SLVS092G - NOVEMBER 1994 - REVISED JANUARY 2003

- Available in 5-V, 4.85-V, and 3.3-V
 Fixed-Output and Adjustable Versions
- Very Low-Dropout Voltage . . . Maximum of 32 mV at I_O = 100 mA (TPS7150)
- Very Low Quiescent Current Independent of Load . . . 285 μA Typ
- Extremely Low Sleep-State Current 0.5 μA Max
- 2% Tolerance Over Specified Conditions For Fixed-Output Versions
- Output Current Range of 0 mA to 500 mA
- TSSOP Package Option Offers Reduced Component Height for Space-Critical Applications
- Power-Good (PG) Status Output

description

The TPS71xx integrated circuits are a family of micropower low-dropout (LDO) voltage regulators. An order of magnitude reduction in dropout voltage and quiescent current over conventional LDO performance is achieved by replacing the typical pnp pass transistor with a PMOS device.

NC – No internal connection † SENSE – Fixed voltage options only (TPS7133, TPS7148, and TPS7150) ‡ FB – Adjustable version only (TPS7101)

Because the PMOS device behaves as a low-value resistor, the dropout voltage is very low (maximum of 32 mV at an output current of 100 mA for the TPS7150) and is directly proportional to the output current (see Figure 1). Additionally, since the PMOS pass element is a voltage-driven device, the quiescent current is very low and remains independent of output loading (typically 285 μ A over the full range of output current, 0 mA to 500 mA). These two key specifications yield a significant improvement in operating life for battery-powered systems. The LDO family also features a sleep mode; applying a TTL high signal to EN (enable) shuts down the regulator, reducing the quiescent current to 0.5 μ A maximum at T₁ = 25°C.

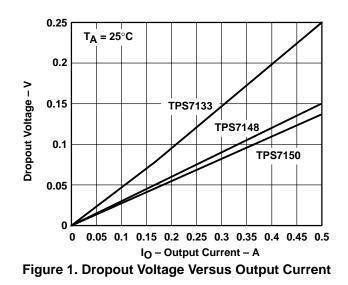


Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.

PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters. Copyright © 2003, Texas Instruments Incorporated

		PACKA VIEW)	
G <u>ND</u> [EN [IN [IN [1 2 3 4	8 7 6 5] PG] SENSE†/FB‡] OUT] OUT
-		CKAG VIEW)	—
GND [1	20	ТРG
GND GND	2	20 19	
GND	3	18	
NC	4	17	лко Л FB‡
	5	16	
EN	6	15	SENSE†
NC	7	14	Тоот
IN 🛛	8	13	Ουτ
IN [9	12] NC
IN [10	11] NC

description (continued)



Power good (PG) reports low output voltage and can be used to implement a power-on reset or a low-battery indicator.

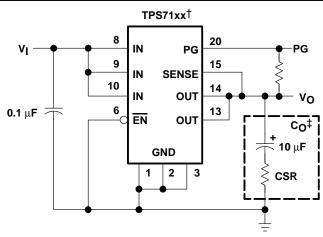
The TPS71xx is offered in 3.3-V, 4.85-V, and 5-V fixed-voltage versions and in an adjustable version (programmable over the range of 1.2 V to 9.75 V). Output voltage tolerance is specified as a maximum of 2% over line, load, and temperature ranges (3% for adjustable version). The TPS71xx family is available in PDIP (8 pin), SO (8 pin), and TSSOP (20-pin) packages. The TSSOP has a maximum height of 1,2 mm.

т.	OUTPU	JT VOLT (V)	AGE	PAG	CKAGED DEVICE	S	CHIP FORM
Тյ	MIN	TYP	MAX	SMALL OUTLINE (D)	PLASTIC DIP (P)	TSSOP (PW)	(Y)
	4.9	5	5.1	TPS7150QD	TPS7150QP	TPS7150QPW	TPS7150Y
	4.75	4.85	4.95	TPS7148QD	TPS7148QP	TPS7148QPW	TPS7148Y
-40°C to 125°C	3.23	3.3	3.37	TPS7133QD	TPS7133QP	TPS7133QPW	TPS7133Y
		ljustable [†] V to 9.75		TPS7101QD	TPS7101QP	TPS7101QPW	TPS7101Y

AVAILABLE OPTIONS

[†]The D and PW packages are available taped and reeled. Add R suffix to device type (e.g., TPS7150QDR). The TPS7101Q is programmable using an external resistor divider (see application information). The chip form is tested at 25°C.



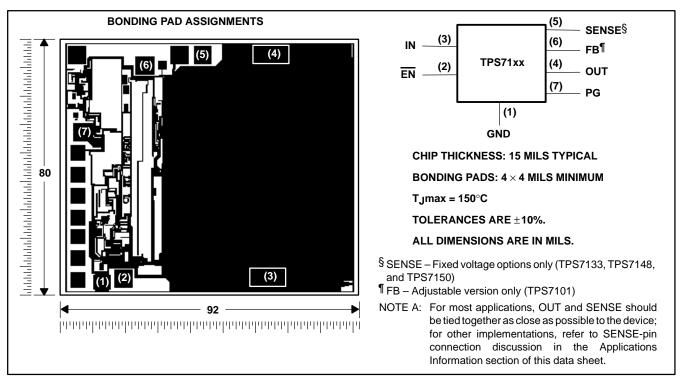


[†] TPS7133, TPS7148, TPS7150 (fixed-voltage options)
 [‡] Capacitor selection is nontrivial. See application information section for details.

Figure 2. Typical Application Configuration

TPS71xx chip information

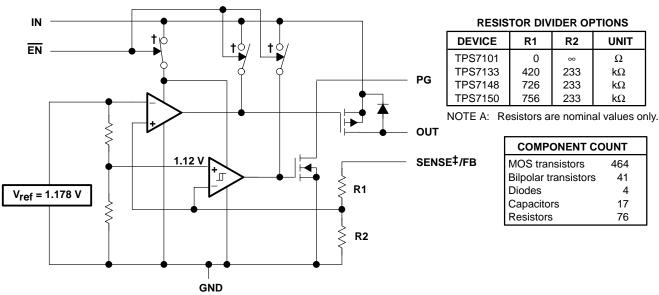
These chips, when properly assembled, display characteristics similar to the TPS71xxQ. Thermal compression or ultrasonic bonding may be used on the doped aluminum bonding pads. The chips may be mounted with conductive epoxy or a gold-silicon preform.



WWW.BSTEFICE BEAGSSER MENTSOM/TI

SLVS092G - NOVEMBER 1994 - REVISED JANUARY 2003

functional block diagram



[†] Switch positions are shown with \overline{EN} low (active).

[‡] For most applications, SENSE should be externally connected to OUT as close as possible to the device. For other implementations, refer to SENSE-pin connection discussion in Applications Information section.

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)§

Input voltage range¶, V _I , PG, SENSE, EN	–0.3 V to 11 V
Output current, IO	
Continuous total power dissipation	See Dissipation Rating Tables 1 and 2
Operating virtual junction temperature range, T _J	–55°C to 150°C
Storage temperature range, T _{stg} Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	–65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	

Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

 \P All voltage values are with respect to network terminal ground.

DISSIPATION RATING TABLE 1 – FREE-AIR TEMPERATURE (see Figure 3)#

			· ·	0 ,
PACKAGE	$T_A \le 25^{\circ}C$ POWER RATING	DERATING FACTOR ABOVE T _A = 25°C	T _A = 70°C POWER RATING	T _A = 125°C POWER RATING
D	725 mW	5.8 mW/°C	464 mW	145 mW
Р	1175 mW	9.4 mW/°C	752 mW	235 mW
PWI	700 mW	5.6 mW/°C	448 mW	140 mW

DISSIPATION RATING TABLE 2 – CASE TEMPERATURE (see Figure 4)[#]

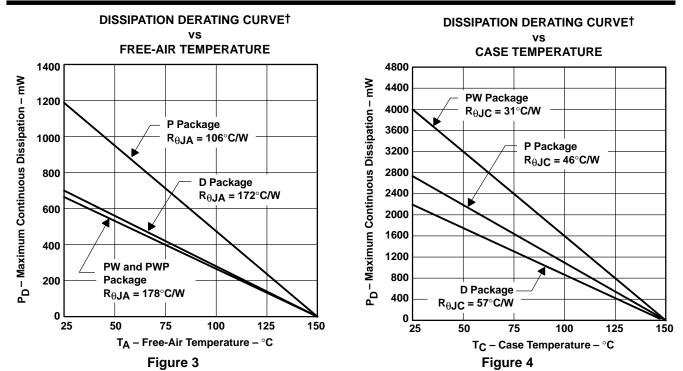
PACKAGE	T _C ≤ 25°C POWER RATING	DERATING FACTOR ABOVE T _C = 25°C	T _C = 70°C POWER RATING	T _C = 125°C POWER RATING
D	2188 mW	17.5 mW/°C	1400 mW	438 mW
Р	2738 mW	21.9 mW/°C	1752 mW	548 mW
PWI	4025 mW	32.2 mW/°C	2576 mW	805 mW

[#] Dissipation rating tables and figures are provided for maintenance of junction temperature at or below absolute maximum temperature of 150°C. For guidelines on maintaining junction temperature within recommended operating range, see the Thermal Information section.

Refer to Thermal Information section for detailed power dissipation considerations when using the TSSOP packages.



SLVS092G - NOVEMBER 1994 - REVISED JANUARY 2003



[†] Dissipation rating tables and figures are provided for maintenance of junction temperature at or below absolute maximum temperature of 150°C. For guidelines on maintaining junction temperature within recommended operating range, see the Thermal Information section.

recommended operating conditions

		MIN	MAX	UNIT
	TPS7101Q	2.5	10	
Input voltage, ∨ _I ‡	TPS7133Q	3.77	10	v
	TPS7148Q	5.2	10	v
	TPS7150Q	5.33	10	
High-level input voltage at EN, V _{IH}		2		V
Low-level input voltage at $\overline{\sf EN}$, V _{IL}			0.5	V
Output current range, IO		0	500	mA
Operating virtual junction temperature ran	ge, TJ	-40	125	°C

[‡] Minimum input voltage defined in the recommended operating conditions is the maximum specified output voltage plus dropout voltage at the maximum specified load range. Since dropout voltage is a function of output current, the usable range can be extended for lighter loads. To calculate the minimum input voltage for your maximum output current, use the following equation: V_{I(min)} = V_{O(max)} + V_{DO(max load)} Because the TPS7101 is programmable, r_{DS(on)} should be used to calculate V_{DO} before applying the above equation. The equation for calculating V_{DO} from r_{DS(on)} is given in Note 2 in the electrical characteristics table. The minimum value of 2.5 V is the absolute lower limit for the recommended input voltage range for the TPS7101.



SLVS092G - NOVEMBER 1994 - REVISED JANUARY 2003

electrical characteristics at I_O = 10 mA, \overline{EN} = 0 V, C_O = 4.7 μ F/CSR[†] = 1 Ω , SENSE/FB shorted to OUT (unless otherwise noted)

PARAMETER	TEST CON	TEST CONDITIONS [‡]		TPS7101Q, TPS7133Q TPS7148Q, TPS7150Q			UNIT	
			TJ	MIN	TYP			
Ground current (active mode)	$\overline{EN} \le 0.5 \text{ V},$	VI = VO + 1 V,	25°C		285	350	μA	
Ground current (active mode)	$0 \text{ mA} \le I_O \le 500 \text{ mA}$		-40°C to 125°C			460	μΑ	
Input current (standby mode)		25°C			0.5	μA		
input current (standby mode)	$\overline{EN} = V_{I},$	$2.7 \text{ V} \leq \text{V}_{I} \leq 10 \text{ V}$	-40°C to 125°C			2	μΑ	
Output current limit	V _O = 0,	Vj = 10 V	25°C		1.2	2	Α	
	VO = 0,	v] = 10 v	-40°C to 125°C			2		
Pass-element leakage current in standby		07// / < 40//</td <td>25°C</td> <td></td> <td></td> <td>0.5</td> <td></td>	25°C			0.5		
mode	$\overline{EN} = V_{I}, \qquad 2.7 \; V \le V_{I} \le 10 \; V$		-40°C to 125°C			1	μA	
PG leakage current	Normal operation,	V _{PG} = 10 V	25°C		0.02	0.5	μA	
			-40°C to 125°C			0.5	μΑ	
Output voltage temperature coefficient			-40°C to 125°C		61	75	ppm/°C	
Thermal shutdown junction temperature					165		°C	
	$2.5 V \le V_{I} \le 6 V$ $6 V \le V_{I} \le 10 V$		-40°C to 125°C	2			v	
EN logic high (standby mode)			-40°C 10 125°C	2.7				
			25°C			0.5	v	
EN logic low (active mode)	$2.7 \text{ V} \le \text{V}_{I} \le 10 \text{ V}$		-40°C to 125°C			0.5		
EN hysteresis voltage			25°C		50		mV	
		0.1/ - 1/ - 40.1/	25°C	-0.5		0.5		
EN input current	$0 V \le V_{I} \le 10 V$	$0 V \le V_I \le 10 V$	-40°C to 125°C	-0.5		0.5	μA	
			25°C		2.05	2.5		
Minimum VI for active pass element			-40°C to 125°C			2.5	V	
	1 200 4	1 200 (25°C		1.06	1.5	V	
Minimum V _I for valid PG	I _{PG} = 300 μA I _{PG} = 300 μA		-40°C to 125°C			1.9		

[†]CSR (compensation series resistance) refers to the total series resistance, including the equivalent series resistance (ESR) of the capacitor, any series resistance added externally, and PWB trace resistance to CO.

[‡]Pulse-testing techniques are used to maintain virtual junction temperature as close as possible to ambient temperature; thermal effects must be taken into account separately.



TPS7101 electrical characteristics at I_O = 10 mA, V_I = 3.5 V, \overline{EN} = 0 V, C_O = 4.7 μ F/CSR[†] = 1 Ω , FB shorted to OUT at device leads (unless otherwise noted)

DADAMETED	TEAT OO	T	T	PS71010	2		
PARAMETER	TEST CO	NDITIONS [‡]	Тј	MIN	TYP	MAX	UNIT
	V _I = 3.5 V,	I _O = 10 mA	25°C		1.178		V
Reference voltage (measured at FB with OUT connected to FB)	$2.5 V \le V_I \le 10 V$, See Note 1	$5 \text{ mA} \le I_O \le 500 \text{ mA},$	-40°C to 125°C	1.143		1.213	V
Reference voltage temperature coefficient			-40°C to 125°C		61	75	ppm/°C
	V ₁ = 2.4 V,	$50 + 4 \leq 1 \leq 150 = 4$	25°C		0.7	1	
	v] = 2.4 v,	$50 \ \mu A \le I_O \le 150 \ mA$	-40°C to 125°C			1	
	V ₁ = 2.4 V,	150 mA ≤ I_{O} ≤ 500	25°C		0.83	1.3	
Pass-element series resistance	$v_{1} = 2.4 v_{2}$	mA	-40°C to 125°C			1.3	Ω
(see Note 2)	V ₁ = 2.9 V,		25°C		0.52	0.85	52
	V] = 2.9 V,	$50 \ \mu A \leq I_{O} \leq 500 \ mA$	-40°C to 125°C			0.85	
	V _I = 3.9 V,	$50 \ \mu A \le I_O \le 500 \ mA$	25°C		0.32		
	V _I = 5.9 V,	$50 \ \mu A \leq I_O \leq 500 \ mA$	25°C		0.23		
Input regulation	V _I = 2.5 V to 10 V, See Note 1	50 $\mu A \leq I_{O} \leq$ 500 mA,	25°C			18	mV
Input regulation			-40°C to 125°C			25	
	$I_{O} = 5 \text{ mA to } 500 \text{ mA},$	$2.5 \text{ V} \leq \text{V}_{I} \leq 10 \text{ V},$	25°C			14	mV
Output regulation	See Note 1		-40°C to 125°C			25	
Oupurregulation	$I_{O} = 50 \ \mu A$ to 500 mA, See Note 1	$2.5 \text{ V} \leq \text{V}_{I} \leq 10 \text{ V},$	25°C			22	mV
			-40°C to 125°C			54	
			25°C	48	59		-
Ripple rejection	f = 120 Hz	I _O = 50 μA	-40°C to 125°C	44			
Ripple rejection	1 = 120 HZ	I _O = 500 mA,	25°C	45	54		dB
		See Note 1	-40°C to 125°C	44			
Output noise-spectral density	f = 120 Hz		25°C		2		μV/√Hz
		C _O = 4.7 μF	25°C		95		
Output noise voltage	10 Hz \leq f \leq 100 kHz, CSR [†] = 1 Ω	C _O = 10 μF	25°C		89		μVrms
	001(1 = 1 22	C _O = 100 μF	25°C		74		
PG trip-threshold voltage§	V _{FB} voltage decreasing	g from above V _{PG}	-40°C to 125°C	1.101		1.145	V
PG hysteresis voltage§	Measured at VFB		25°C		12		mV
		N/ 0.40.\/	25°C		0.1	0.4	
PG output low voltage§	I _{PG} = 400 μA,	V _I = 2.13 V	-40°C to 125°C			0.4	- V
			25°C	-10	0.1	10	
FB input current			-40°C to 125°C	-20		20	nA

[†]CSR refers to the total series resistance, including the ESR of the capacitor, any series resistance added externally, and PWB trace resistance to C_O.

[‡] Pulse-testing techniques are used to maintain virtual junction temperature as close as possible to ambient temperature; thermal effects must be taken into account separately.

§ Output voltage programmed to 2.5 V with closed-loop configuration (see application information).

NOTES: 1. When $V_I < 2.9$ V and $I_O > 150$ mA simultaneously, pass element $r_{DS(On)}$ increases (see Figure 27) to a point such that the resulting dropout voltage prevents the regulator from maintaining the specified tolerance range.

2. To calculate dropout voltage, use equation:

 $V_{DO} = I_O \cdot r_{DS(on)}$

 $r_{DS(on)}$ is a function of both output current and input voltage. The parametric table lists $r_{DS(on)}$ for V_I = 2.4 V, 2.9 V, 3.9 V, and 5.9 V, which corresponds to dropout conditions for programmed output voltages of 2.5 V, 3 V, 4 V, and 6 V, respectively. For other programmed values, refer to Figure 26.



SLVS092G - NOVEMBER 1994 - REVISED JANUARY 2003

TPS7133 electrical characteristics at I_O = 10 mA, V_I = 4.3 V, \overline{EN} = 0 V, C_O = 4.7 μ F/CSR[†] = 1 Ω , SENSE shorted to OUT (unless otherwise noted)

	TEOTOOL	TEST CONDITIONS [‡]			PS71330	2	
PARAMETER	TEST CON	IDITIONS+	ΤJ	MIN	TYP	MAX	UNIT
Output welfe as	V _I = 4.3 V,	I _O = 10 mA	25°C		3.3		v
Output voltage	$4.3 \text{ V} \le \text{V}_{I} \le 10 \text{ V},$	$5 \text{ mA} \le \text{IO} \le 500 \text{ mA}$	-40°C to 125°C	3.23		3.37	v
	1- 10		25°C		4.5	7	
	I _O = 10 mA,	V _I = 3.23 V	-40°C to 125°C			8	
Dreneutureltere	1- 100	\/ 2 22 \/	25°C		47	60	mV
Dropout voltage	I _O = 100 mA,	V _I = 3.23 V	-40°C to 125°C			80	mv
	$I_{O} = 500 \text{ mA},$	VI = 3.23 V	25°C		235	300	
	IO = 500 IIIA,	v] = 3.23 v	-40°C to 125°C			400	
Pass-element series resistance	(3.23 V – V _O)/I _O ,	V _I = 3.23 V,	25°C		0.47	0.6	Ω
Fass-element series resistance	I _O = 500 mA		-40°C to 125°C			0.8	52
Input regulation	V _I = 4.3 V to 10 V,	$50 \ \mu A \le I_O \le 500 \ mA$	25°C			20	mV
Input regulation	$v_{\rm I} = 4.3 v 10 10 v,$	$30 \ \mu A \ge 10 \ge 300 \ \text{IIIA}$	-40°C to 125°C	27	IIIV		
	I _O = 5 mA to 500 mA,	$4.3~V \le V_I \le 10~V$	25°C		21	38	mV
Output regulation			-40°C to 125°C			75	IIIV
Output regulation	$h_{0} = 50 \text{ m} \text{ A to } 500 \text{ m} \text{ A}$	$4.3 \text{ V} \leq \text{V}_{I} \leq 10 \text{ V}$	25°C		30	60	mV
	$10 = 30 \mu \text{A} \text{ to } 300 \text{mA},$		-40°C to 125°C			120	
		I _O = 50 μA	25°C	43	54		
Ripple rejection	f = 120 Hz		-40°C to 125°C	40			dB
	1 = 120112	I _O = 500 mA	25°C	39	49		uв
		10 = 300 MA	-40°C to 125°C	36			
Output noise-spectral density	f = 120 Hz		25°C		2		μV/√Hz
		C _O = 4.7 μF	25°C		274		
Output noise voltage	10 Hz ≤ f ≤ 100 kHz, CSR [†] = 1 Ω	C _O = 10 μF	25°C		228		μVrms
	001(1 = 1 52	C _O = 100 μF	25°C		159		
PG trip-threshold voltage	VO voltage decreasing	•	-40°C to 125°C	2.868		3	V
PG hysteresis voltage			25°C		35		mV
	1 4	N 0.0.1/	25°C		0.22	0.4	
PG output low voltage	I _{PG} = 1 mA,	V _I = 2.8 V	-40°C to 125°C			0.4	V

† CSR refers to the total series resistance, including the ESR of the capacitor, any series resistance added externally, and PWB trace resistance to CO.

[‡] Pulse-testing techniques are used to maintain virtual junction temperature as close as possible to ambient temperature; thermal effects must be taken into account separately.



TPS7148 electrical characteristics at $I_0 = 10 \text{ mA}$, $V_1 = 5.85 \text{ V}$, $\overline{\text{EN}} = 0 \text{ V}$, $C_0 = 4.7 \,\mu\text{F/CSR}^{\dagger} = 1 \,\Omega$, SENSE shorted to OUT (unless otherwise noted)

	TEST CON	TEST CONDITIONS [‡]			PS71480	2	
PARAMETER	TEST COM	IDITIONS+	Тj	MIN	TYP	MAX	UNIT
Output voltogo	VI = 5.85 V,	I _O = 10 mA	25°C		4.85		v
Output voltage	5.85 V \leq V _I \leq 10 V,	$5 \text{ mA} \le \text{IO} \le 500 \text{ mA}$	-40°C to 125°C	4.75		4.95	l v
	10	V _I = 4.75 V	25°C		2.9	6	
Dropout voltage	I _O = 10 mA,		-40°C to 125°C			8	
	la 100 mA	\/. 47E\/	25°C		30	37	
	I _O = 100 mA,	V _I = 4.75 V	-40°C to 125°C			54	mV
	La 500 m A		25°C		150	180	
	I _O = 500 mA,	V _I = 4.75 V	-40°C to 125°C			250	
	(4.75 V – V _O)/I _O ,	VI = 4.75 V,	25°C		0.32	0.35	
Pass-element series resistance	$I_{O} = 500 \text{ mA}$	•	-40°C to 125°C			0.52	2 Ω
	V _I = 5.85 V to 10 V,	$50 \ \mu A \le I_{O} \le 500 \ mA$	25°C			27	mV
nput regulation			-40°C to 125°C			37	
	I _O = 5 mA to 500 mA,	$5.85~V \le V_I \le 10~V$	25°C		12	42	
			-40°C to 125°C			80	mV
Output regulation	$I_{O} = 50 \ \mu A \text{ to } 500 \ \text{mA},$	5.85 V \leq VJ \leq 10 V	25°C		42	60	mV
			-40°C to 125°C			130	
		I _O = 50 μA	25°C	42	53		
Dinale rejection	f = 120 Hz		-40°C to 125°C	39			
Ripple rejection	1 = 120 HZ		25°C	39	50		dB
		I _O = 500 mA	-40°C to 125°C	35			
Output noise-spectral density	f = 120 Hz		25°C		2		μV/√Hz
		C _O = 4.7 μF	25°C		410		
Output noise voltage	10 Hz ≤ f ≤ 100 kHz, CSR [†] = 1 Ω	C _O = 10 μF	25°C		328		μVrms
	03(1 = 1 22	C _O = 100 μF	25°C		212		
PG trip-threshold voltage	VO voltage decreasing	U	-40°C to 125°C	4.5		4.7	V
PG hysteresis voltage			25°C		50		mV
		V: 442.V	25°C		0.2	0.4	V
PG output low voltage	IPG = 1.2 mA,	V _I = 4.12 V	-40°C to 125°C			0.4	V

[†]CSR refers to the total series resistance, including the ESR of the capacitor, any series resistance added externally, and PWB trace resistance to C_O.

[‡] Pulse-testing techniques are used to maintain virtual junction temperature as close as possible to ambient temperature; thermal effects must be taken into account separately.



SLVS092G - NOVEMBER 1994 - REVISED JANUARY 2003

TPS7150 electrical characteristics at I_O = 10 mA, V_I = 6 V, \overline{EN} = 0 V, C_O = 4.7 μ F/CSR[†] = 1 Ω , SENSE shorted to OUT (unless otherwise noted)

	TEAT OOL	τ.	TPS7150Q				
PARAMETER	TEST CON	IDITIONS+	Тј	MIN	TYP	MAX	UNIT
Outrout velta an	V _I = 6 V,	I _O = 10 mA	25°C		5		v
Output voltage	$6 \text{ V} \leq \text{V}_I \leq 10 \text{ V},$	$5 \text{ mA} \le \text{I}_{O} \le 500 \text{ mA}$	-40°C to 125°C	4.9		5.1	V V
	1. 10	N/ 100 N/	25°C		2.9	6	
Dropout voltage	I _O = 10 mA,	V _I = 4.88 V	-40°C to 125°C			8	1
	1- 100 1	\/	25°C		27	32	
	I _O = 100 mA,	V _I = 4.88 V	-40°C to 125°C			47	mV
	La 500 m A	\/- 4.00.\/	25°C		146	170	1
	I _O = 500 mA,	V _I = 4.88 V	-40°C to 125°C			230	1
	(4.88 V – V _O)/I _O ,	VI = 4.88 V,	25°C		0.29	0.32	
Pass-element series resistance	$I_{O} = 500 \text{ mA}$	•	-40°C to 125°C			0.47	Ω
	$V_{I} = 6 V \text{ to } 10 V,$	$50 \ \mu A \le I_{O} \le 500 \ mA$	25°C			25	mV
put regulation			-40°C to 125°C			32	III V
	$I_{O} = 5 \text{ mA to } 500 \text{ mA},$	$6 V \le V_I \le 10 V$	25°C		30	45	mV
			-40°C to 125°C			86	mv
Output regulation		$6 V \le V_I \le 10 V$	25°C		45	65	mV
	$10 = 50 \mu A \ 10 \ 500 \mu A$		-40°C to 125°C			140	
		l _O = 50 μA	25°C	45	55		dB
Ripple rejection	f = 120 Hz		-40°C to 125°C	40			
Rippie rejection	T = 120 HZ	I _O = 500 mA	25°C	42	52		
		IQ = 300 IIIA	-40° C to 125° C	36			
Output noise-spectral density	f = 120 Hz		25°C		2		μV/√Hz
		C _O = 4.7 μF	25°C		430		
Output noise voltage	10 Hz ≤ f ≤ 100 kHz, CSR† = 1 Ω	C _O = 10 μF	25°C		345		μVrms
	0311 = 1 22	C _O = 100 μF	25°C		220		1
PG trip-threshold voltage	VO voltage decreasing	-	-40°C to 125°C	4.55		4.75	V
PG hysteresis voltage			25°C		53		mV
-		N/ 4.05.1/	25°C		0.2	0.4	
PG output low voltage	I _{PG} = 1.2 mA,	V _I = 4.25 V	-40°C to 125°C			0.4	V

[†]CSR refers to the total series resistance, including the ESR of the capacitor, any series resistance added externally, and PWB trace resistance to CO.

[‡] Pulse-testing techniques are used to maintain virtual junction temperature as close as possible to ambient temperature; thermal effects must be taken into account separately.



electrical characteristics at I_O = 10 mA, \overline{EN} = 0 V, C_O = 4.7 μ F/CSR[†] = 1 Ω , T_J = 25°C, SENSE/FB shorted to OUT (unless otherwise noted)

PARAMETER	TEST CONDITIONS [‡]	TPS7101Y, TPS7133Y TPS7148Y, TPS7150Y	UNIT
		MIN TYP MAX	
Ground current (active mode)	$\label{eq:linear} \begin{array}{ll} \overline{\text{EN}} \leq 0.5 \text{ V}, & \text{V}_{\text{I}} = \text{V}_{\text{O}} + 1 \text{ V}, \\ 0 \text{ mA} \leq \text{I}_{\text{O}} \leq 500 \text{ mA} \end{array}$	285	μΑ
Output current limit	$V_{O} = 0, \qquad \qquad V_{I} = 10 \text{ V}$	1.2	А
PG leakage current	Normal operation, $V_{PG} = 10 V$	0.02	μA
Thermal shutdown junction temperature		165	°C
EN hysteresis voltage		50	mV
Minimum VI for active pass element		2.05	V
Minimum V _I for valid PG	I _{PG} = 300 μA	1.06	V

PARAMETER	TEAT A	NDITIONOT	T	TPS7101Y		
PARAMETER	TEST CO	ONDITIONS [‡]	MIN	MIN TYP MAX		
Reference voltage (measured at FB with OUT connected to FB)	V _I = 3.5 V,	I _O = 10 mA		1.178		V
	V _I = 2.4 V,	$50 \ \mu A \le I_O \le 150 \ mA$		0.7		
	V _I = 2.4 V,	$150~mA \le I_{O} \le 500~mA$		0.83		
Pass-element series resistance (see Note 2)	V _I = 2.9 V,	$50 \ \mu A \le I_O \le 500 \ mA$		0.52		Ω
	V _I = 3.9 V,	$50 \ \mu A \leq I_O \leq 500 \ mA$		0.32		
	V _I = 5.9 V,	$50 \ \mu A \leq I_O \leq 500 \ mA$		0.23		
Input regulation	V _I = 2.5 V to 10 V, See Note 1	50 $\mu A \leq I_{O} \leq$ 500 mA,			18	mV
	$2.5 V \le V_I \le 10 V$, See Note 1	$I_{O} = 5 \text{ mA to } 500 \text{ mA},$			14	mV
Output regulation	$2.5 V \le V_I \le 10 V$, See Note 1	$I_{O} = 50 \ \mu A \text{ to } 500 \ m A,$			22	mV
Ripple rejection	V _I = 3.5 V, I _O = 50 μA	f = 120 Hz,		59		dB
Output noise-spectral density	V _I = 3.5 V,	f = 120 Hz		2		μV/√Hz
	V _I = 3.5 V,	C _O = 4.7 μF		95		
Output noise voltage	$10 \text{ Hz} \le \text{f} \le 100 \text{ kHz},$	C _O = 10 μF		89		μVrms
	$CSR^{\dagger} = 1 \Omega$	C _O = 100 μF		74		
PG hysteresis voltage§	V _I = 3.5 V,	Measured at V _{FB}		12		mV
PG output low voltage§	V _I = 2.13 V,	I _{PG} = 400 μA		0.1		V
FB input current	VI = 3.5 V	VI = 3.5 V		0.1		nA

† CSR refers to the total series resistance, including the ESR of the capacitor, any series resistance added externally, and PWB trace resistance to C_O.

[‡]Pulse-testing techniques are used to maintain virtual junction temperature as close as possible to ambient temperature; thermal effects must be taken into account separately.

§ Output voltage programmed to 2.5 V with closed-loop configuration (see application information).

NOTES: 1. When VI < 2.9 V and IO > 150 mA simultaneously, pass element rDS(on) increases (see Figure 27) to a point such that the resulting dropout voltage prevents the regulator from maintaining the specified tolerance range.

2. To calculate dropout voltage, use equation:

 $V_{DO} = I_O \cdot r_{DS(on)}$

rDS(on) is a function of both output current and input voltage. The parametric table lists rDS(on) for VI = 2.4 V, 2.9 V, 3.9 V, and 5.9 V, which corresponds to dropout conditions for programmed output voltages of 2.5 V, 3 V, 4 V, and 6 V, respectively. For other programmed values, refer to Figure 26.



SLVS092G - NOVEMBER 1994 - REVISED JANUARY 2003

electrical characteristics at I_O = 10 mA, \overline{EN} = 0 V, C_O = 4.7 μ F/CSR[†] = 1 Ω , T_J = 25°C, SENSE shorted to OUT (unless otherwise noted) (continued)

	7507.00		TPS7133Y				
PARAMETER	TEST CC	NDITIONS [‡]	MIN	TYP	MAX	UNIT	
Output voltage	V _I = 4.3 V,	I _O = 10 mA		3.3		V	
	V _I = 3.23 V,	I _O = 10 mA		0.02			
Dropout voltage	V _I = 3.23 V,	I _O = 100 mA		47		mV	
	V _I = 3.23 V,	I _O = 500 mA		235			
Pass-element series resistance	$(3.23 V - V_O)/I_O,$ $I_O = 500 \text{ mA}$	V _I = 3.23 V,		0.47		Ω	
Output regulation	$4.3 \text{ V} \le \text{V}_{I} \le 10 \text{ V},$	$I_{O} = 5 \text{ mA to } 500 \text{ mA}$		21		mV	
Output regulation	4.3 V \leq V _I \leq 10 V,	$I_{O} = 50 \ \mu A$ to 500 mA		30		mV	
Pipple rejection	V _I = 4.3 V,	I _O = 50 μA		54	dB		
Ripple rejection	f = 120 Hz	I _O = 500 mA		49		uБ	
Output noise-spectral density	V _I = 4.3 V,	f = 120 Hz		2		μV/√ Hz	
	V _I = 4.3 V,	C _O = 4.7 μF		274		μVrms	
Output noise voltage	10 Hz \leq f \leq 100 kHz,	C _O = 10 μF		228			
	$CSR^{\dagger} = 1 \Omega$	C _O = 100 μF		159			
PG hysteresis voltage	V _I = 4.3 V	-		35		mV	
PG output low voltage	V _I = 2.8 V,	I _{PG} = 1 mA		0.22		V	

	TEAT OF	TEST CONDITIONS [‡]				
PARAMETER	TEST CO	INDITIONS+	MIN	TYP	MAX	UNIT
Output voltage	V _I = 5.85 V,	I _O = 10 mA		4.85		V
	V _I = 4.75 V,	I _O = 10 mA		0.08		
Dropout voltage	V _I = 4.75 V,	I _O = 100 mA		30		mV
	V _I = 4.75 V,	I _O = 500 mA		150		
Pass-element series resistance	(4.75 V – V _O)/I _O , I _O = 500 mA	V _I = 4.75 V,		0.32		Ω
Output regulation	5.85 V ≤ V _I ≤ 10 V,	5.85 V \leq V $_{I}$ \leq 10 V, \qquad I $_{O}$ = 5 mA to 500 mA		12		mV
	5.85 V \leq V _I \leq 10 V,	I_{O} = 50 μ A to 500 mA		42		mV
	Vj = 5.85 V,	I _O = 50 μA		53	dB	
Ripple rejection	f = 120 Hz	IO = 500 mA		50		uБ
Output noise-spectral density	V _I = 5.85 V,	f = 120 Hz		2		μV/√Hz
	V _I = 5.85 V,	C _O = 4.7 μF		410		
Output noise voltage	10 Hz \leq f \leq 100 kHz,	C _O = 10 μF		328		μVrms
	$CSR^{\dagger} = 1 \Omega$	C _O = 100 μF		212		
PG hysteresis voltage	V _I = 5.85 V	•		50		mV
PG output low voltage	V _I = 4.12 V,	I _{PG} = 1.2 mA		0.2	0.4	V

† CSR refers to the total series resistance, including the ESR of the capacitor, any series resistance added externally, and PWB trace resistance to CO.

[‡] Pulse-testing techniques are used to maintain virtual junction temperature as close as possible to ambient temperature; thermal effects must be taken into account separately.



SLVS092G - NOVEMBER 1994 - REVISED JANUARY 2003

electrical characteristics at I_O = 10 mA, \overline{EN} = 0 V, C_O = 4.7 μ F/CSR[†] = 1 Ω , T_J = 25°C, SENSE shorted to OUT (unless otherwise noted) (continued)

			TPS7150Y				
PARAMETER	TEST C	ONDITIONS [‡]	MIN	TYP	MAX	UNIT	
Output voltage	V _I = 6 V,	I _O = 10 mA		5		V	
	V _I = 4.88 V,	I _O = 10 mA		0.13			
Dropout voltage	V _I = 4.88 V,	I _O = 100 mA		27		mV	
	V _I = 4.88 V,	I _O = 500 μA		146			
Pass-element series resistance	$(4.88 V - V_O)/I_O,$ $I_O = 500 mA$	V _I = 4.88 V,		0.29		Ω	
Output regulation	6 V ≤ V _I ≤ 10 V,	$6 \text{ V} \le \text{V}_{I} \le 10 \text{ V}, \qquad I_{O} = 5 \text{ mA to } 500 \text{ mA}$				mV	
	$6 V \le V_I \le 10 V$,	$I_{O} = 50 \ \mu A$ to 500 mA		45		mV	
	VI = 6 V,	I _O = 50 μA		55	dB		
Ripple rejection	f = 120 Hz	IO = 500 mA		52		aв	
Output noise-spectral density	V _I = 6 V,	f = 120 Hz		2		μV/√Hz	
	VI = 6 V,	C _O = 4.7 μF		430			
Output noise voltage	10 Hz ≤ f ≤ 100 kHz,	C _O = 10 μF		345		μVrms	
	$CSR^{\dagger} = 1 \Omega$	C _O = 100 μF		220			
PG hysteresis voltage	V _I = 6 V	-		53		mV	
PG output low voltage	V _I = 4.25 V,	PG = 1.2 mA		0.2		V	

[†]CSR refers to the total series resistance, including the ESR of the capacitor, any series resistance added externally, and PWB trace resistance to CO.

[‡]Pulse-testing techniques are used to maintain virtual junction temperature as close as possible to ambient temperature; thermal effects must be taken into account separately.

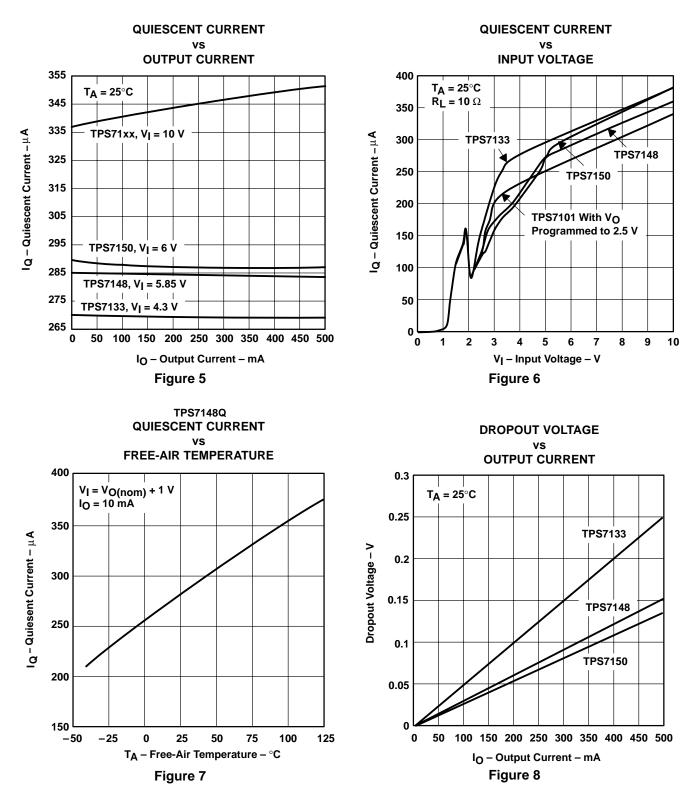


TYPICAL CHARACTERISTICS

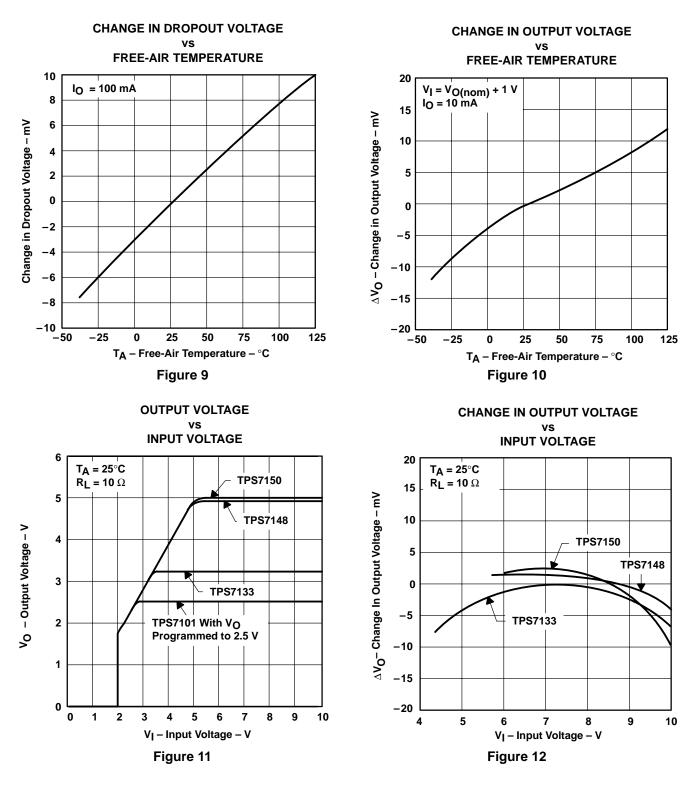
Table of Graphs

			FIGURE
		vs Output current	5
IQ	Quiescent current	vs Input voltage	6
		vs Free-air temperature	7
VDO	Dropout voltage	vs Output current	8
ΔVDO	Change in dropout voltage	vs Free-air temperature	9
ΔVO	Change in output voltage	vs Free-air temperature	10
VO	Output voltage	vs Input voltage	11
ΔVO	Change in output voltage	vs Input voltage	12
			13
N/ -	Output up the set		14
Vo	Output voltage	vs Output current	15
			16
			17
	Pipple rejection		18
	Ripple rejection	vs Frequency	19
			20
			21
			22
	Output spectral noise density	vs Frequency	23
			24
^r DS(on)	Pass-element resistance	vs Input voltage	25
R	Divider resistance	vs Free-air temperature	26
II(SENSE)	SENSE pin current	vs Free-air temperature	27
	FB leakage current	vs Free-air temperature	28
ν .	Minimum input voltage for active-pass element	vs Free-air temperature	29
VI	Minimum input voltage for valid PG	vs Free-air temperature	30
l _{l(EN)}	Input current (EN)	vs Free-air temperature	31
	Output voltage response from Enable (\overline{EN})		32
V _{PG}	Power-good (PG) voltage	vs Output voltage	33
			34
CSR	Compensation series resistance	vs Output current	35
000			36
CSR	Compensation series resistance	vs Added ceramic capacitance	37
000			38
CSR	Compensation series resistance	vs Output current	39
000			40
CSR	Compensation series resistance	vs Added ceramic capacitance	41





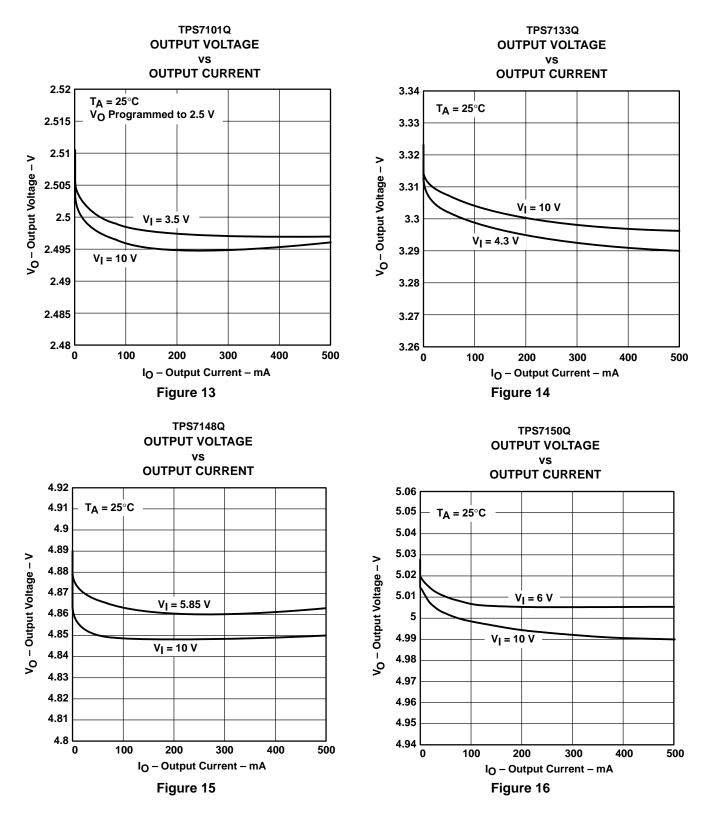






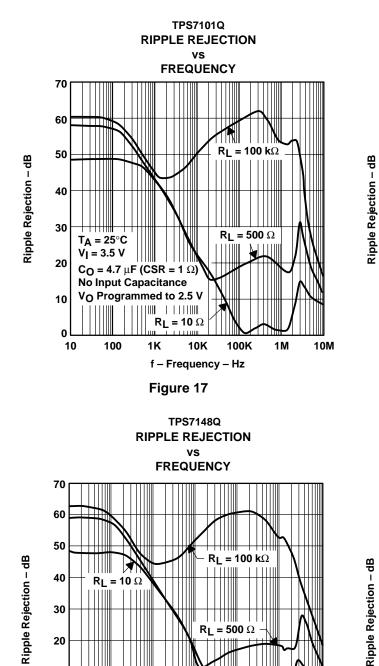
SLVS092G - NOVEMBER 1994 - REVISED JANUARY 2003

TYPICAL CHARACTERISTICS



TEXAS RUMENTS WWW.

TYPICAL CHARACTERISTICS



R_I = 500 Ω

100 k

1 M

10 M

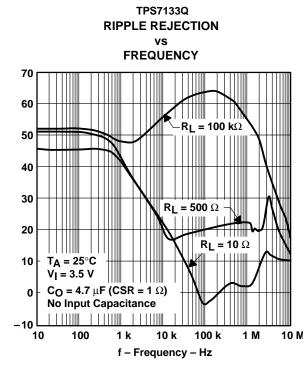
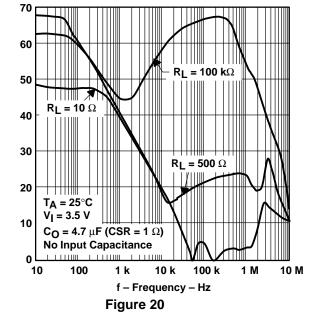


Figure 18

TPS7150Q RIPPLE REJECTION vs FREQUENCY





40

30

20

10

0

-10

10

R_L = 10 Ω

T∆ = 25°C

1.1.1.1.00

= 3.5 V

100

 $C_0 = 4.7 \ \mu F (CSR = 1 \ \Omega)$

No Input Capacitance

1.1.1.100

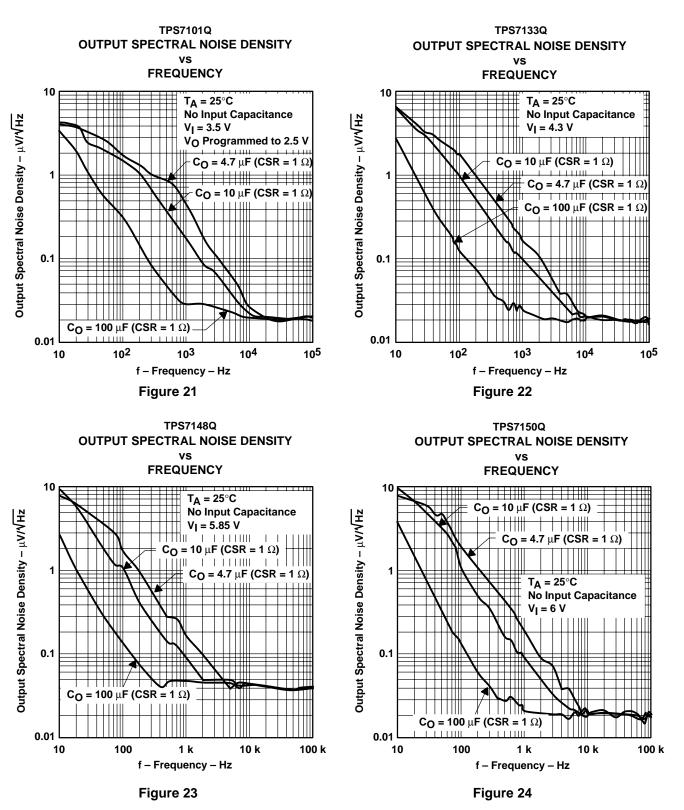
1 k

1.1.1.111

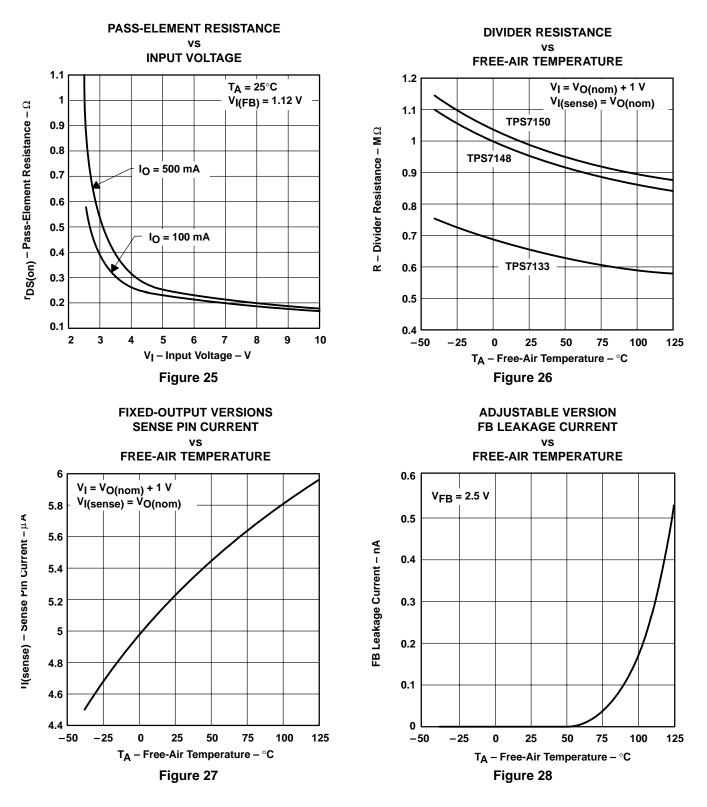
Figure 19

10 k

f - Frequency - Hz

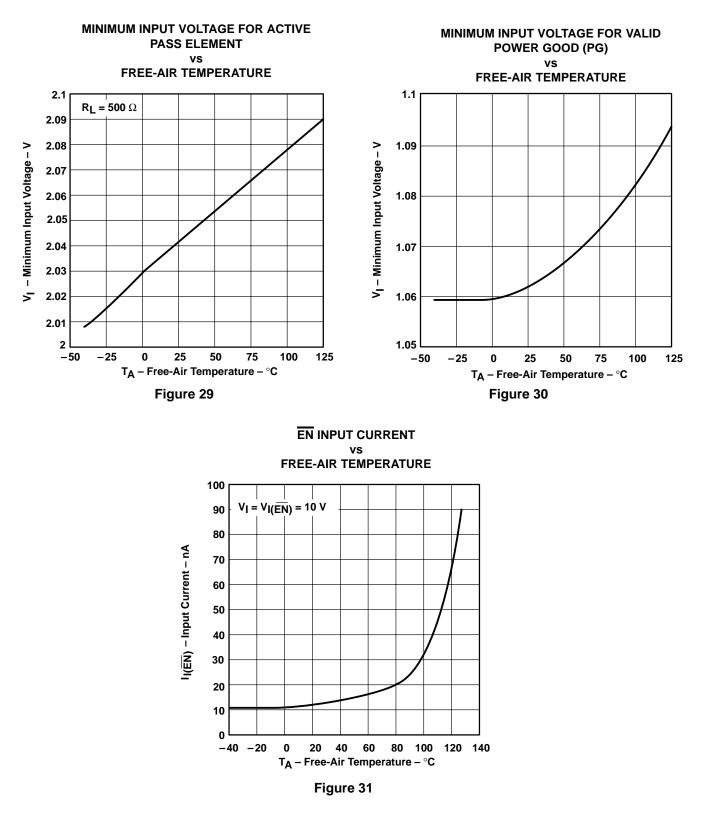








SLVS092G - NOVEMBER 1994 - REVISED JANUARY 2003





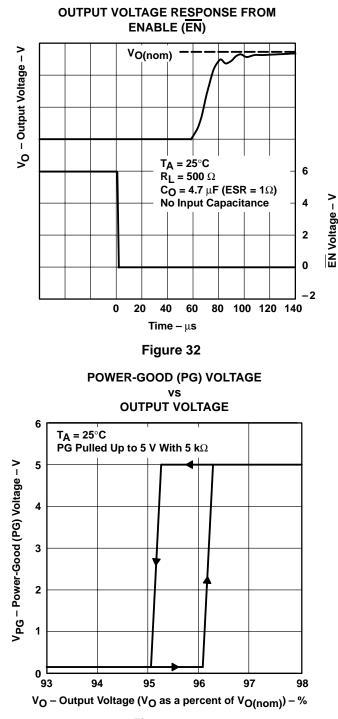
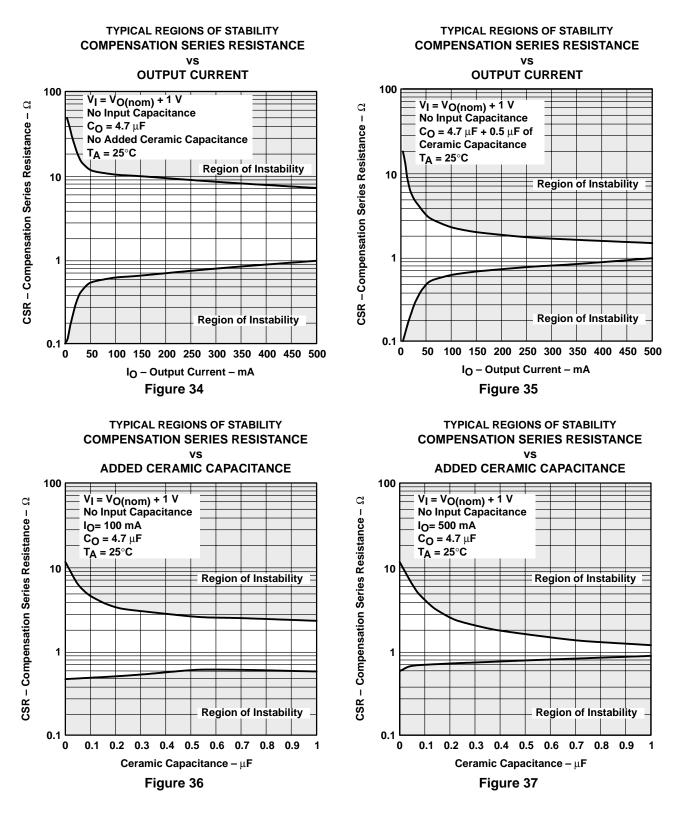
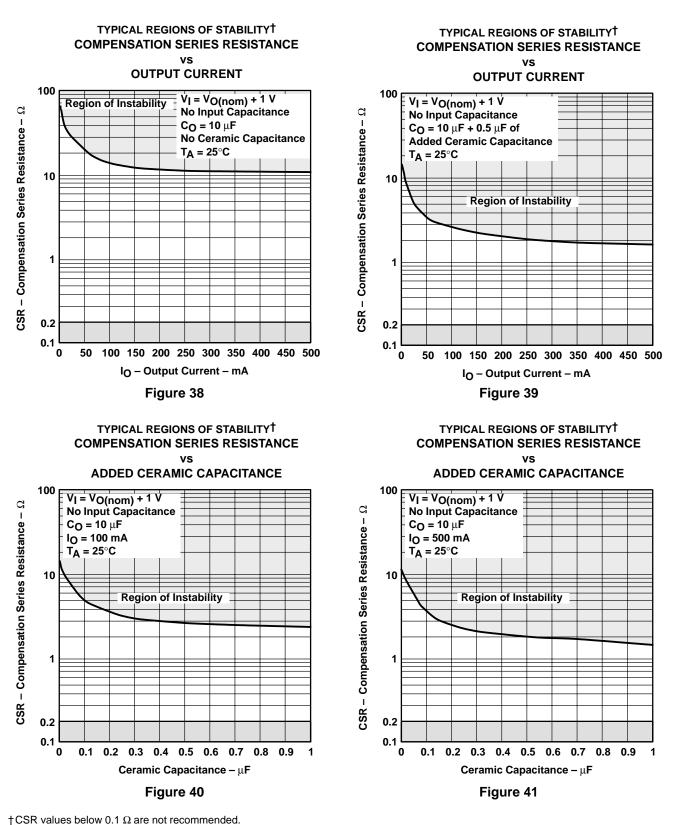


Figure 33



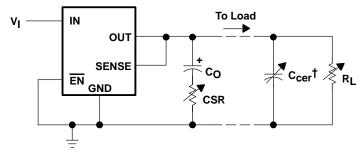








TYPICAL CHARACTERISTICS



[†]Ceramic capacitor





SLVS092G - NOVEMBER 1994 - REVISED JANUARY 2003

APPLICATION INFORMATION

The TPS71xx series of low-dropout (LDO) regulators is designed to overcome many of the shortcomings of earlier-generation LDOs, while adding features such as a power-saving shutdown mode and a power-good indicator. The TPS71xx family includes three fixed-output voltage regulators: the TPS7133 (3.3 V), the TPS7148 (4.85 V), and the TPS7150 (5 V). The family also offers an adjustable device, the TPS7101 (adjustable from 1.2 V to 9.75 V).

device operation

The TPS71xx, unlike many other LDOs, features very low quiescent currents that remain virtually constant even with varying loads. Conventional LDO regulators use a pnp-pass element, the base current of which is directly proportional to the load current through the regulator ($I_{\rm B} = I_{\rm C}/\beta$). Close examination of the data sheets reveals that those devices are typically specified under near no-load conditions; actual operating currents are much higher as evidenced by typical quiescent current versus load current curves. The TPS71xx uses a PMOS transistor to pass current; because the gate of the PMOS element is voltage driven, operating currents are low and invariable over the full load range. The TPS71xx specifications reflect actual performance under load.

Another pitfall associated with the pnp-pass element is its tendency to saturate when the device goes into dropout. The resulting drop in β forces an increase in I_B to maintain the load. During power up, this translates to large start-up currents. Systems with limited supply current may fail to start up. In battery-powered systems, it means rapid battery discharge when the voltage decays below the minimum required for regulation. The TPS71xx guiescent current remains low even when the regulator drops out, eliminating both problems.

Included in the TPS71xx family is a 4.85-V regulator, the TPS7148. Designed specifically for 5-V cellular systems, its 4.85-V output, regulated to within \pm 2%, allows for operation within the low-end limit of 5-V systems specified to \pm 5% tolerance; therefore, maximum regulated operating lifetime is obtained from a battery pack before the device drops out, adding crucial talk minutes between charges.

The TPS71xx family also features a shutdown mode that places the output in the high-impedance state (essentially equal to the feedback-divider resistance) and reduces quiescent current to under 2 µA. If the shutdown feature is not used. EN should be tied to ground. Response to an enable transition is guick; regulated output voltage is reestablished in typically 120 µs.

minimum load requirements

The TPS71xx family is stable even at zero load; no minimum load is required for operation.

SENSE-pin connection

The SENSE pin of fixed-output devices must be connected to the regulator output for proper functioning of the regulator. Normally, this connection should be as short as possible; however, the connection can be made near a critical circuit (remote sense) to improve performance at that point. Internally, SENSE connects to a high-impedance wide-bandwidth amplifier through a resistor-divider network and noise pickup feeds through to the regulator output. Routing the SENSE connection to minimize/avoid noise pickup is essential. Adding an RC network between SENSE and OUT to filter noise is not recommended because it can cause the regulator to oscillate.

external capacitor requirements

An input capacitor is not required; however, a ceramic bypass capacitor (0.047 pF to 0.1 μ F) improves load transient response and noise rejection if the TPS71xx is located more than a few inches from the power supply. A higher-capacitance electrolytic capacitor may be necessary if large (hundreds of milliamps) load transients with fast rise times are anticipated.

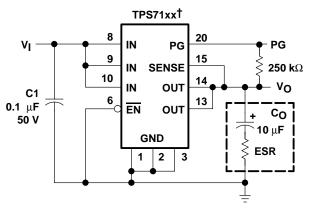


APPLICATION INFORMATION

external capacitor requirements (continued)

As with most LDO regulators, the TPS71xx family requires an output capacitor for stability. A $10-\mu$ F solid-tantalum capacitor connected from the regulator output to ground is sufficient to ensure stability over the full load range (see Figure 43). Adding high-frequency ceramic or film capacitors (such as power-supply bypass capacitors for digital or analog ICs) can cause the regulator to become unstable unless the ESR of the tantalum capacitor is less than 1.2 Ω over temperature. Where component height and/or mounting area is a problem, physically smaller, $10-\mu$ F devices can be screened for ESR. Figures 34 through 41 show the stable regions of operation using different values of output capacitance with various values of ceramic load capacitance.

In applications with little or no high-frequency bypass capacitance (< 0.2μ F), the output capacitance can be reduced to 4.7 μ F, provided ESR is maintained between the values shown in figures 34 through 41. Because minimum capacitor ESR is seldom if ever specified, it may be necessary to add a $0.5-\Omega$ to $1-\Omega$ resistor in series with the capacitor and limit ESR to 1.5Ω maximum.



[†] TPS7133, TPS7148, TPS7150 (fixed-voltage options)

Figure 43. Typical Application Circuit

programming the TPS7101 adjustable LDO regulator

Programming the adjustable regulators is accomplished using an external resistor divider as shown in Figure 44. The equation governing the output voltage is:

$$V_{O} = V_{ref} \cdot \left(1 + \frac{R1}{R2}\right)$$

where

V_{ref} = reference voltage, 1.178 V typ



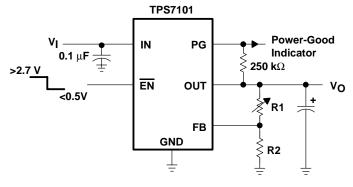
SLVS092G - NOVEMBER 1994 - REVISED JANUARY 2003

APPLICATION INFORMATION

programming the TPS7101 adjustable LDO regulator (continued)

Resistors R1 and R2 should be chosen for approximately 7-µA divider current. A recommended value for R2 is 169 k Ω with R1 adjusted for the desired output voltage. Smaller resistors can be used, but offer no inherent advantage and consume more power. Larger values of R1 and R2 should be avoided as leakage currents at FB introduce an error. Solving equation 1 for R1 yields a more useful equation for choosing the appropriate resistance:

$$R1 = \left(\frac{V_0}{V_{ref}} - 1\right) \cdot R2$$



OUTPUT VOLTAGE PROGRAMMING GUIDE

OUTPUT VOLTAGE	R1	R2	UNIT
2.5 V	191	169	kΩ
3.3 V	309	169	kΩ
3.6 V	348	169	kΩ
4 V	402	169	kΩ
5 V	549	169	kΩ
6.4 V	750	169	kΩ



power-good indicator

The TPS71xx features a power-good (PG) output that can be used to monitor the status of the regulator. The internal comparator monitors the output voltage: when the output drops to between 92% and 98% of its nominal regulated value, the PG output transistor turns on, taking the signal low. The open-drain output requires a pullup resistor. If not used, it can be left floating. PG can be used to drive power-on reset circuitry or as a low-battery indicator. PG does not assert itself when the regulated output voltage falls outside the specified 2% tolerance, but instead reports an output voltage low, relative to its nominal regulated value.

regulator protection

The TPS71xx PMOS-pass transistor has a built-in back diode that safely conducts reverse currents when the input voltage drops below the output voltage (e.g., during power down). Current is conducted from the output to the input and is not internally limited. When extended reverse voltage is anticipated, external limiting may be appropriate.

The TPS71xx also features internal current limiting and thermal protection. During normal operation, the TPS71xx limits output current to approximately 1 A. When current limiting engages, the output voltage scales back linearly until the overcurrent condition ends. While current limiting is designed to prevent gross device failure, care should be taken not to exceed the power dissipation ratings of the package. If the temperature of the device exceeds 165°C, thermal-protection circuitry shuts it down. Once the device has cooled, regulator operation resumes.





www.ti.com

23-Jul-2010

PACKAGING INFORMATION

Orderable Device	Status ⁽¹⁾	Package Type	Package Drawing	Pins	Package Qty	Eco Plan ⁽²⁾	Lead/ Ball Finish	MSL Peak Temp ⁽³⁾	Samples (Requires Login)
TPS7101QD	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	Contact TI Distributor or Sales Office
TPS7101QDG4	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	Contact TI Distributor or Sales Office
TPS7101QDR	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	Request Free Sample
TPS7101QDRG4	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	Request Free Sample
TPS7101QP	ACTIVE	PDIP	Р	8	50	Pb-Free (RoHS)	CU NIPDAU	N / A for Pkg Type	Request Free Sample
TPS7101QPE4	ACTIVE	PDIP	Р	8	50	Pb-Free (RoHS)	CU NIPDAU	N / A for Pkg Type	Request Free Sample
TPS7101QPW	ACTIVE	TSSOP	PW	20	70	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	Contact TI Distributo or Sales Office
TPS7101QPWG4	ACTIVE	TSSOP	PW	20	70	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	Contact TI Distributo or Sales Office
TPS7101QPWLE	OBSOLETE	TSSOP	PW	20		TBD	Call TI	Call TI	Samples Not Availabl
TPS7101QPWR	ACTIVE	TSSOP	PW	20	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	Request Free Sample
TPS7101QPWRG4	ACTIVE	TSSOP	PW	20	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	Request Free Sample
TPS7133QD	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	Request Free Sample
TPS7133QDG4	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	Request Free Sample
TPS7133QDR	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	Purchase Samples
TPS7133QDRG4	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	Purchase Samples
TPS7133QP	ACTIVE	PDIP	Р	8	50	Pb-Free (RoHS)	CU NIPDAU	N / A for Pkg Type	Request Free Sample
TPS7133QPE4	ACTIVE	PDIP	Р	8	50	Pb-Free (RoHS)	CU NIPDAU	N / A for Pkg Type	Request Free Sample
TPS7133QPW	ACTIVE	TSSOP	PW	20	70	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	Purchase Samples
TPS7133QPWG4	ACTIVE	TSSOP	PW	20	70	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	Purchase Samples



PACKAGE OPTION ADDENDUM



www.ti.com

23-Jul-2010

Orderable Device	Status ⁽¹⁾	Package Type	Package Drawing	Pins	Package Qty	Eco Plan ⁽²⁾	Lead/ Ball Finish	MSL Peak Temp ⁽³⁾	Samples (Requires Login)
TPS7133QPWLE	OBSOLETE	TSSOP	PW	20		TBD	Call TI	Call TI	Samples Not Available
TPS7133QPWPLE	OBSOLETE	TSSOP	PW	20		TBD	Call TI	Call TI	Samples Not Availabl
TPS7133QPWR	ACTIVE	TSSOP	PW	20	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	Request Free Sample
TPS7133QPWRG4	ACTIVE	TSSOP	PW	20	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	Request Free Sample
TPS7148QD	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	Request Free Sample
TPS7148QDG4	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	Request Free Sample
TPS7148QDR	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	Purchase Samples
TPS7148QDRG4	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	Purchase Samples
TPS7148QP	ACTIVE	PDIP	Р	8	50	Pb-Free (RoHS)	CU NIPDAU	N / A for Pkg Type	Request Free Sample
TPS7148QPE4	ACTIVE	PDIP	Р	8	50	Pb-Free (RoHS)	CU NIPDAU	N / A for Pkg Type	Request Free Sample
TPS7148QPWLE	OBSOLETE	TSSOP	PW	20		TBD	Call TI	Call TI	Samples Not Availabl
TPS7150QD	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	Request Free Sample
TPS7150QDG4	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	Request Free Sample
TPS7150QDR	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	Purchase Samples
TPS7150QDRG4	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	Purchase Samples
TPS7150QP	ACTIVE	PDIP	Р	8	50	Pb-Free (RoHS)	CU NIPDAU	N / A for Pkg Type	Request Free Sample
TPS7150QPE4	ACTIVE	PDIP	Р	8	50	Pb-Free (RoHS)	CU NIPDAU	N / A for Pkg Type	Request Free Sample
TPS7150QPW	ACTIVE	TSSOP	PW	20		Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	Purchase Samples
TPS7150QPWG4	ACTIVE	TSSOP	PW	20		Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	Purchase Samples
TPS7150QPWLE	OBSOLETE	TSSOP	PW	20		TBD	Call TI	Call TI	Samples Not Availabl

⁽¹⁾ The marketing status values are defined as follows: **ACTIVE:** Product device recommended for new designs.



PACKAGE OPTION ADDENDUM



23-Jul-2010

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect. NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design. PREVIEW: Device has been announced but is not in production. Samples may or may not be available. OBSOLETE: TI has discontinued the production of the device.

⁽²⁾ Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes. **Pb-Free (RoHS Exempt):** This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

⁽³⁾ MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

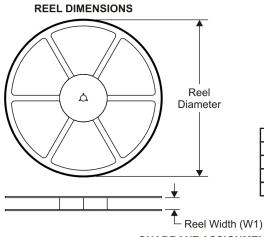
Important Information and Disclaimer: The information provided on this page represents TI's knowledge and belief as of the date that it is provided. TI bases its knowledge and belief on information provided by third parties, and makes no representation or warranty as to the accuracy of such information. Efforts are underway to better integrate information from third parties. TI has taken and continues to take reasonable steps to provide representative and accurate information but may not have conducted destructive testing or chemical analysis on incoming materials and chemicals. TI and TI suppliers consider certain information to be proprietary, and thus CAS numbers and other limited information may not be available for release.

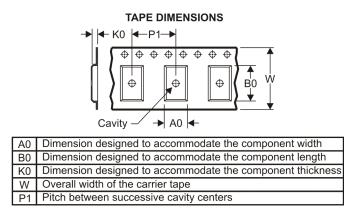
In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.



TEXAS INSTRUMENTS www.ti.com

TAPE AND REEL INFORMATION





QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



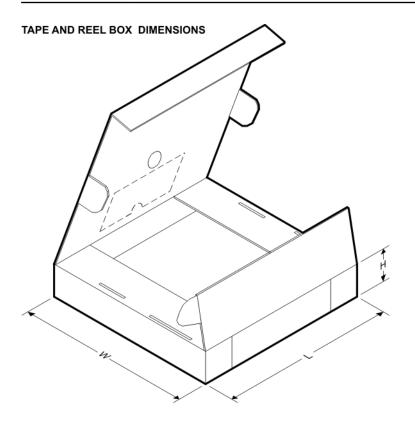
*All dimensions are nominal												
Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TPS7101QDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
TPS7101QPWR	TSSOP	PW	20	2000	330.0	16.4	6.95	7.1	1.6	8.0	16.0	Q1
TPS7133QDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
TPS7133QPWR	TSSOP	PW	20	2000	330.0	16.4	6.95	7.1	1.6	8.0	16.0	Q1
TPS7148QDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
TPS7150QDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1

www.BDTIC.com/TI



PACKAGE MATERIALS INFORMATION

9-Aug-2008



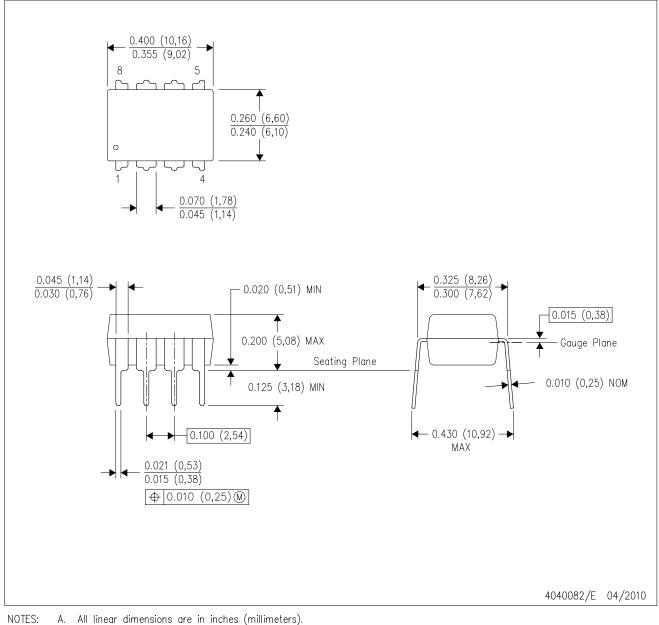
*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TPS7101QDR	SOIC	D	8	2500	346.0	346.0	29.0
TPS7101QPWR	TSSOP	PW	20	2000	346.0	346.0	33.0
TPS7133QDR	SOIC	D	8	2500	346.0	346.0	29.0
TPS7133QPWR	TSSOP	PW	20	2000	346.0	346.0	33.0
TPS7148QDR	SOIC	D	8	2500	346.0	346.0	29.0
TPS7150QDR	SOIC	D	8	2500	346.0	346.0	29.0

www.BDTIC.com/TI

P (R-PDIP-T8)

PLASTIC DUAL-IN-LINE PACKAGE

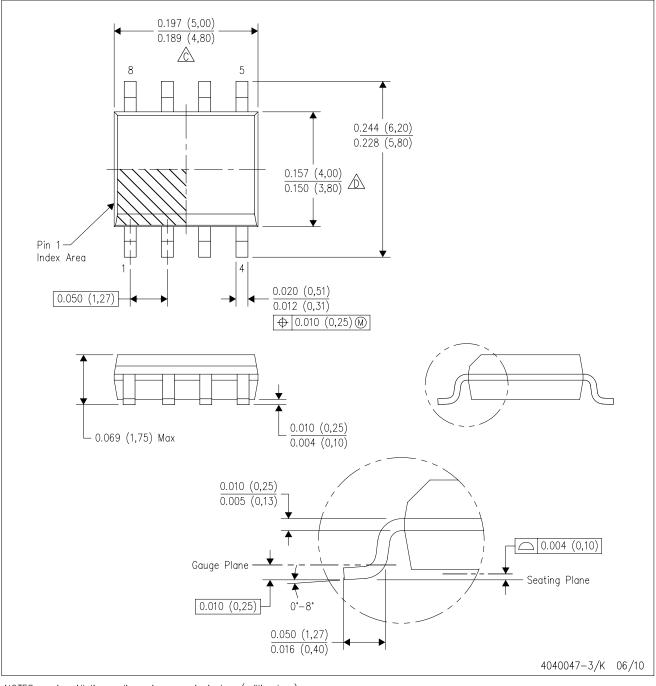


- A. All linear dimensions are in inches (millimeters).B. This drawing is subject to change without notice.
- C. Falls within JEDEC MS-001 variation BA.



D (R-PDSO-G8)

PLASTIC SMALL-OUTLINE PACKAGE



NOTES: A. All linear dimensions are in inches (millimeters).

B. This drawing is subject to change without notice.

Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed .006 (0,15) per end.

Body width does not include interlead flash. Interlead flash shall not exceed .017 (0,43) per side.

E. Reference JEDEC MS-012 variation AA.



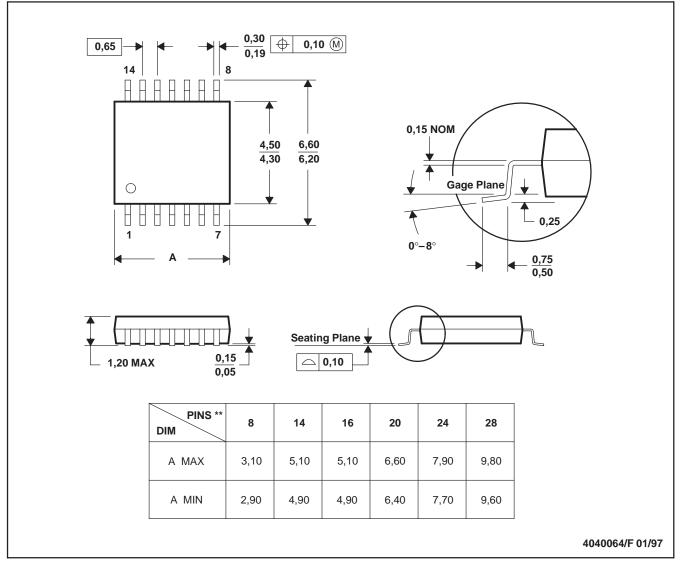
MECHANICAL DATA

MTSS001C - JANUARY 1995 - REVISED FEBRUARY 1999

PW (R-PDSO-G**)

PLASTIC SMALL-OUTLINE PACKAGE

14 PINS SHOWN



NOTES: A. All linear dimensions are in millimeters.

- B. This drawing is subject to change without notice.
- C. Body dimensions do not include mold flash or protrusion not to exceed 0,15.
- D. Falls within JEDEC MO-153



IMPORTANT NOTICE

Texas Instruments Incorporated and its subsidiaries (TI) reserve the right to make corrections, modifications, enhancements, improvements, and other changes to its products and services at any time and to discontinue any product or service without notice. Customers should obtain the latest relevant information before placing orders and should verify that such information is current and complete. All products are sold subject to TI's terms and conditions of sale supplied at the time of order acknowledgment.

TI warrants performance of its hardware products to the specifications applicable at the time of sale in accordance with TI's standard warranty. Testing and other quality control techniques are used to the extent TI deems necessary to support this warranty. Except where mandated by government requirements, testing of all parameters of each product is not necessarily performed.

TI assumes no liability for applications assistance or customer product design. Customers are responsible for their products and applications using TI components. To minimize the risks associated with customer products and applications, customers should provide adequate design and operating safeguards.

TI does not warrant or represent that any license, either express or implied, is granted under any TI patent right, copyright, mask work right, or other TI intellectual property right relating to any combination, machine, or process in which TI products or services are used. Information published by TI regarding third-party products or services does not constitute a license from TI to use such products or services or a warranty or endorsement thereof. Use of such information may require a license from a third party under the patents or other intellectual property of the third party, or a license from TI under the patents or other intellectual property of TI.

Reproduction of TI information in TI data books or data sheets is permissible only if reproduction is without alteration and is accompanied by all associated warranties, conditions, limitations, and notices. Reproduction of this information with alteration is an unfair and deceptive business practice. TI is not responsible or liable for such altered documentation. Information of third parties may be subject to additional restrictions.

Resale of TI products or services with statements different from or beyond the parameters stated by TI for that product or service voids all express and any implied warranties for the associated TI product or service and is an unfair and deceptive business practice. TI is not responsible or liable for any such statements.

TI products are not authorized for use in safety-critical applications (such as life support) where a failure of the TI product would reasonably be expected to cause severe personal injury or death, unless officers of the parties have executed an agreement specifically governing such use. Buyers represent that they have all necessary expertise in the safety and regulatory ramifications of their applications, and acknowledge and agree that they are solely responsible for all legal, regulatory and safety-related requirements concerning their products and any use of TI products in such safety-critical applications, notwithstanding any applications-related information or support that may be provided by TI. Further, Buyers must fully indemnify TI and its representatives against any damages arising out of the use of TI products in such safety-critical applications.

TI products are neither designed nor intended for use in military/aerospace applications or environments unless the TI products are specifically designated by TI as military-grade or "enhanced plastic." Only products designated by TI as military-grade meet military specifications. Buyers acknowledge and agree that any such use of TI products which TI has not designated as military-grade is solely at the Buyer's risk, and that they are solely responsible for compliance with all legal and regulatory requirements in connection with such use.

TI products are neither designed nor intended for use in automotive applications or environments unless the specific TI products are designated by TI as compliant with ISO/TS 16949 requirements. Buyers acknowledge and agree that, if they use any non-designated products in automotive applications, TI will not be responsible for any failure to meet such requirements.

Following are URLs where you can obtain information on other Texas Instruments products and application solutions:

Products		Applications	
Amplifiers	amplifier.ti.com	Audio	www.ti.com/audio
Data Converters	dataconverter.ti.com	Automotive	www.ti.com/automotive
DLP® Products	www.dlp.com	Communications and Telecom	www.ti.com/communications
DSP	dsp.ti.com	Computers and Peripherals	www.ti.com/computers
Clocks and Timers	www.ti.com/clocks	Consumer Electronics	www.ti.com/consumer-apps
Interface	interface.ti.com	Energy	www.ti.com/energy
Logic	logic.ti.com	Industrial	www.ti.com/industrial
Power Mgmt	power.ti.com	Medical	www.ti.com/medical
Microcontrollers	microcontroller.ti.com	Security	www.ti.com/security
RFID	www.ti-rfid.com	Space, Avionics & Defense	www.ti.com/space-avionics-defense
RF/IF and ZigBee® Solutions	www.ti.com/lprf	Video and Imaging	www.ti.com/video
		Wireless	www.ti.com/wireless-apps

Mailing Address: Texas Instruments, Post Office Box 655303, Dallas, Texas 75265 Copyright © 2010, Texas Instruments Incorporated

www.BDTIC.com/TI