

135µA, 14nV/√Hz, Rail-to-Rail Output Precision Op Amp with Shutdown

FEATURES

- 35µV Maximum Offset Voltage
- 110pA Maximum Input Bias Current
- 135µA Supply Current
- Rail-to-Rail Output Swing
- 12µA Supply Current in Shutdown
- 120dB Minimum Voltage Gain (V_S = ±15V)
- 0.8µV/°C Maximum V_{OS} Drift
- 14nV/√Hz Input Noise Voltage
- 2.7V to ±18V Supply Voltage Operation
- Operating Temperature Range: -40°C to 85°C
- Space Saving 3mm × 3mm DFN Package

APPLICATIONS

- Thermocouple Amplifiers
- Precision Photo Diode Amplifiers
- Instrumentation Amplifiers
- Battery-Powered Precision Systems

DESCRIPTION

The LT®6010 op amp combines low noise and high precision input performance with low power consumption and rail-to-rail output swing.

Input offset voltage is trimmed to less than $35\mu V$. The low drift and excellent long-term stability guarantee a high accuracy over temperature and over time. The 110pA maximum input bias current and 120dB minimum voltage gain further maintain this precision over operating conditions.

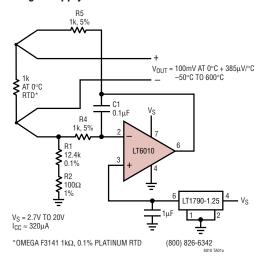
The LT6010 works on any power supply voltage from 2.7V to 36V, and draws only 135 μ A of supply current on a 5V supply. A power saving shutdown feature reduces supply current to 12 μ A. The output voltage swings to within 40mV of either supply rail, making the amplifier a good choice for low voltage single supply operation.

The LT6010 is fully specified at 5V and ± 15 V supplies and from -40° C to 85° C. The device is available in SO-8 and space-saving 3mm \times 3mm DFN packages. This op amp is also available in dual (LT6011) and quad (LT6012) packages.

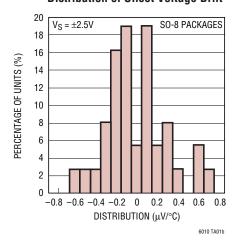
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TYPICAL APPLICATION

Single Supply Current Source for Platinum RTD



Distribution of Offset Voltage Drift



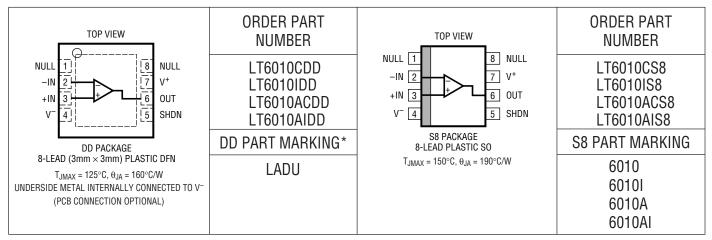


ABSOLUTE MAXIMUM RATINGS (Note 1)

Total Supply Voltage (V ⁺ to V ⁻)	40V
Differential Input Voltage (Note 2)	10V
Input Voltage, Shutdown Voltage	V+ to V ⁻
Input Current (Note 2)	±10mA
Output Short-Circuit Duration (Note 3)	Indefinite
Operating Temperature Range (Note 4)	40°C to 85°C
Specified Temperature Range (Note 5) –	40°C to 85°C

Maximum Junction Temperature	
DD Package	125°C
SO-8 Package	150°C
Storage Temperature Range	
DD Package65°C to	125°C
SO-8 Package65°C to	150°C
Lead Temperature (Soldering, 10 sec)	.300°C

PACKAGE/ORDER INFORMATION



^{*}Temperature grades are identified by a label on the shipping container. Consult LTC Marketing for parts specified with wider operating temperature ranges.

ELECTRICAL CHARACTERISTICS The \bullet denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25^{\circ}C$. $V_S = 5V$, OV; $V_{CM} = 2.5V$; R_L to OV; SHDN = 0.2V, unless otherwise specified. (Note 5)

SYMBOL	PARAMETER	CONDITIONS		MIN	TYP	MAX	UNITS
V _{OS}	Input Offset Voltage (Note 7)	LT6010AS8			10	35	μV
		$T_A = 0$ °C to 70 °C	•			60	μV
		$T_A = -40^{\circ}C$ to $85^{\circ}C$	•			75	μV
		LT6010S8			20	55	μV
		$T_A = 0$ °C to 70 °C	•			85	μV
		$T_A = -40^{\circ}C$ to $85^{\circ}C$	•			110	μV
		LT6010ADD			20	60	μV
		$T_A = 0$ °C to 70 °C	•			85	μV
		$T_A = -40^{\circ}C$ to $85^{\circ}C$	•			100	μV
		LT6010DD			30	80	μV
		$T_A = 0$ °C to 70 °C	•			110	μV
		$T_A = -40^{\circ}C$ to $85^{\circ}C$	•			135	μV
$\Delta V_{OS}/\Delta T$	Input Offset Voltage Drift (Note 6)	LT6010AS8, LT6010S8	•		0.2	0.8	μV/°C
		LT6010ADD,LT6010DD	•		0.2	1.3	μV/°C



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SYMBOL	PARAMETER	CONDITIONS		MIN	TYP	MAX	UNITS
I _{OS}	Input Offset Current (Note 7)	LT6010AS8 $T_A = 0^{\circ}C \text{ to } 70^{\circ}C$ $T_A = -40^{\circ}C \text{ to } 85^{\circ}C$	•		20	110 150 200	pA pA pA
		LT6010S8 $T_A = 0^{\circ}\text{C to } 70^{\circ}\text{C}$ $T_A = -40^{\circ}\text{C to } 85^{\circ}\text{C}$	•		40	200 300 400	pA pA pA
		LT6010ADD $T_A = 0^{\circ}C$ to $70^{\circ}C$ $T_A = -40^{\circ}C$ to $85^{\circ}C$	•		20	200 300 400	pA pA pA
		LT6010DD $T_A = 0^{\circ}C$ to $70^{\circ}C$ $T_A = -40^{\circ}C$ to $85^{\circ}C$	•		40	300 400 500	pA pA pA
I _B	Input Bias Current (Note 7)	LT6010AS8 $T_A = 0^{\circ}C \text{ to } 70^{\circ}C$ $T_A = -40^{\circ}C \text{ to } 85^{\circ}C$	•		20	±110 ±150 ±200	pA pA pA
		LT6010S8 $T_A = 0^{\circ}\text{C to } 70^{\circ}\text{C}$ $T_A = -40^{\circ}\text{C to } 85^{\circ}\text{C}$	•		40	±200 ±300 ±400	pA pA pA
		LT6010ADD $T_A = 0^{\circ}\text{C to } 70^{\circ}\text{C}$ $T_A = -40^{\circ}\text{C to } 85^{\circ}\text{C}$	•		20	±200 ±300 ±400	pA pA pA
		LT6010DD $T_A = 0^{\circ}C$ to $70^{\circ}C$ $T_A = -40^{\circ}C$ to $85^{\circ}C$	•		40	±300 ±400 ±500	pA pA pA
	Input Noise Voltage	0.1Hz to 10Hz			400		nV _{P-P}
e _n	Input Noise Voltage Density	f = 1kHz			14		nV/√Hz
i _n	Input Noise Current Density	f = 1kHz			0.1		pA/√Hz
R _{IN}	Input Resistance	Common Mode, V _{CM} = 1V to 3.8V Differential		10	120 20		GΩ MΩ
CIN	Input Capacitance				4		pF
V _{CM}	Input Voltage Range (Positive) Input Voltage Range (Negative)	Guaranteed by CMRR Guaranteed by CMRR	•	3.8	4 0.7	1	V V
CMRR	Common Mode Rejection Ratio	V _{CM} = 1V to 3.8V	•	107	135		dB
	Minimum Supply Voltage	Guaranteed by PSRR	•		2.4	2.7	V
PSRR	Power Supply Rejection Ratio	$V_S = 2.7V \text{ to } 36V, V_{CM} = 1/2V_S$	•	112	135		dB
A _{VOL}	Large-Signal Voltage Gain	$R_L = 10k$, $V_{OUT} = 1V$ to $4V$ $R_L = 2k$, $V_{OUT} = 1V$ to $4V$	•	300 250	2000 2000		V/mV V/mV
V _{OUT}	Maximum Output Swing (Positive, Referred to V+)	No Load, 50mV Overdrive	•		35	55 65	mV mV
		I _{SOURCE} = 1mA, 50mV Overdrive	•		120	170 220	mV mV
	Maximum Output Swing (Negative, Referred to 0V)	No Load, 50mV Overdrive	•		40	55 65	mV mV
		I _{SINK} = 1mA, 50mV Overdrive	•		150	225 275	mV mV



ELECTRICAL CHARACTERISTICS The \bullet denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25^{\circ}C$. $V_S = 5V$, 0V; $V_{CM} = 2.5V$; R_L to 0V; SHDN = 0.2V, unless otherwise specified. (Note 5)

SYMBOL	PARAMETER	CONDITIONS		MIN	TYP	MAX	UNITS
I _{SC} Output Short-Circuit Curr	Output Short-Circuit Current (Note 3)	V _{OUT} = 0V, 1V Overdrive (Source)	•	10 4	14		m A mA
		V _{OUT} = 5V, -1V Overdrive (Sink)	•	10 4	21		m A mA
SR	Slew Rate	$A_V = -10$, $R_F = 50k$, $R_G = 5k$ $T_A = 0^{\circ}C$ to $70^{\circ}C$ $T_A = -40^{\circ}C$ to $85^{\circ}C$	•	0.06 0.05 0.04	0.09		V/μs V/μs V/μs
GBW	Gain Bandwidth Product	f = 10kHz	•	250 225	330		kHz kHz
t _s	Settling Time	$A_V = -1$, 0.01%, $V_{OUT} = 1.5V$ to 3.5V			45		μs
t_r, t_f	Rise Time, Fall Time	A _V = 1, 10% to 90%, 0.1V Step			1		μs
I _{SHDN}	SHDN Pin Current	SHDN ≤ V ⁻ + 0.2V (On)	•			0.25	μА
		SHDN = V ⁻ + 2.0V (Off)	•		15	25	μА
t _{SHDN}	SHDN Turn-On, Turn-Off Time	SHDN = V^- (On) to V^- + 2.0V (Off) SHDN = V^- + 2.0V (Off) to V^- (On)			25 25		μs μs
I _S	Supply Current	SHDN \leq V ⁻ + 0.2V (On) T _A = 0°C to 70°C T _A = -40°C to 85°C	•		135	150 190 210	μΑ μΑ μΑ
		SHDN = V ⁻ + 2.0V (Off)	•		12	25 50	μA μA

The \bullet denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25^{\circ}C$. $V_S = \pm 15V$, $V_{CM} = 0V$, R_L to 0V; SHDN = -14.8V, unless otherwise specified. (Note 5)

SYMBOL	PARAMETER	CONDITIONS		MIN	TYP	MAX	UNITS
V _{OS} Input Offset Voltage (Note 7)	Input Offset Voltage (Note 7)	LT6010AS8 T _A = 0°C to 70°C T _A = -40°C to 85°C	•		10	60 80 110	μV μV μV
		LT6010S8 $T_A = 0^{\circ}\text{C to } 70^{\circ}\text{C}$ $T_A = -40^{\circ}\text{C to } 85^{\circ}\text{C}$	•		20	85 120 160	μV μV μV
		LT6010ADD $T_A = 0^{\circ}C \text{ to } 70^{\circ}C$ $T_A = -40^{\circ}C \text{ to } 85^{\circ}C$	•		20	85 105 135	μV μV μV
		LT6010DD T _A = 0°C to 70°C T _A = -40°C to 85°C	•		30	110 145 185	μV μV μV
$\Delta V_{0S}/\Delta T$	Input Offset Voltage Drift (Note 6)	LT6010AS8, LT6010S8 LT6010ADD,LT6010DD	•		0.2 0.2	0.8 1.3	μV/°C μV/°C
I _{OS}	Input Offset Current (Note 7)	LT6010AS8 T _A = 0°C to 70°C T _A = -40°C to 85°C	•		20	110 150 200	pA pA pA
	LT6010S8 $T_A = 0^{\circ}\text{C to } 70^{\circ}\text{C}$ $T_A = -40^{\circ}\text{C to } 85^{\circ}\text{C}$	•		40	200 300 400	pA pA pA	
		LT6010ADD T _A = 0°C to 70°C T _A = -40°C to 85°C	•		20	200 300 400	pA pA pA



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SYMBOL	PARAMETER	CONDITIONS		MIN	TYP	MAX	UNITS
I _{OS}	Input Offset Current (Note 7)	LT6010DD			40	300	pA
		$T_A = 0$ °C to 70°C	•			400	рA
		$T_A = -40^{\circ}\text{C to } 85^{\circ}\text{C}$	•			500	pA
I_{B}	Input Bias Current (Note 7)	LT6010AS8			20	±110	pA
		T _A = 0°C to 70°C	•			±150	p A
		$T_A = -40^{\circ}\text{C to } 85^{\circ}\text{C}$	•			±200	pA
		LT6010S8			40	±200	pA
		$T_A = 0^{\circ}C \text{ to } 70^{\circ}C$ $T_A = -40^{\circ}C \text{ to } 85^{\circ}C$				±300 ±400	pA
		**			00		pA
		LT6010ADD T _A = 0°C to 70°C			20	±200 ±300	pA pA
		$T_A = -40^{\circ}\text{C to } 70^{\circ}\text{C}$				±300 ±400	pA pA
		LT6010DD			40	±300	
		$T_A = 0^{\circ}C \text{ to } 70^{\circ}C$			40	±300 ±400	pA pA
		$T_A = -40^{\circ}\text{C to } 85^{\circ}\text{C}$				±500	pA
	Input Noise Voltage	0.1Hz to 10Hz			400		nV _{P-P}
e _n	Input Noise Voltage Density	f = 1kHz			13		nV/√Hz
in	Input Noise Current Density	f = 1kHz			0.1		pA/√Hz
R _{IN}	Input Resistance	Common Mode, V _{CM} = ±13.5V		50	400		GΩ
-		Differential			20		MΩ
C _{IN}	Input Capacitance				4		pF
V _{CM}	Input Voltage Range	Guaranteed by CMRR	•	±13.5	±14		V
CMRR	Common Mode Rejection Ratio	$V_{CM} = -13.5V \text{ to } 13.5V$		115	135		d B
			•	112			dB
	Minimum Supply Voltage	Guaranteed by PSRR	•		±1.2	±1.35	V
PSRR	Power Supply Rejection Ratio	$V_S = \pm 1.35 V \text{ to } \pm 18 V$	•	112	135		dB
A _{VOL}	Large-Signal Voltage Gain	$R_L = 10k$, $V_{OUT} = -13.5V$ to 13.5V	•	1000 600	2000		V/mV V/mV
		$R_L = 5k$, $V_{OUT} = -13.5V$ to 13.5V		500	1500		V/mV
			•	300			V/mV
V _{OUT}	Maximum Output Swing (Positive, Referred to V+)	No Load, 50mV Overdrive	•		45	80 100	m V mV
		I _{SOURCE} = 1mA, 50mV Overdrive	•		140	195 240	m V mV
	Maximum Output Swing	No Load, 50mV Overdrive			45	80	m V
	(Negative, Referred to 0V)		•			100	mV
		I _{SINK} = 1mA, 50mV Overdrive			150	250	m V
			•			300	mV
I _{SC}	Output Short-Circuit Current (Note 3)	V _{OUT} = 0V, 1V Overdrive (Source)		10	15		m A
			•	5			mA_
		$V_{OUT} = 0V, -1V$ Overdrive (Sink)		10	20		m A
			•	5			mA

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SYMBOL	PARAMETER	CONDITIONS		MIN	TYP	MAX	UNITS
SR	Slew Rate	$A_V = -10$, $R_F = 50k$, $R_G = 5k$ $T_A = 0^{\circ}C$ to $70^{\circ}C$ $T_A = -40^{\circ}C$ to $85^{\circ}C$	•	0.08 0.07 0.05	0.11		V/µs V/µs V/µs
GBW	Gain Bandwidth Product	f = 10kHz	•	275 250	350		kHz kHz
t _s	Settling Time	$A_V = -1$, 0.01%, $V_{OUT} = 0V$ to 10V			85		μS
t _r , t _f	Rise Time, Fall Time	A _V = 1, 10% to 90%, 0.1V Step			1		μs
I _{SHDN}	SHDN Pin Current	$SHDN \le V^- + 0.2V (On)$	•			0.25	μА
		SHDN = $V^- + 2.0V$ (Off)	•		15	25	μА
t _{SHDN}	SHDN Turn-On, Turn-Off Time	SHDN = V^- (On) to V^- + 2.0V (Off) SHDN = V^- + 2.0V (Off) to V^- (On)			25 25		μs μs
I _S	Supply Current	SHDN \leq V ⁻ + 0.2V (On) $T_A = 0^{\circ}C$ to $70^{\circ}C$ $T_A = -40^{\circ}C$ to $85^{\circ}C$	•		260	330 380 400	μΑ μΑ μΑ
		SHDN = $V^- + 2.0V$ (Off)			18	50	μА

Note 1: Absolute Maximum Ratings are those beyond which the life of the device may be impaired.

Note 2: The inputs are protected by back-to-back diodes and internal series resistors. If the differential input voltage exceeds 10V, the input current must be limited to less than 10mA.

Note 3: A heat sink may be required to keep the junction temperature below absolute maximum ratings.

Note 4: Both the LT6010C and LT6010I are guaranteed functional over the operating temperature range of -40° C to 85° C.

Note 5: The LT6010C is guaranteed to meet the specified performance

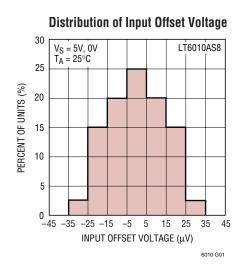
from 0°C to 70°C and is designed, characterized and expected to meet specified performance from -40°C to 85°C but is not tested or QA sampled at these temperatures. The LT6010I is guaranteed to meet specified performance from -40°C to 85°C.

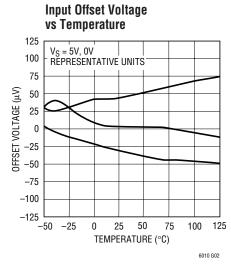
Note 6: This parameter is not 100% tested.

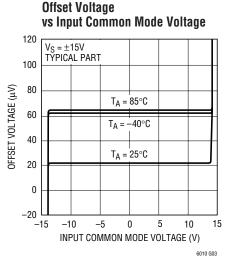
Note 7: The specifications for V_{OS} , I_B and I_{OS} depend on the grade and on the package. The following table clarifies the notations used in the specification table:

	Standard Grade	A Grade
S8 Package	LT6010S8	LT6010AS8
DFN Package	LT6010DD	LT6010ADD

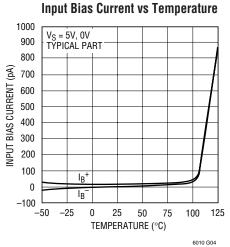
TYPICAL PERFORMANCE CHARACTERISTICS

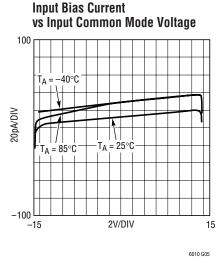


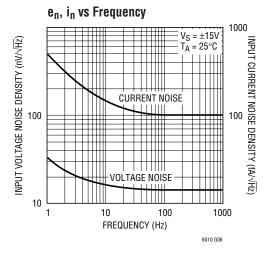




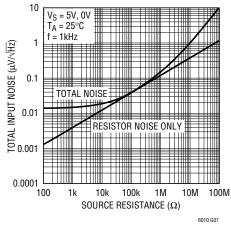
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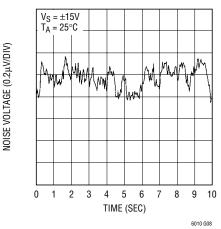




Total Input Noise vs Source Resistance

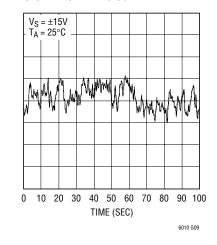




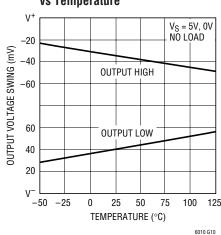


0.01Hz to 1Hz Noise

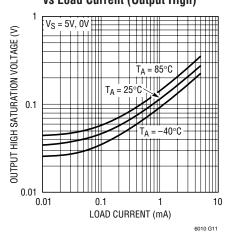
NOISE VOLTAGE (0.2µV/DIV)



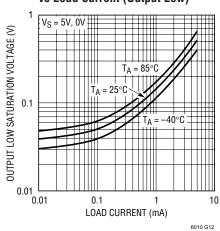
Output Voltage Swing vs Temperature



Output Saturation Voltage vs Load Current (Output High)

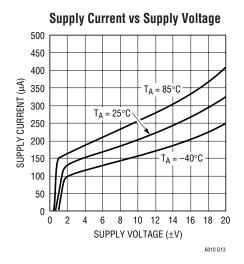


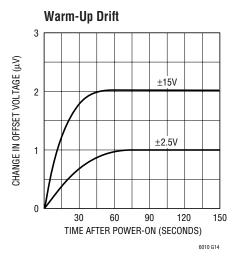
Output Saturation Voltage vs Load Current (Output Low)

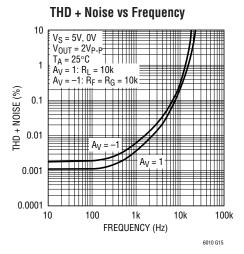


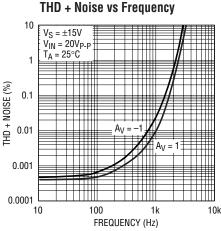


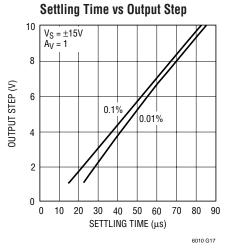
TYPICAL PERFORMANCE CHARACTERISTICS

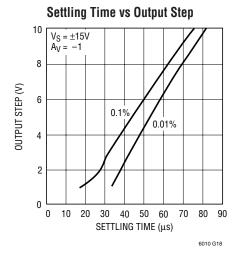


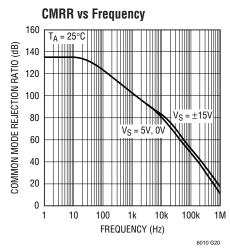




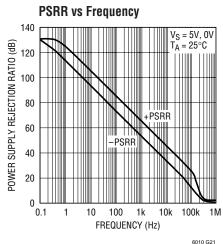




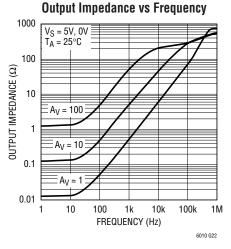


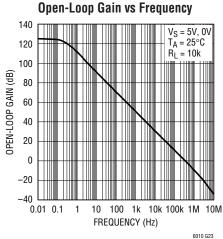


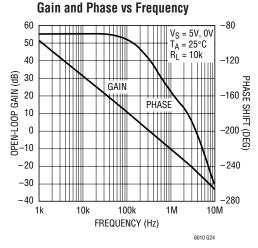
6010 G16

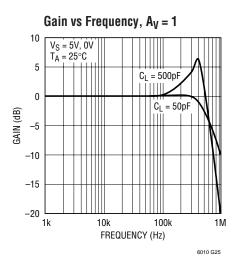


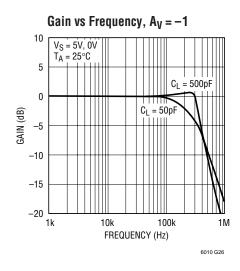
TYPICAL PERFORMANCE CHARACTERISTICS

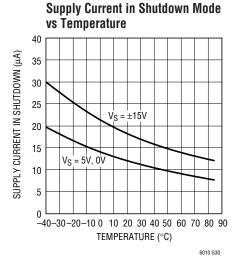


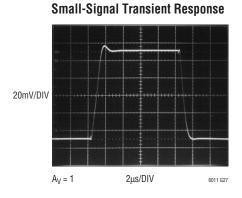


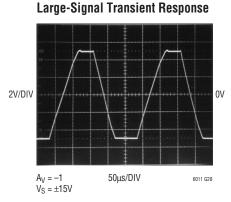


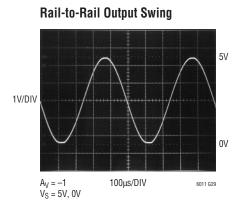














APPLICATIONS INFORMATION

Preserving Input Precision

Preserving the input accuracy of the LT6010 requires that the applications circuit and PC board layout do not introduce errors comparable to or greater than the $20\mu V$ typical offset of the amplifier. Temperature differentials across the input connections can generate thermocouple voltages of 10's of microvolts, so the connections to the input leads should be short, close together, and away from heat dissipating components. Air currents across the board can also generate temperature differentials.

The extremely low input bias currents (20pA typical) allow high accuracy to be maintained with high impedance sources and feedback resistors. The LT6010 low input bias currents are obtained by a cancellation circuit onchip. The input bias currents are permanently trimmed at wafer testing to a low level. Do not try to balance the input resistances in each input lead; instead, keep the resistance at either input as low as possible for maximum accuracy.

Leakage currents on the PC board can be higher than the LT6010's input bias current. For example, $10G\Omega$ of leakage between a 15V supply lead and an input lead will generate 1.5nA! Surround the input leads by a guard ring, driven to the same potential as the input common mode, to avoid excessive leakage in high impedance applications.

Input Protection

The LT6010 features on-chip back-to-back diodes between the input devices, along with 500Ω resistors in series with either input. This internal protection limits the input current to approximately 10mA (the maximum

allowed) for a 10V differential input voltage. Use additional external series resistors to limit the input current to 10mA in applications where differential inputs of more than 10V are expected. For example, a 1k resistor in series with each input provides protection against 30V differential voltage.

Input Common Mode Range

The LT6010 output is able to swing nearly to each power supply rail (rail-to-rail out), but the input stage is limited to operating between $V^- + 1V$ and $V^+ - 1.2V$. Exceeding this common mode range will cause the gain to drop to zero, however no phase reversal will occur.

Total Input Noise

The LT6010 amplifier contributes negligible noise to the system when driven by sensors (sources) with impedance between $20k\Omega$ and $1M\Omega$. Throughout this range, total input noise is dominated by the $4kTR_S$ noise of the source. If the source impedance is less than $20k\Omega$, the input voltage noise of the amplifier starts to contribute with a minimum noise of $14nV/\sqrt{Hz}$ for very low source impedance. If the source impedance is more than $1M\Omega$, the input current noise of the amplifier, multiplied by this high impedance, starts to contribute and eventually dominate. Total input noise spectral density can be calculated as:

$$v_{n(TOTAL)} = \sqrt{{e_n}^2 + 4kTR_S + (i_nR_S)^2}$$

where $e_n=14nV/\sqrt{Hz}$, $i_n=0.1pA/\sqrt{Hz}$ and R_S the total impedance at the input, including the source impedance.



APPLICATIONS INFORMATION

Offset Voltage Adjustment

The input offset voltage of the LT6010 and its drift with temperature are permanently trimmed at wafer testing to the low level as specified in the electrical characteristic. However, if further adjustment of V_{OS} is desired, nulling with a 50k potentiometer is possible and will not degrade drift with temperature. Trimming to a value other than zero

creates a drift of $(V_{OS}/300\mu V)\,\mu V/^{\circ}C$, e.g., if V_{OS} is adjusted to $300\mu V$, the change in drift will be $1\mu V/^{\circ}C$. The adjustment range with a 50k pot is approximately $\pm 0.9 mV$ (see Figures 1A and 1B). The sensitivity and resolution of the nulling can be improved by using a smaller pot in conjunction with fixed resistors. The configuration shown has an approximate nulling range of $\pm 150\mu V$ (see Figures 2A and 2B).

Standard Adjustment

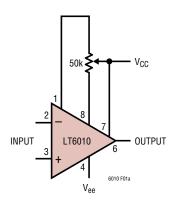


Figure 1A

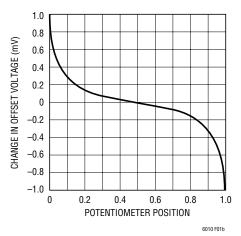


Figure 1B

Improved Sensitivity Adjustment

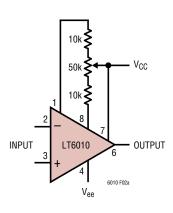


Figure 2A

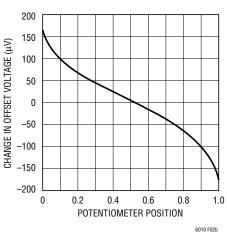


Figure 2B



APPLICATIONS INFORMATION

Shutdown

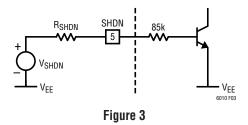
The LT6010 can be put into shutdown mode to conserve power. When the SHDN pin is biased at less than 0.2V above the negative supply, the part operates normally. When pulled 2V or more above V^- , the supply current drops to about $12\mu A$, shutting down the op amp.

The output of the LT6010 op amp is not isolated from the inputs while in shutdown mode. Therefore, this shutdown feature cannot be used for multiplexing applications.

There is an internal 85k resistor at the SHDN pin. If the SHDN voltage source is more than 2V above the negative supply, an external series resistor can be placed between the source and SHDN pin to reduce SHDN pin current (see Figure 3). For an example of suggested values see Table 1. The resistors listed ensure that the voltage at the SHDN pin is 2V above the negative supply.

Table 1

V _{SHDN} (V)	R_{SHDN} ($k\Omega$)
2	NONE
3	77k
4	153k
5	230k



Capacitive Loads

The LT6010 can drive capacitive loads up to 500pF in unity gain. The capacitive load driving capability increases as the amplifier is used in higher gain configurations. A small series resistance between the output and the load further increases the amount of capacitance that the amplifier can drive.

Rail-to-Rail Operation

The LT6010 outputs can swing to within millivolts of either supply rail, but the inputs cannot. However, for most op amp configurations, the inputs need to swing less than the outputs. Figure 4 shows the basic op amp configurations, lists what happens to the op amp inputs and specifies whether or not the op amp must have rail-to-rail inputs. Select a rail-to-rail input op amp only when really necessary, because the input precision specifications are usually inferior.

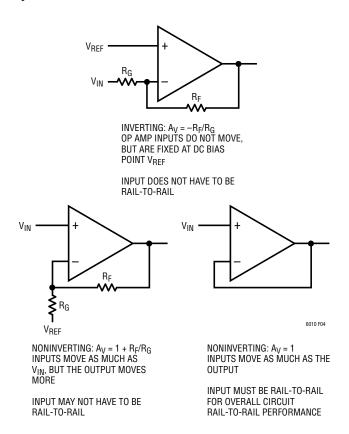
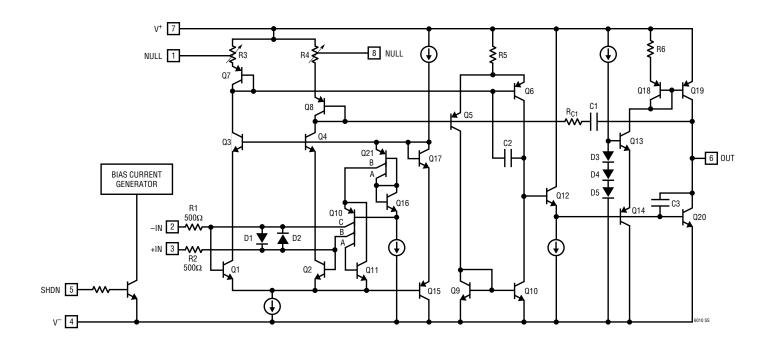


Figure 4. Some Op Amp Configurations Do Not Require Rail-to-Rail Inputs to Achieve Rail-to-Rail Outputs

LINEAR

SIMPLIFIED SCHEMATIC

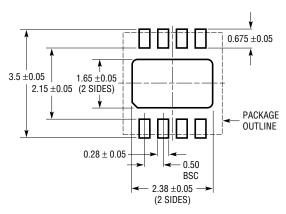




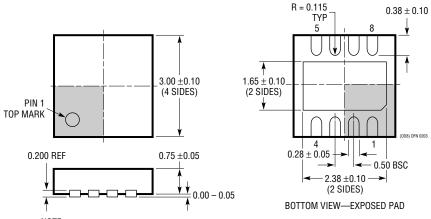
PACKAGE DESCRIPTION

$\begin{array}{c} \textbf{DD Package} \\ \textbf{8-Lead Plastic DFN (3mm} \times 3mm) \end{array}$

(Reference LTC DWG # 05-08-1698)



RECOMMENDED SOLDER PAD PITCH AND DIMENSIONS



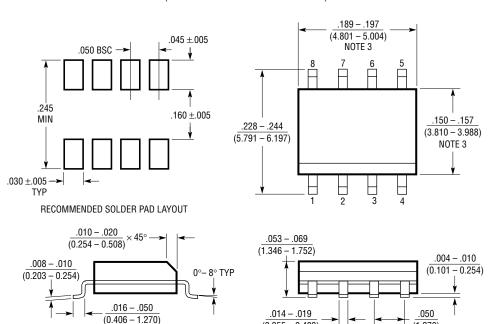
- NOTE:
- 1. DRAWING TO BE MADE A JEDEC PACKAGE OUTLINE MO-229 VARIATION OF (WEED-1)
- 2. ALL DIMENSIONS ARE IN MILLIMETERS
- 3. DIMENSIONS OF EXPOSED PAD ON BOTTOM OF PACKAGE DO NOT INCLUDE MOLD FLASH. MOLD FLASH, IF PRESENT, SHALL NOT EXCEED 0.15mm ON ANY SIDE
- 4. EXPOSED PAD SHALL BE SOLDER PLATED



PACKAGE DESCRIPTION

S8 Package 8-Lead Plastic Small Outline (Narrow .150 Inch)

(Reference LTC DWG # 05-08-1610)



(0.355 - 0.483)

TYP

NOTE: NOTE:
1. DIMENSIONS IN (MILLIMETERS)

2. DRAWING NOT TO SCALE

3. THESE DIMENSIONS DO NOT INCLUDE MOLD FLASH OR PROTRUSIONS.
MOLD FLASH OR PROTRUSIONS SHALL NOT EXCEED .006" (0.15mm)

S08 0303

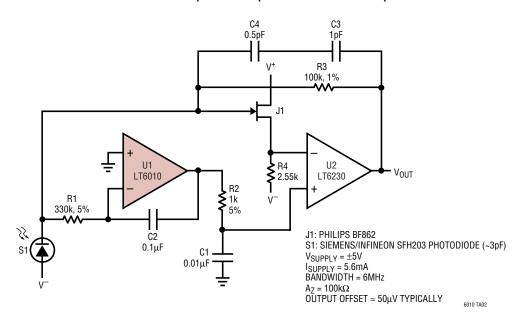
(1.270)

BSC



TYPICAL APPLICATION

Precision JFET Input Transimpedance Photodiode Amplifier



RELATED PARTS

PART NUMBER	DESCRIPTION	COMMENTS
LT6011/6012	Dual/Quad Precision Op Amps	135μA, Rail-to-Rail Output
LT1001	Low Power, Picoamp Input Precision Op Amp	250pA Input Bias Current
LT1880	Rail-to-Rail Output, Picoamp Input Precision Op Amp	C _{LOAD} up to 1000pF