storage — The third phase of the clock is identical to the first phase
 the charge transferred in C produces

oration 8720 Late Road, Femoral CA,

-5V in the negative terminal of  $C_1$ , which is applied to the negative side of capacitor  $C_2$ . Since  $C_2^+$  is at +5V, the voltage potential across  $C_2$  is l0V.

#### Phase 4

-  $\rm V_{DD}$  transfer - The fourth phase of the clock connects the negative terminal of  $\rm C_2$  to ground and transfers the generated IOV across  $\rm C_2$  to  $\rm C_4$ , the  $\rm V_{DD}$  storage capacitor. Again, simultaneously with this, the positive side of capacitor  $\rm C_1$  is switched to +5V and the negative side is connected to ground, and the cycle begins again.

Since both V+ and V<sup>-</sup> are separately generated from  $V_{\rm cc}$  in a no–load condition, V+ and V<sup>-</sup> will be symmetrical. Older charge pump approaches that generate V<sup>-</sup> from V+ will show a decrease in the magnitude of V<sup>-</sup> compared to V+ due to the inherent inefficiencies in the design.

The clock rate for the charge pump typically operates at 15kHz. The external capacitors must be a minimum of 0.1µF with a 16V breakdown rating.

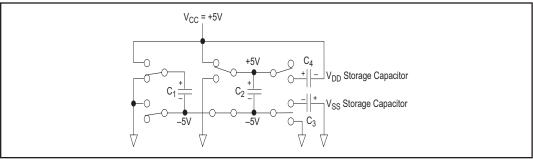


Figure 6, Charge Pump - Phase 1

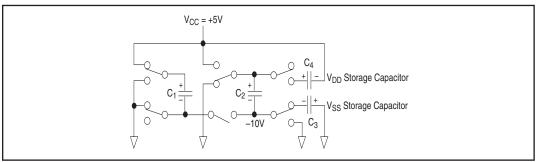


Figure 7, Charge Pump - Phase 2

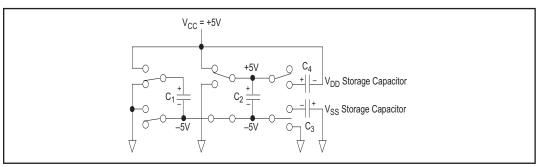


Figure 8, Charge Pump - Phase 3

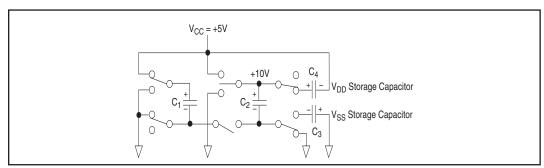


Figure 9, Charge Pump - Phase 4

#### **ESD TOLERANCE**

The **SP211EH/213EH** family incorporates ruggedized ESD cells on all driver output and receiver input pins. The ESD structure is improved over our previous family for more rugged applications and environments sensitive to electro-static discharges and associated transients. The improved ESD tolerance is at least ±15kV without damage nor latch-up.

There are different methods of ESD testing applied:

- a) MIL-STD-883, Method 3015.7
- b) IEC61000-4-2 Air-Discharge
- c) IEC61000-4-2 Direct Contact

The Human Body Model has been the generally accepted ESD testing method for semiconductors. This method is also specified in MIL-STD-883, Method 3015.7 for ESD testing. The premise of this ESD test is to simulate the human body's potential to store electro-static energy and discharge it to an integrated circuit. The simulation is performed by using a test model as shown in Figure 10. This method will test the IC's capability to withstand an ESD transient during normal handling such as in manufacturing areas where the IC's tend to be handled frequently.

The IEC-61000-4-2, formerly IEC801-2, is generally used for testing ESD on equipment

and systems. For system manufacturers, they must guarantee a certain amount of ESD protection since the system itself is exposed to the outside environment and human presence. The premise with IEC61000-4-2 is that the system is required to withstand an amount of static electricity when ESD is applied to points and surfaces of the equipment that are accessible to personnel during normal usage. The transceiver IC receives most of the ESD current when the ESD source is applied to the connector pins. The test circuit for IEC61000-4-2 is shown on Figure 11. There are two methods within IEC61000-4-2, the Air Discharge method and the Contact Discharge method.

With the Air Discharge Method, an ESD voltage is applied to the equipment under test (EUT) through air. This simulates an electrically charged person ready to connect a cable onto the rear of the system only to find an unpleasant zap just before the person touches the back panel. The high energy potential on the person discharges through an arcing path to the rear panel of the system before he or she even touches the system. This energy, whether discharged directly or through air, is predominantly a function of the discharge current rather than the discharge voltage. Variables with an air discharge such as approach speed of the object carrying the ESD potential to the system and humidity will tend to change the discharge current. For example, the rise time of the discharge current varies with the approach speed.

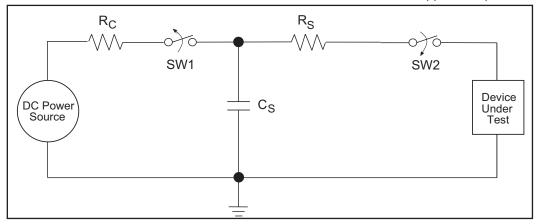


Figure 10. ESD Test Circuit for Human Body Model

Exam Lonyoration 92.00 to Road, Example SA, 1524-010 888-007, white large on SF2 15 13 H 101\_060111

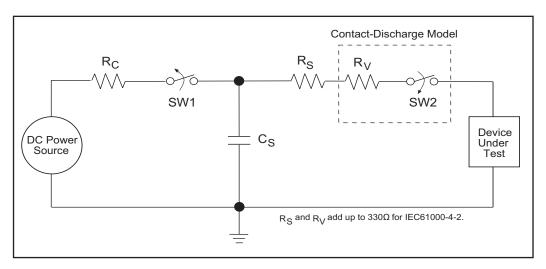


Figure 11. ESD Test Circuit for IEC61000-4-2

The Contact Discharge Method applies the ESD current directly to the EUT. This method was devised to reduce the unpredictability of the ESD arc. The discharge current rise time is constant since the energy is directly transferred without the air-gap arc. In situations such as hand held systems, the ESD charge can be directly discharged to the equipment from a person already holding the equipment. The current is transferred on to the keypad or the serial port of the equipment directly and then travels through the PCB and finally to the IC.

The circuit model in Figures 10 and 11 represent the typical ESD testing circuit used for all three methods. The  $C_{\rm S}$  is initially charged with the DC power supply when the first switch (SW1) is on. Now that the capacitor is charged, the second switch (SW2) is on while SW1 switches off. The voltage stored in the capacitor is then applied through  $R_{\rm S}$ , the current limiting resistor, onto the device under test (DUT). In ESD tests, the SW2 switch is pulsed so that the device under test receives a duration of voltage.

For the Human Body Model, the current limiting resistor ( $R_{\rm S}$ ) and the source capacitor

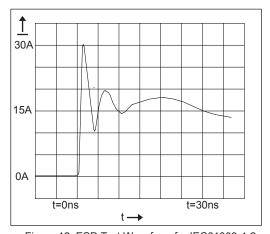


Figure 12. ESD Test Waveform for IEC61000-4-2

 $(\text{C}_{\text{S}})$  are 1.5k $\Omega$  an 100pF, respectively. For IEC-61000-4-2, the current limiting resistor  $(\text{R}_{\text{S}})$  and the source capacitor  $(\text{C}_{\text{S}})$  are 330 $\Omega$  an 150pF, respectively.

The higher  $C_s$  value and lower  $R_s$  value in the IEC61000-4-2 model are more stringent than the Human Body Model. The larger storage capacitor injects a higher voltage to the test point when SW2 is switched on. The lower current limiting resistor increases the current charge onto the test point.

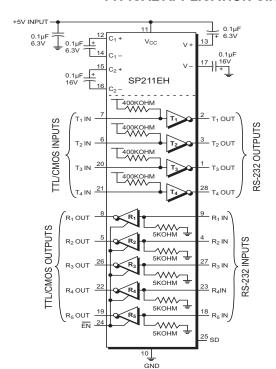
#### **EIA STANDARDS**

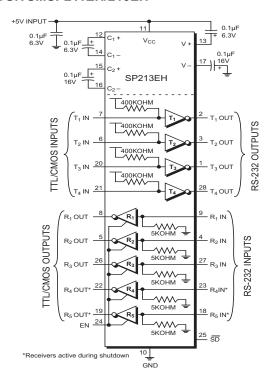
The Electronic Industry Association (EIA) developed several standards of data transmission which are revised and updated in order to meet the requirements of the industry. In data processing, there are two basic means of communicating between systems and components. The RS--232 standard was first introduced in 1962 and, since that time, has become an industry standard. The RS-232 is a relatively slow data exchange protocol, with a maximum baud rate of only 20kbps, which can be transmitted over a maximum copper wire cable length of 50 feet. The SP211EH/213EH series of data communications interface products have been designed to meet both the EIA protocol standards, and the needs of the industry.

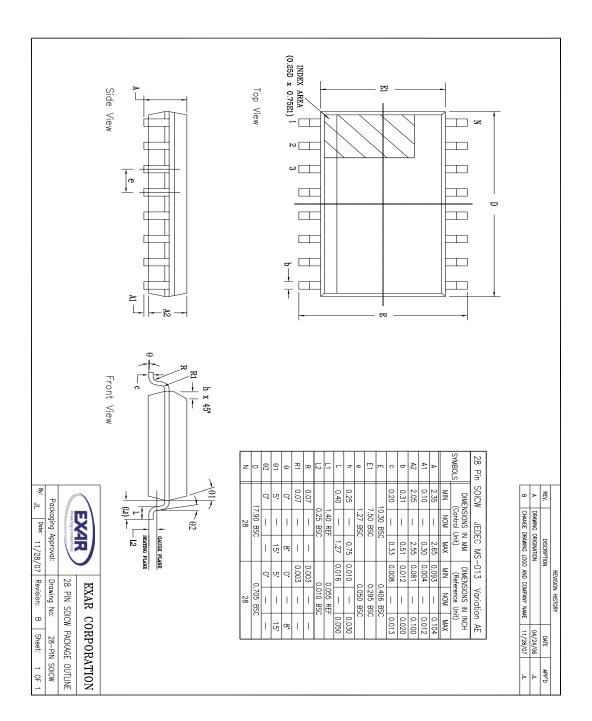
DEVICE PIN	HUMAN BODY	IEC16000-4-2				
TESTED	MODEL	Air Discharge	Direct Contact	Level		
Driver Outputs Receiver Inputs	±15kV ±15kV	<u>+</u> 15kV <u>+</u> 15kV	<u>+</u> 8kV <u>+</u> 8kV	4 4		

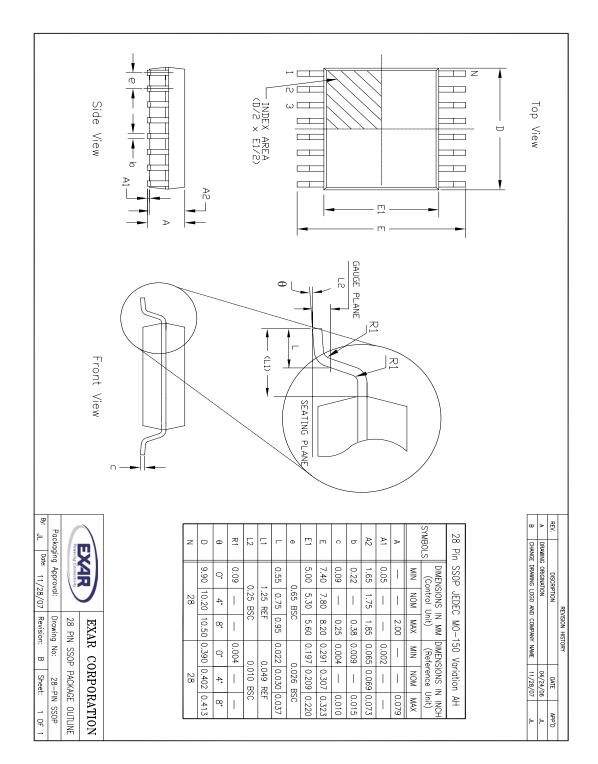
Table 3. Transceiver ESD Tolerance Levels

## TYPICAL APPLICATION CIRCUITS...SP211EH/213EH









## ORDERING INFORMATION RS232 Transceivers with Low-Power Shutdown and Tri-state Enable: RS232 Transceivers with Low-Power Shutdown, Tri-state Enable, and Wake-Up Function:

DATE	REVISION	DESCRIPTION
08		Legacy Sipex Datasheet
10/28/10	1.0.0	Convert to Exar format, update ordering information and change rev to 1.0.0
June 2011	1.0.1	All options of SP208EH are discontinued per PDN 110510-01

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