

# 2.7V to 5.5V, 12-Bit, Octal DIGITAL-TO-ANALOG CONVERTER in a Wafer Chip-Scale Package—Pb-Free/Green

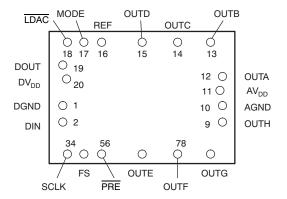
### **FEATURES**

- Eight, 12-Bit, Voltage Output DACs in One Package
- Pin-Compatible, Pb-Free, RoHS-Compliant Upgrade to the TLV5610IYE
- Programmable Settling Time versus Power Consumption:
  - 1µs in Fast Mode
  - 3µs in Slow Mode
- Compatible With TMS320<sup>™</sup> DSP Family and SPI<sup>™</sup> Serial Ports
- Differential and Single-Ended Analog Input/Output
- Separate Software Control for ADC and DAC Power Down
- Monotonic Over Temperature
- Low Power Consumption:
  - 18mW in Slow Mode at 3V
  - 48mW in Fast Mode at 3V
- Power Down Mode
- Buffered, High-Impedance Reference Inputs
- Data Output for Daisy-Chaining

### **APPLICATIONS**

- Digital Servo-Control Loops
- Digital Offset and Gain Adjustment
- Industrial Process Control
- Machine and Motion Control Devices
- Mass Storage Devices

# WAFER CHIP-SCALE PACKAGE (BOTTOM VIEW)



### **DESCRIPTION**

The TLV5610IYZ is the lead-free version of the TLV5610IYE, an eight-channel, 12-bit, voltage output digital-to-analog converter (DAC) with a flexible serial interface. The serial interface allows glueless interface to TMS320 and SPI, QSPI™, and Microwire™ serial ports. It is programmed with a 16-bit serial string containing four control and 12 data bits.

Additional features include a power-down mode, an  $\overline{\text{LDAC}}$  input for simultaneous update of all eight DAC outputs, and a data output that can be used to cascade multiple devices.

The resistor string output voltage is buffered by a rail-to-rail output amplifier with a programmable settling time to allow the designer to optimize speed versus power dissipation. The buffered, high-impedance reference input can be connected to the supply voltage.

The TLV5610IYZ is implemented in a CMOS process and is available in a 20-terminal wafer chip-scale package (WCSP). The TLV5610IYZ is characterized for operation from -40°C to +85°C in a wire-bonded small outline (SO) package.

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### SBAS389A-JULY 2006-REVISED JULY 2006



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

### ORDERING INFORMATION

T <sub>A</sub>	WCSP-20 PACKAGE <sup>(1)</sup>
-40°C to +85°C	TLV5610IYZ

<sup>(1)</sup> For the most current package and ordering information, see the Package Option Addendum at the end of this document, or see the TI web site at www.ti.com.

### **ABSOLUTE MAXIMUM RATINGS**

Over operating free-air temperature range (unless otherwise noted)<sup>(1)</sup>

	TLV5610IYZ	UNIT
Supply voltage (AV <sub>DD</sub> , DV <sub>DD</sub> to GND)	7	V
Reference input voltage range	- 0.3 to AV <sub>DD</sub> + 0.3	V
Digital input voltage range	- 0.3 to DV <sub>DD</sub> + 0.3	V
Operating free-air temperature range, T <sub>A</sub>	-40 to +85	°C

<sup>(1)</sup> Stresses beyond those listed under absolute maximum ratings may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these or any other conditions beyond those indicated under recommended operating conditions is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

### RECOMMENDED OPERATING CONDITIONS

		MIN	NOM	MAX	UNIT
Complementary AV DV	5V operation	4.5	5	5.5	V
Supply voltage, AV <sub>DD</sub> , DV <sub>DD</sub>	3V operation	2.7	3	3.3	V
High-level digital input, V <sub>IH</sub>	DV <sub>DD</sub> = 2.7V to 5.5V	2			V
Low-level digital input, V <sub>IL</sub>	DV <sub>DD</sub> = 2.7V to 5.5V			08	V
Deference voltage V	$AV_{DD} = 5V$	GND	4.096	$AV_{DD}$	V
Reference voltage, V <sub>REF</sub>	$AV_{DD} = 3V$	GND	2.048	$AV_{DD}$	V
Load resistance. R <sub>L</sub>		2			kΩ
Load capacitance, C <sub>L</sub>				100	pF
Clock frequency, f <sub>CLK</sub>				30	MHz
Operating free-air temperature, T	A	-40		+85	°C



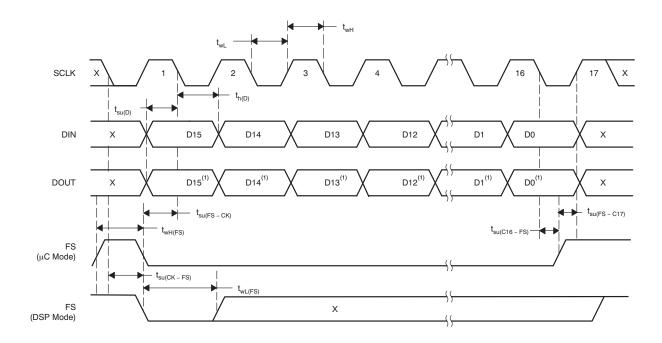
### **ELECTRICAL CHARACTERISTICS**

Over operating free-air temperature range (unless otherwise noted)

	PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
Power S	Supply						
	Davier august august	Fast	No load, V <sub>REF</sub> = 4.096V, all inputs = DV <sub>DD</sub> or		16	21	mA
I <sub>DD</sub>	Power-supply current	Slow	GND		6	8	mA
	Power-down supply current				0.1		μΑ
POR	Power-on threshold				2		V
PSRR	Power-supply rejection ratio		Full-scale; see Note (1)		-60		dB
Static D	AC Specifications						
	Resolution				12		Bits
INL	Integral nonlinearity		Code 40 to 4095, V <sub>REF</sub> = 2V, 4V		±2	±6	LSB
DNL	Differential nonlinearity		Code 40 to 4095, V <sub>REF</sub> = 2V, 4V		±0.5	±1	LSB
E <sub>ZS</sub>	Zero-scale error (offset error at zero-s	scale)				±30	mV
E <sub>ZS</sub> TC	Zero-scale error temperature coefficie	ent			30		μV/°C
$E_G$	Gain error					±0.6	% Full-Scale V
E <sub>G</sub> TC	Gain error temperature coefficient				10		ppm/°C
Output	Specifications						
Vo	Voltage output range		$R_L = 10k\Omega$	0		$AV_{DD} - 4$	V
	Output load regulation accuracy		$R_L = 2k\Omega \text{ vs } 10k\Omega$			±0.3	% Full-Scale V
Referen	ce Input						
VI	Input voltage range			0		$AV_{DD}$	V
R <sub>I</sub>	Input resistance				100		kΩ
Cı	Input capacitance Fast				5		pF
			V 0.4V + 2.049V input code 0v900		2.2		MHz
	Reference input bandwidth	Slow	$V_{REF} = 0.4V_{PP} + 2.048V_{DC}$ , input code = 0x800		1.9		MHz
	Reference feedthrough		$V_{REF} = 2V_{PP}$ at 1kHz + 2.048 $V_{DC}^{(2)}$		-84		dB
Digital I	nputs						
I <sub>IH</sub>	High-level digital input current		$V_{I} = DV_{DD}$			1	μΑ
I <sub>IL</sub>	Low-level digital input current		V <sub>I</sub> = 0V	-1			μΑ
Cı	Input capacitance				8		pF
Digital (	Outputs						
V <sub>OH</sub>	High-level digital output voltage		$R_L = 10k\Omega$	2.6			V
V <sub>OL</sub>	Low-level digital output voltage		$R_L = 10k\Omega$			0.4	V
	Output voltage rise time		$R_L = 10k\Omega$ , $C_L = 20pF$ , incl. propagation delay		7	20	ns
Analog	Output Dynamic Performance					<u> </u>	
+ (EC)	Output pottling time (3) full post-	Fast	B = 10k0 C = 1000nE		1	3	μs
t <sub>S</sub> (FS)	Output settling time (3), full-scale	Slow	$R_L = 10k\Omega, C_L = 1000pF$		3	7	μs
+ (CC)	Output pottling time (4)	Fast	B 40k0 C 4000pF		0.5	1	μs
t <sub>S</sub> (CC)	Output settling time (4), code to code	Slow	$R_L = 10k\Omega, C_L = 1000pF$		1	2	μs
CD	Claurata (5)	Fast	B 40k0 C 4000=F	4	10		V/μs
SR	Slew rate <sup>(5)</sup>	Slow	$R_L = 10k\Omega, C_L = 1000pF$	1	3		V/μs
	Glitch energy		See Note <sup>(6)</sup>		4		nV-s
	Channel crosstalk		10kHz sine, 4V <sub>PP</sub>		-90		dB

- (1) Power-supply rejection ratio at full-scale is measured by varying AV<sub>DD</sub> and is given by the following equation:
- PSRR = 20  $\log[(\text{EG(AV_{DD}max}) \text{EG(AV_{DD}min}))]/V_{DD}max$ . Reference feedthrough is measured at the DAC output with an input code = 0x000. Settling time is the time for the output signal to remain within  $\pm 0.5$  LSB of the final measured value for a digital input code change of (3)0x80 to 0xFFF and 0xFFF to 0x080, respectively. Assured by design; not production tested.
- Settling time is the time for the output signal to remain within ±0.5 LSB of the final measured value for a digital input code change of one count. The max time applies to code changes near zero scale or full-scale. Assured by design; not production tested.
- Slew rate determines the time it takes for a change of the DAC output from 10% to 90% full-scale voltage.
- Code transition: 0x7FF to 0x800.





NOTE: (1) Previous input data.

Figure 1. Serial Interface Timing

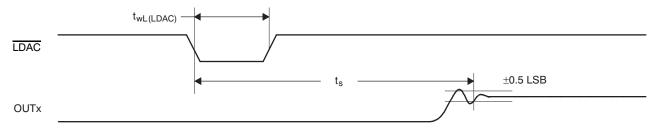


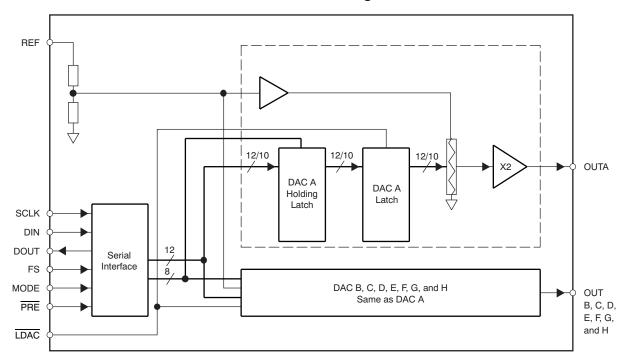
Figure 2. Output Timing

### **DIGITAL INPUT TIMING REQUIREMENTS**

	PARAMETER	MIN	TYP	MAX	UNIT
t <sub>su(FS-CK)</sub>	Setup time, FS low before next negative SCLK edge	8			ns
t <sub>su(C16-FS)</sub>	Setup time, 16th negative edge after FS low on which bit D0 is sampled before rising edge of FS. $\mu C$ mode only	10			ns
t <sub>su(FS-C17)</sub>	μC mode, setup time, FS high before 17th positive SCLK.	10			ns
t <sub>su(CK-FS)</sub>	DSP mode, setup time, SLCK low before FS low.	5			ns
t <sub>wL(LDAC)</sub>	LDAC duration low	10			ns
t <sub>wH</sub>	SCLK pulse duration high	16			ns
t <sub>wL</sub>	SCLK pulse duration low	16			ns
t <sub>su(D)</sub>	Setup time, data ready before SCLK falling edge	8			ns
t <sub>h(D)</sub>	Hold time, data held valid after SCLK falling edge	5			ns
t <sub>wH(FS)</sub>	FS duration high	10			ns
t <sub>wL(FS)</sub>	FS duration low	10			ns
t <sub>s</sub>	Settling time	See Analog O Performance	utput Dyr	namic	



### **Functional Block Diagram**



### **Terminal Functions**

TERM	IINAL							
NAME	NO.	1/0	DESCRIPTION					
AGND	10	Р	Analog ground					
$AV_{DD}$	11	Р	Analog power supply					
DGND	1	Р	Digital ground					
DIN	2	I	Digital serial data input					
DOUT	19	0	Digital serial data output					
$DV_DD$	20	Р	Digital power supply					
FS	4	I	Frame sync input					
LDAC	18	I	Load DAC. The DAC outputs are only updated if this signal is low. It is an asynchronous input.					
MODE	17	I	DSP/ $\mu$ C mode pin. High = $\mu$ C mode; NC = DSP mode.					
PRE	5	I	Preset input					
REF	16	I	Voltage reference input					
SCLK	3	I	Serial clock input					
OUTA-OUTH	12–15, 6–9	0	DAC outputs A, B, C, D, E, F, G and H					



### TYPICAL CHARACTERISTICS

1.0

0.2

0.1

0

0



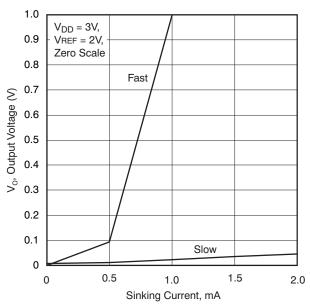


Figure 3.

# 0.9 | V<sub>DD</sub> = 5V, V<sub>REF</sub> = 4V Zero Scale | 0.7 | 0.6 | 0.5 | 0.5 | 0.4 | 0.5 | 0.5 | 0.7 | 0.4 | 0.5 | 0.5 | 0.3 | 0.4 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 | 0.5 |

Slow

0.5

**OUTPUT LOAD REGULATION** 

Sinking Current (mA)

Figure 4.

1.0

1.5

2.0

### **OUTPUT LOAD REGULATION**

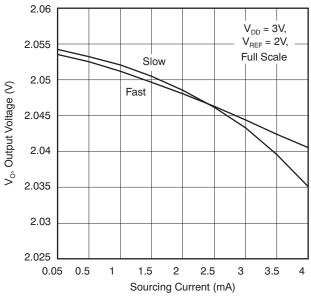


Figure 5.

### **OUTPUT LOAD REGULATION**

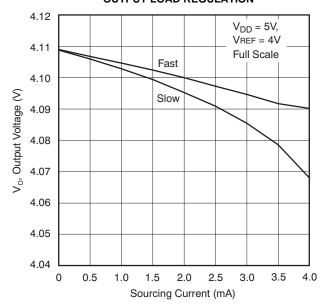
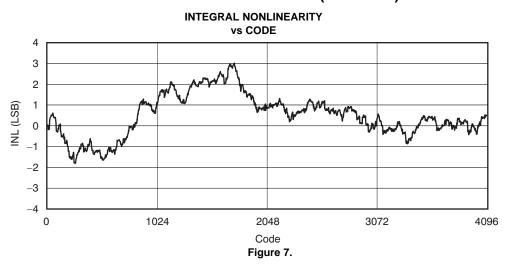
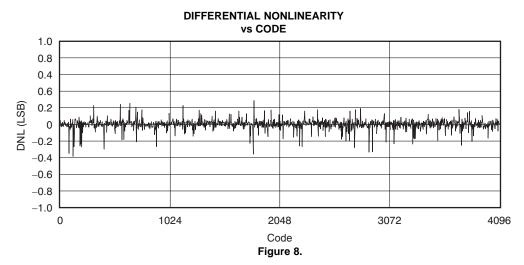


Figure 6.



### **TYPICAL CHARACTERISTICS (continued)**







### **APPLICATION INFORMATION**

### **GENERAL FUNCTION**

The TLV5610IYZ is an eight-channel, 12-bit, single-supply DAC, based on a resistor string architecture. The TLV5610IYZ, a green/PB-free device, is pin-compatible with the TLV5610IYE. The TLV5610IYZ consists of a serial interface, a speed and power-down control logic, a reference input buffer, a resistor string, and a rail-to-rail output buffer.

The output voltage (full-scale determined by external reference) for each channel is given by:

$$REF \frac{CODE}{0x1000}[V]$$
 (1)

Where:

REF is the reference voltage.

CODE is the digital input value.

The input range is 0x000 to 0xFFF.

A power on reset initially puts the internal latches to a defined state (all bits zero).

### **SERIAL INTERFACE**

A falling edge of FS starts shifting the data on DIN, starting with the MSB to the internal register on the falling edges of SCLK. After 16 bits have been transferred, the content of the shift register is moved to one of the DAC holding registers, depending on the address bits within the data word. A logic '0' on the LDAC pin is required to transfer the content of the DAC holding register to the DAC latch and to update the DAC outputs. LDAC is an asynchronous input; it can be held low if a simultaneous update of all eight channels is not needed.

For daisy-chaining, DOUT provides the data sampled on DIN with a delay of 16 clock cycles.

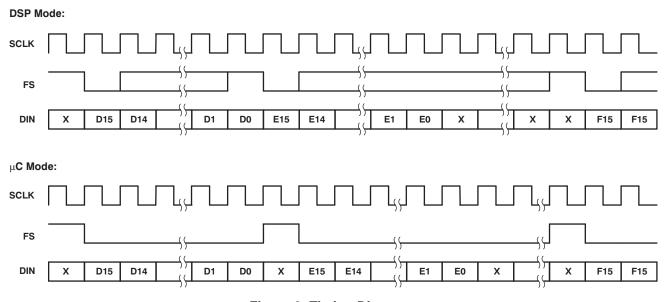


Figure 9. Timing Diagrams

The differences between DSP mode (MODE = NC or 0) and μC (MODE = 1) mode:

- In μC mode, FS must be held low until all 16 data bits have been transferred. If FS is driven high before the 16th falling clock edge, the data transfer is cancelled. The DAC is updated after a rising edge on FS.
- In DSP mode, FS must only stay low for 20ns and can go high before the 16th falling clock edge.



### **APPLICATION INFORMATION (continued)**

### SERIAL CLOCK FREQUENCY AND UPDATE RATE

The maximum serial clock frequency is given by:

$$f_{\text{sclkmax}} = \frac{1}{t_{\text{whmin}} + t_{\text{wlmin}}} = 30\text{MHz}$$
(2)

The maximum update rate is given by:

$$f_{updatemax} = \frac{1}{16(t_{whmin} + t_{wlmin})} = 1.95MHz$$
(3)

Note that the maximum update rate is just a theoretical value for the serial interface because the settling time of the DAC must also be considered.

### **DATA FORMAT**

The 16 bit data word consists of two parts:

- 1. Address bits (D15...D12)
- 2. Data bits (D11...D0)

### Table 1. Data Format<sup>(1)</sup>

D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
А3	A2	A1	A0						Da	ata					

(1) Ax: Address bits. See Table 2.

### **REGISTER MAP**

Table 2. Register Map

			-	•
А3	A2	A1	A0	FUNCTION
0	0	0	0	DAC A
0	0	0	1	DAC B
0	0	1	0	DAC C
0	0	1	1	DAC D
0	1	0	0	DAC E
0	1	0	1	DAC F
0	1	1	0	DAC G
0	1	1	1	DAC H
1	0	0	0	CTRL0
1	0	0	1	CTRL1
1	0	1	0	Preset
1	0	1	1	Reserved
1	1	0	0	DAC A and B
1	1	0	1	DAC C and $\overline{D}$
1	1	1	0	DAC E and F
1	1	1	1	DAC G and H



### DAC A-H AND TWO-CHANNEL REGISTERS

Writing to DAC A–H sets the output voltage of channel A–H. It is possible to automatically generate the complement of one channel by writing to one of the four, two-channel registers (DAC A and  $\overline{B}$ , etc.).

The TLV5610IYZ decodes all 12 data bits.

### **PRESET**

The outputs of all DAC channels can be driven to a predefined value stored in the preset register by driving the PRE input low. The PRE input is asynchronous to the clock.

### CTRL0

### Table 3. CTRL0 Bit Register

D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
X	Х	Х	Х	Х	Х	Х	PD	DO	X	X	IM

### **Bit Definitions:**

PD Full Device Power Down.

0 = Normal

1 = Power Down

**DO** Digital Output Enable.

0 = Disable

1 = Enable

IM Input Mode.

0 = Straight Binary

1 = Twos Complement

X Reserved.

If DOUT is enabled, the data input on DIN is output on DOUT with a 16-cycle delay. This feature makes it possible to daisy-chain multiple DACs on one serial bus.

### CTRL1

### Table 4. CTRL1 Bit Register

D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
Χ	Х	Χ	Х	Χ	Х	Х	PD	DO	Х	Х	IM

### **Bit Definitions:**

Power Down DAC $_{XY}$ . In power-down mode, the amplifiers of the selected DAC pair within the device are disabled and the total power consumption of the device is significantly reduced. Power-down mode of a specific DAC pair can be selected by setting the  $P_{XY}$  bit within the data word to '1'.

0 = Normal

1 = Power Down

 $S_{XY}$  DAC<sub>XY</sub> Speed Mode. There are two settling time modes: fast and slow. Fast mode of a DAC pair is selected by setting  $S_{XY}$  to '1', and slow mode is selected by setting  $S_{XY}$  to '0'.

0 = Slow

1 = Fast

NOTE: XY refers to DAC pair AB, CD, EF, or GH.



### **USING THE TLV5610IYZ WAFER CHIP-SCALE PACKAGE (WCSP)**

TLV5610 qualifications are done using a wire-bonded small outline (SO) package. The qualifications include: steady state life, thermal shock, ESD, latch-up, biased HAST, autoclave, and characterization. These qualified devices are orderable as TLV5610IDW.

**NOTE:** The wafer chip-scale package (WCSP) for the TLV5610IYZ uses the same *die* as TLV5610IDW, but is not qualified. WCSP qualification, including board level reliability (BLR), is the responsibility of the customer.

It is recommended that underfill be used for increased reliability. BLR is application-dependent, but may include tests such as: temperature cycling, drop test, key push, bend, vibration, and package shear.

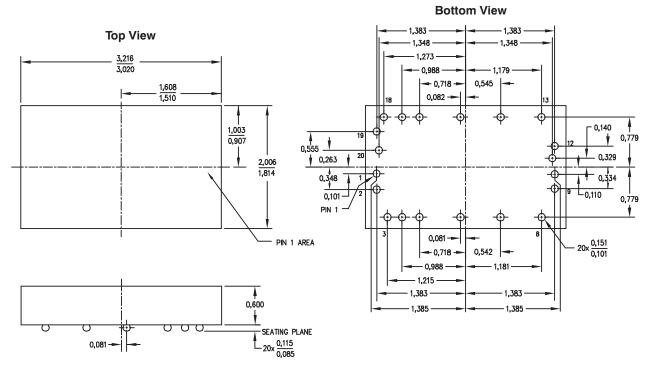
For general guidelines on board assembly of the WCSP, the following documentation provides more details:

- Application Report NanoStar™ & NanoFree™ 300µm Solder Bump WCSP Application—SBVA017
- Design Summary WCSP Little Logic—SCET007B

NOTE: The use of underfill is required and greatly reduces the risk of thermal mismatch fails.

Underfill is an epoxy/adhesive that may be added during the board assembly process to improve board level/system level reliability. The process is to dispense the epoxy under the dice after die attach reflow. The epoxy adheres to the body of the device and to the printed-circuit board. It reduces stress placed upon the solder joints because of thermal coefficient of expansion (TCE) mismatch between the board and the component. Underfill material is highly filled with silica or other fillers to increase epoxy modulus, reduce creep sensitivity, and decrease material TCE.

**NOTE:** The recommendation for peak flow temperatures of +250°C to +260°C is based on general empirical results that indicate that this temperature range is needed to facilitate good wetting of the solder bump to the substrate or circuit board pad. Lower peak temperatures may cause nonwets (cold solder joints).



- A. All linear dimensions are in millimeters.
- B. This drawing is subject to change without notice.

Figure 10. TLV5610IYZ Wafer Chip-Scale Package



## **Revision History**

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

CI	hanges from Original (July 2006) to A Revision	Page	;
•	Deleted separate INL and DNL specifications for Code 20 to 1023; not relevant to this device	3	3
•	Replaced timing diagram	4	ļ
•	Updated Digital Input Timing Requirements Table	4	



### PACKAGE OPTION ADDENDUM

www.ti.com 16-Apr-2009

### PACKAGING INFORMATION

Orderable Device	Status <sup>(1)</sup>	Package Type	Package Drawing	Pins I	Package Qty	e Eco Plan <sup>(2)</sup>	Lead/Ball Finish	MSL Peak Temp <sup>(3)</sup>
TLV5610IYZR	ACTIVE	DIESALE	YZ	20	3000	Green (RoHS & no Sb/Br)	SNAGCU	Level-NC-NC-UNLIM
TLV5610IYZT	ACTIVE	DIESALE	YZ	20	250	Green (RoHS & no Sb/Br)	SNAGCU	Level-NC-NC-UNLIM

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

**Pb-Free** (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

**Pb-Free (RoHS Exempt):** This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

**Green (RoHS & no Sb/Br):** TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

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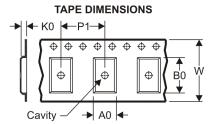




4-Feb-2009

### TAPE AND REEL INFORMATION





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	A0	Dimension designed to accommodate the component width
	B0	Dimension designed to accommodate the component length
		Dimension designed to accommodate the component thickness
	W	Overall width of the carrier tape
Γ	P1	Pitch between successive cavity centers

### QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE

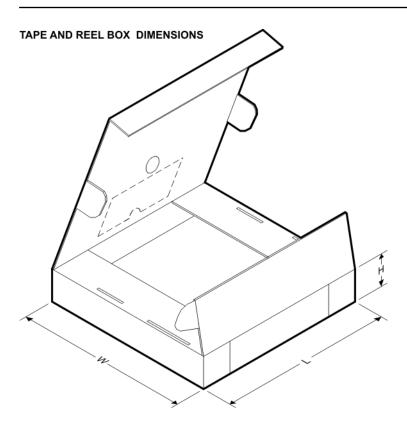


\*All dimensions are nominal

Device	Package Type	Package Drawing			Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TLV5610IYZR	DIESALE	YZ	20	3000	180.0	8.4	2.1	3.4	0.85	4.0	8.0	Q2
TLV5610IYZT	DIESALE	YZ	20	250	180.0	8.4	2.1	3.4	0.85	4.0	8.0	Q2

# PACKAGE MATERIALS INFORMATION

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### \*All dimensions are nominal

Device	Package Type	Package Drawing	Pins SPQ		Length (mm)	Width (mm)	Height (mm)	
TLV5610IYZR	DIESALE	YZ	20	3000	220.0	220.0	34.0	
TLV5610IYZT	DIESALE	YZ	20	250	220.0	220.0	34.0	

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